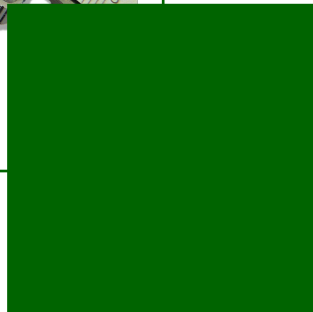




GC-NIP

Datasheet

Version: 1.4
Date: 09/04/2020



Revision history

Date	Revision	Changes
21/05/2013	1.0	Initial version
17/06/2013	1.1	Operating mode changes
03/03/2014	1.2	Characteristic values updated Updated schematics
04/04/2017	1.3	change to new AMAC document layout
09/04/2020	1.4	<p>At Overview switching filter changed to low pass filter At Features the measured delay values added. 4.4 us for two channel and 8 us for nonius. Ordering Information updated AM IPE NONIUS,. Converter BiSS added. Program service added. Figure numbering corrected. Pin assignment and package renamed before package. Pin assignment fit to one page. Package assigned 5.1 Figure Input signals (single ended) update Figure Input signals (differential) update Interpolation rate / nonius pitch description IRDSEL added. Maximum input frequency sorted to match table order. Sources added at Sensor monitoring table. Table Recommended configuration of the sensor monitoring recommended value "00BF" corrected at ABZ interface. Table Configuration options completed, registers matched to length. Interpolation rate / nonius pitch clarification of IRDIV added. At Edge distance control / Interval time tpp / Hysteresis CFG1/DH(2:0) corrected to CFG3/DH(2:0) At Index signal /Reference signal/ Signal Z CFG3/ZPOS(4:0) corrected to CFG2/ZPOS(4:0). Range added. At table ABZ modes IRQ changed to nINT. Adjustment of the reference (index) point with trigger mode update ZMODE. Table Propagation delay analogue (tdANA) corrected to CFGGAIN Table Register overview Access for registers from MVAL to NONIUS completed as read access.</p> <p>At CFG1 description CFGTPP ranges added, TRGSLP 1 rising corrected. IRDIV completed. At CFG2 range added for ZPOS. Validity when CFG3/NOSEL = 0, added. At CFG3 DH description ranges added. IRDIV2 completed. At STAT/ID/REV ASICREV reset value 0010. Configuring tpp and fosz example sorted and corrected. Table tpp added.</p>

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1 Overview

The two-channel interpolation circuit GC-NIP serves to increase the resolution of absolute position and angular measuring systems with two sinusoidal output signals (nonius signal). Aside from the calculation of the absolute position, the GC-NIP may also operate as one- or two-channel incremental measuring system.

The input signals are subjected to an AMAC-specific internal gain and offset control. Additionally, the phase deviation of the input signals can be adjusted statically by a digital potentiometer. By dividing the signal period of the input signals up to 8,192 times, the incremental position on both channels is calculated as well as the absolute position using the nonius calculation. The distance information can be passed on to processing components via a fast SPI interface, an SSI interface, a BiSS interface or by conventional ABZ-square-wave signals.

Inputs and outputs of the GC-NIP are designed for 3.3V interfaces. The IC comprises six instrumentation amplifiers with adjustable gain factors. Encoders with voltage interface or measuring bridges can be connected directly. Sensors with current interface and photodiode-arrays are adapted by a simple external circuit. The IC operates on both single-ended or differential input signals. The noise of the sensor signals is prevented by a configurable low pass analogue filter. Additionally, a digital hysteresis can suppress the edge noise of the output signals at low input frequencies and at standstill. Thus, in case of short-time disturbance of the input signals, a subsequent interpolation counter will operate without errors.

The quality of the signals issued by the sensors is monitored in the IC. For that purpose it is possible to activate 9 sources separately producing an error signal. For the calculation of the absolute position a set of sensor- or scale-specific correction coefficients can be stored in the EEPROM of the IC. In that way, harmonics of the sinusoidal signals or inaccuracies of the measuring scale do not lead to errors in the absolute position value. The designation of the correction coefficients is realized by a simple software-based calibration procedure.

Providing absolute position and incremental square-wave-signals (ABZ) in parallel, the GC-NIP is well-suited for the use in motor-feedback-systems. The four integrated output interfaces (ABZ/SPI/SSI/BiSS) and further features like the multistage trigger signal processing, the processing of distance coded reference marks, the possibility to adjust the reference mark as well as adjustment and storage of the zero position make the IC suitable for direct use in industrial controls or in fast absolute or multichannel incremental position measuring systems. A selectable master SPI interface allows the user to modify the SSI/BiSS-data by providing additional information, for example data from an external multiturn counter or error information.

The GC-NIP can be configured according to specific applications using the integrated EEPROM, via configuration inputs or via the serial interface (SPI/BiSS).

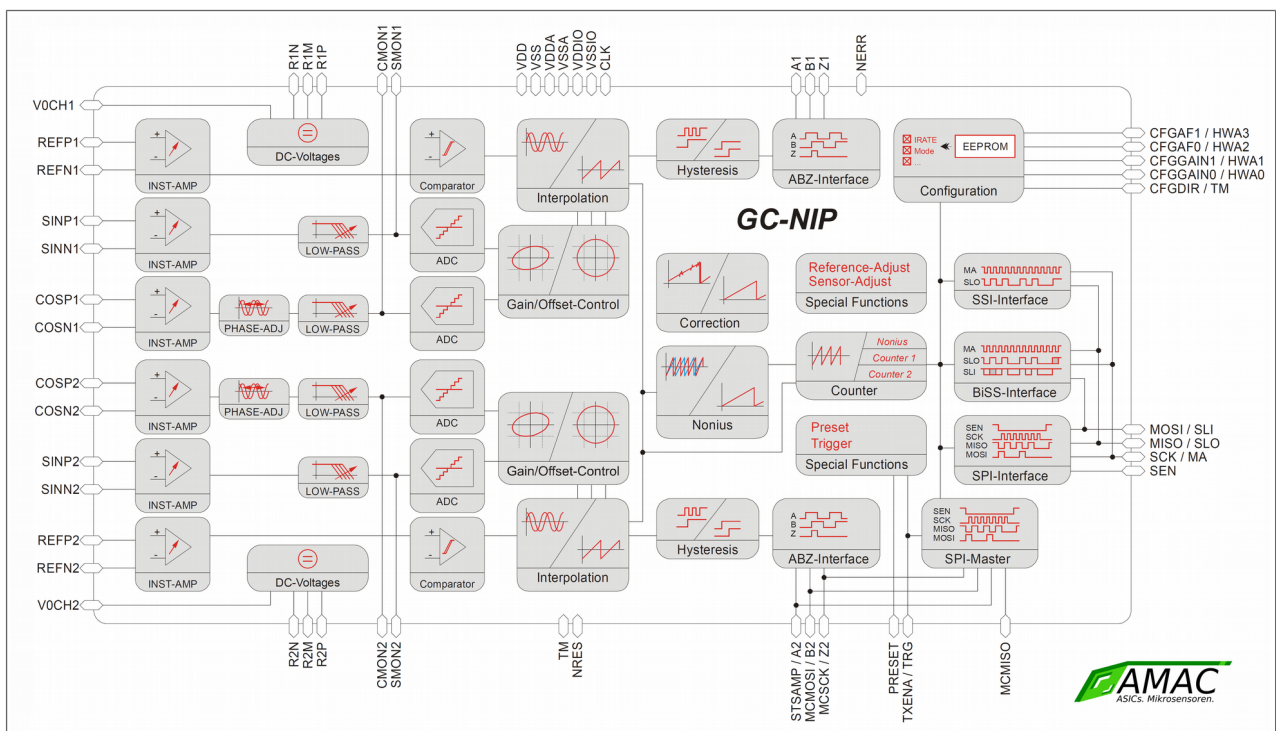


Figure 1: Block diagram

2 Features

Interfaces	
Analogue input	Sinusoidal / cosinusoidal / reference (index) signals, differential or single-ended Adjustable amplification for 660 mV _{PP} / 250 mV _{PP} / 120 mV _{PP} / 60 mV _{PP} Input frequency max. 130 kHz for nonius calculation Input frequency max. 90 kHz for interpolation
ABZ	90° square-wave sequences (A/B/Z) Adjustable width of zero signal Z to ¼ or 1 period A/B Error signal Interrupt signal for external processing Service signals for sensor adjustment
SPI	30-bit counter value for the interpolation channels Up to 22-bit resolution for the absolute position 9-bit sensor status information on each channel Compatible to Standard-SPI: 16-bit, MSB first, up to 15 MHz
SSI and BiSS	Up to 30-bit counter value 2-bit sensor status Gray code / binary code adjustable timing SSI ring operation
Additional inputs	Trigger input for storage of the measured value Preset signal for adjustment and storage of the counter values Reference position alignment using external signal preset impulse
Configuration options	Integrated EEPROM Configuration inputs Serial Interface (SPI/BiSS)

Interpolation / nonius calculation / signal processing	
Interpolation rate	256 to 8192, divisible by 8 Adjustable Divider 1/2/4/8 for the A/B-signals on each channel
Nonius pitch	Number of periods per turn for absolute position calculation Interpolation rate / [8 / 16 / 32 / 64]
Nonius correction	Correction coefficients stored in EEPROM Software based calibration process for determination of the correction coefficients
Signal correction	AMAC-specific digital controller for the offset, control range ±10% of the standard amplitude AMAC-specific digital controller for the amplitude, control range 60% ... 120% of the standard amplitude Digital potentiometer with 64 steps for phase correction; selectable range ±5° or ±10° Input signal monitoring with configurable error indication
Suppression of disturbances	Adjustable low pass filter 10 kHz, 75 kHz, 150 kHz Digital hysteresis for suppression of the edge noise at the output (configurable 0...7) Selectable minimum edge distance at the output (bandwidth limitation)
Reference signal processing	Adjustable reference mark position in 32 steps (0 ... 360°) Optional: high precision alignment of the reference mark position (configuration via external signal possible) Processing of distance coded reference marks Measured-value trigger at the reference mark position
Miscellaneous	Optional Master-SPI interface for output and manipulation of SSI/BiSS-Data 2-stage measured value trigger Constant delay between sampling and measurement value for all resolutions: Two channel 4.4 us, Nonius: 8 us

Important characteristics	
Package	QFN64 (9 x 9 mm)
Operating voltage	3.3V
Temperature range	-40 ... 125 °C
Max interface clock	SPI 15 MHz, BiSS 10 MHz, SSI 5 MHz

3 Ordering Information

Product Type	Description	Article Number
GC-NIP	Interpolation Circuit GC-NIP, QFN64	PR-44800-00
GC-LS	4-channel / Analogue level shifter 5V to 3.3V, QFN32	PR-44500-00
AM-IPE-NONIUS	Demoboard for Interpolation circuit GC-NIP	PR-44820-00
USB to SPI converter	USB adapter for the SPI interface	PR-44025-10
USB to BiSS converter	USB adapter for the BiSS interface	PR-44030-00
Programming service	User specific programming of EEPROM (possible for a number of units)	360-060

4 Typical applications

Table 1: Applications overview

Signal form (Sensor)	Application of GC-NIP
Sinusoidal, Voltage	Direct connection of GC-NIP to sensor.
Sinusoidal, Current	Additional resistors required
Reference- (Index-) Track	Direct connection of GC-NIP to sensor.
Square wave	IC is not suitable in principle.
Signal Form (Sensor)	Application of GC-NIP
1 V _{pp} nominal	Use GC-LS for signal conversion or external resistors.
660 mV _{pp} nominal	Direct connection of GC-NIP to sensor.
330 mV _{pp} nominal	Use GC-LS for signal conversion or external resistors.
250 mV _{pp} nominal	Direct connection of GC-NIP to sensor.
120 mV _{pp} nominal	Direct connection of GC-NIP to sensor.
80 mV _{pp} nominal	Use GC-LS for signal conversion or external resistors.
60 mV _{pp} nominal	Direct connection of GC-NIP to sensor.
2 V _{pp} nominal	External resistors required (see 11).
Differential signal, DC-Reference Voltage 0.82...1.8V	Direct connection of GC-NIP to sensor.
Single-Ended, DC-Reference Source in Sensor	Direct connection of GC-NIP to sensor.
Single-Ended, DC-Reference Source not in Sensor	Direct connection of GC-NIP to sensor.
Photodiodes 0.5 μA _{pp}	External resistors required. (see 11).
Photodiodes 11 μA _{pp} ...16 μA _{pp}	External resistors required. (see 11).
Resistive bridges (magnetic sensors)	Direct connection of GC-NIP to sensor.
Unstable amplitude of sensor	GC-NIP contains automatic controller for amplitudes.
Offset not correctable at sensor	GC-NIP contains automatic controller for offset.
Phase not correctable at sensor	GC-NIP contains potentiometer for phase correction.
Variable Reference mark position	Reference mark position is freely adjustable.
Distance coded reference marks	Evaluation support via SPI using the internal trigger-function.
Subsequent processing	Application of GC-NIP
Interface to microcontroller/DSP/FPGA	Use SPI-Interface
Interface to external interpolation counter	ABZ-Interface for both channels
Usage on industrial control	SSI, BiSS or ABZ-Interface
System includes more than one channel	Two incremental interpolation channels included. Possibility to use simultaneously on single SPI-Bus
Real-Time-System / equidistantly measurement	Constant propagation delay for all resolutions, trigger input
IC-Configuration	Integrated EEPROM, all registers are configurable via SPI/BiSS
Signal specification LVCMOS	In-/Outputs used directly
Signal specification RS422	Driver-IC required
Maximum signal frequencies (f _{osz} = 26 MHz)	
Rotary encoder: $f_{\max} = (\text{revolutions / minute}) \cdot (\text{signal periods / revolution}) / 60$	
Linear encoder: $f_{\max} = (v_{\max} [\text{in m/s}] / (\text{signal periods} [\text{in mm}])) \cdot 1000$	
f _{max} < 150 kHz	All interpolation rates up to 8,192 via SPI/SSI/BiSS
f _{max} < 23 MHz / Interpolation rate	If ABZ-outputs are used
Max frequency of interpolation counter on ABZ is known	Configuration of the minimum edge distance possible via CFGL/CFGTPP

5 Pin assignment and package

5.1 Pin assignment

Table 2: Pin assignment QFN64

Pin	Name	Type	Description
1	VDDA	AP	Supply voltage Analogue +3.3V
2	VSSA	AP	Analogue GND
3	R2P	AO	ADC2-reference voltage high
4	R2M	AO	ADC2-reference voltage mid
5	R2N	AO	ADC2-reference voltage low
6	SMON2	AO	Monitor output at instrumentation amplifier sine channel 2
7	CMON2	AO	Monitor output at instrumentation amplifier cosine channel 2
8	N.C.	n.c.	Do not connect
9	VDD	DP	Supply voltage digital +3.3V
10	VSS	DP	Digital GND
11	CFGAF1/HWA3	DI / Pull-Down	Configuration Analogue filter / HWA<3>
12	CFGAF0/HWA2	DI / Pull-Down	Configuration Analogue filter / HWA<2>
13	SCK/MA	DI / Pull-Down	SPI/BiSS/SSI: clock
14	SEN	DI / Pull-Up	SPI: select / during Reset: select interface SPI / BiSS or SSI
15	MOSI/SLI	DI / Pull-Down	SPI/BiSS: data in GC-NIP
16	MISO/SLO	DO / Open-Drain	SPI/BiSS/SSI: data out GC-NIP
17	N.C.	n.c.	Do not connect
18	VDDIO	DP	Supply voltage digital (IO) +3.3V
19	VSSIO	DP	Digital GND
20	MCCK / Z2	DO / Tristate	Controller interface – clock / Output for Zero signal Z (reference signal)
21	MCMOSI / B2	DO / Tristate	Controller interface – data out / Incremental output B channel 2
22	STSAMP / A2	DO / Tristate	Controller interface – sync signal / Incremental output A channel 2
23	MCMISO	DI	Controller interface – data in
24	TXENA / TRG	DI	Controller interface – enable / trigger
25	CLK	DI	Clock input
26	PRESET	DI / Pull-Up	Input for the preset function
27	CFGDIR / TM2	DI / Pull-Down	Configuration counter direction for nonius method/ Test mode select, if TM = high
28	Z1	DO / Tristate	Output for Zero signal Z (reference signal) channel 1
29	B1	DO / Tristate	Incremental Output B channel 1
30	A1	DO / Tristate	Incremental Output A channel 1
31	VSSIO	DP	Digital GND
32	VDDIO	DP	Supply Voltage digital (IO) +3.3V
33	NERR	DO / Open-Drain	Error signal
34	CFGGAIN0/HWA0	DI / Pull-Down	Configuration Gain / HWA<0>
35	CFGGAIN1/HWA1	DI / Pull-Down	Configuration Gain / HWA<1>
36	TM	DI	Test mode; Connect to VSS
37	NRES	In-/Output Analogue; Pull-Up	Reset
38	VS	AP/AI	Test EEPROM
39	VCG	AP/AI	Test EEPROM
40	VSS	DP	Digital GND
41	VDD	DP	Supply voltage digital +3.3V
42	CMON1	AO	Monitor Output at instrumentation amplifier cosine channel 1
43	SMON1	AO	Monitor Output at instrumentation amplifier sine channel 1
44	R1N	AO	ADC1-reference voltage low
45	R1M	AO	ADC1-reference voltage mid
46	R1P	AO	ADC1-reference voltage high
47	VSSA	AP	Analogue GND
48	VDDA	AP	Supply voltage Analogue +3.3V
49	N.C.	n.c.	Do not connect
50	REFP2	AI	Input reference signal positive channel 2
51	REFN2	AI	Input reference signal negative channel 2
52	REFP1	AI	Input reference signal positive channel 1
53	REFN1	AI	Input reference signal negative channel 1
54	V0CH1	AO	Mean voltage channel 1 for sensor supply
55	SINP1	AI	Sinusoidal signal at input, positive channel 1
56	SINN1	AI	Sinusoidal signal at input, negative channel 1
57	COSN1	AI	Cosinusoidal signal at input, negative channel 1
58	COSP1	AI	Cosinusoidal signal at input, positive channel 1
59	COSP2	AI	Cosinusoidal signal at input, positive channel 2
60	COSN2	AI	Cosinusoidal signal at input, negative channel 2
61	SINN2	AI	Sinusoidal signal at input, negative channel 1
62	SINP2	AI	Sinusoidal signal at input, positive channel 1
63	V0CH2	AO	Mean voltage channel 2 for sensor supply
64	N.C.	n.c.	Do not connect
EXP		Exposed Pad	

AP = Analogue power supply, DP = digital power supply, AI = Analogue input, AO = Analogue output

5.2 Package

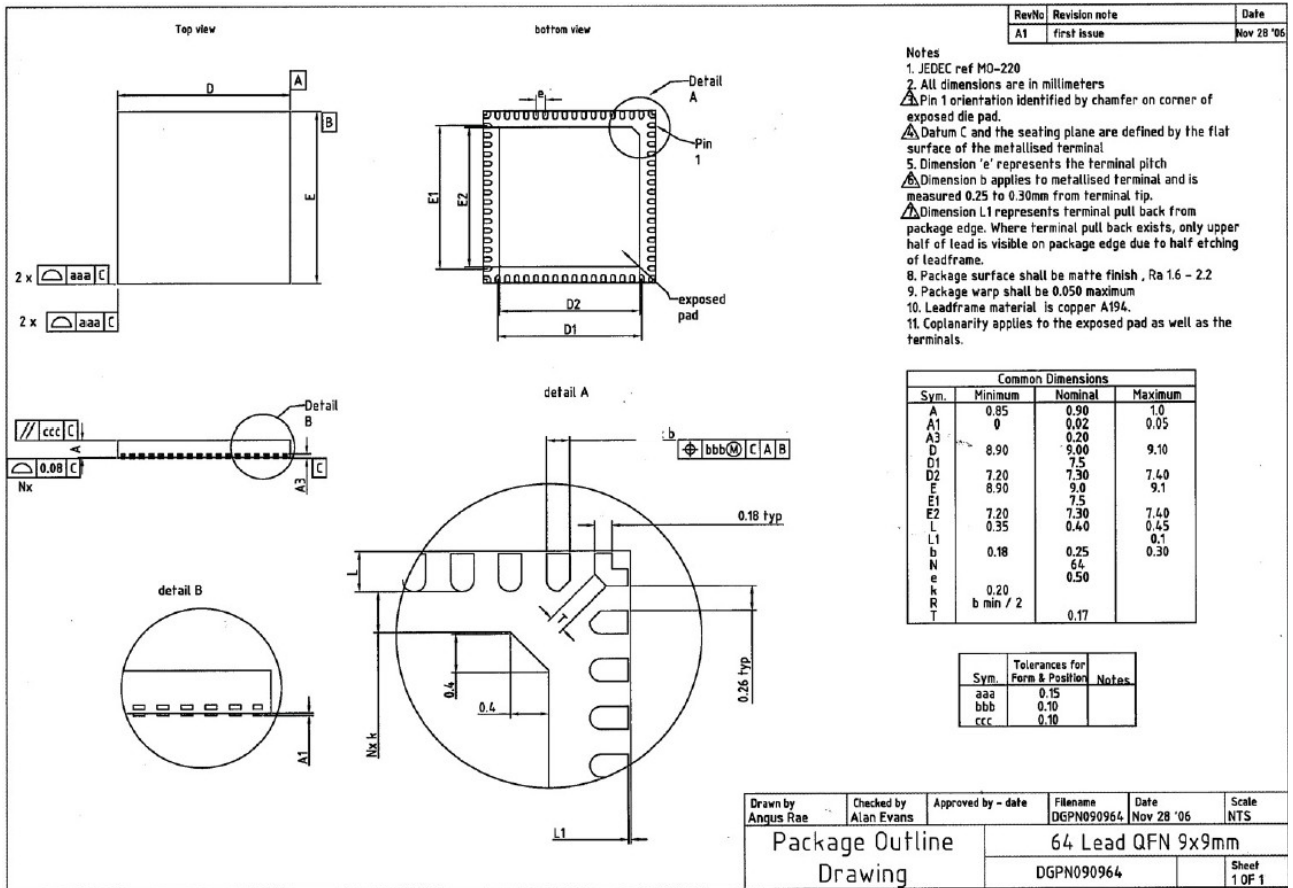


Figure 2: Package QFN64 drawing

6 Start up Behaviour / Configuration Options

6.1 Reset

During reset of the IC, the digital interface is selected (SPI or SSI/BiSS) and all registers are initialized with their default values. The initialization of the circuit is performed either from the internal EEPROM or from the configuration pins. The internal EEPROM has to be programmed with a valid identifier at EEPROM address 0x00 in order to be used for configuration after reset. The configuration of the interpolation rate is either done from the EEPROM (if valid) or with a fixed interpolation rate of 2,000. Another valid identifier on EEPROM address 0x01 decides, if the correction coefficients – also located in the EEPROM of the IC – are loaded during reset and later be used for absolute position calculation.

During the whole reset sequence, a pin - dependent on the selected interface - NERR or MISO, is maintained at L level. Up to this point, the serial interfaces may not be activated. Subsequently, the configuration registers can be modified using the SPI- or BiSS-interface. The following tables provide an overview of the configuration possibilities for the GC-NIP.

Table 3: Selection of the serial interface

Interface	Pin SEN reset value	BIT SSI ¹⁾	Pin MISO / SLO	Pin MOSI / SLI	Pin SCK	Pin SEN	Ready-Signal
SPI	1	any value	SPI-MISO	SPI-MOSI	SPI-SCK	SPI-SEN	at MISO
SSI	0	1	SSI-DATA	-	SSI-MA	0	at NERR
BiSS	0	0	BISS-SLO	BISS-SLI	BiSS-MA	0	at NERR

¹⁾ Register CFGBiSS / Bit 31

Table 4: Switching the configuration source

Content of EEPROM address 0x00	Configuration
unequal 0x134A	Basic configuration / see Tables 8 and 9 (Pin)
0x134A	read from EEPROM / Tables 8 and 9 (EEPROM)

Table 5: Configuration of the interpolation rate and nonius pitch

Content of EEPROM address 0x00	Bit IRMAP ¹⁾	Configuration source	Interpolation rate IRATE	Nonius pitch (Signal on channel 1)	Interpolation rate ABZ
unequal 0x134A	default value: 1	Manufacturer EEPROM	2000	125	2000
0x134A	0	EEPROM	any between [256 ... 8192] divisible by 8	IRATE / [16,32,64,128]	IRATE / [1,2,4,8]
0x134A	1	Manufacturer EEPROM	2000	125	2000 / [1,2,4,8]

¹⁾ Register CFG1 / Bit 3

Table 6: Configuration correction coefficients for nonius calculation

Content of EEPROM address 0x01	correction coefficients value
unequal 0x134A	All are '0'
0x134A	Read from EEPROM

6.2 Configuration

The IC can be matched to most varied measuring systems and subsequent electronic systems via the configuration registers. If the IC is initialized using the integrated EEPROM or a serial interface (SPI/BiSS), full configuration possibilities are available. If the initialization is performed via the configuration pins, selected parameters can be set externally. Table 7 below provides an overview of the configuration possibilities of the GC-NIP. Detailed description of the configuration register set can be found in section 9 of this document.

Table 7: Configuration options

Parameter	Possible values	Register / Bit
Interpolation rate IRATE	Configurable interpolation table in EEPROM Alternative: fixed table in manufacturer-PROM Divider for square-wave-signals (ABZ) and counter Separate divider for channel 1 and 2	From EEPROM CFG1 / IRMAP CFG1 / IRDIV1(1:0) CFG1 / IRD2SEL, CFG3 / IRDIV2(1:0)
Operating mode	Nonius + ABZ 2 Channel, Nonius + DSP, Calibration Mode, etc.	CFG1 / MODE(3:0)
Nonius pitch	Interpolation rate / [8 / 16 / 32 / 64]	From EEPROM
Nonius correction	Correction values Correction value resolution Activate / deactivate correction from EEPROM	From EEPROM CFG3 / MXSHR CFG3 / MXFEED
Min. edge interval t_{pp}	1, 2, 4, 8, 16, 32, 64, 128	CFG1 / CFGTPP(2:0)
Reference point	Enable / disable, delayed Width 1 Increment / 4 Increments Position 0°-360°, step size 11.25° Mode reset, trigger, adjust, distance coded Position 0°-360°, step size 360°/IRATE ¹⁾	CFG3 / DISZ(1:0), CFG3 / ZDEL(1:0) CFG2 / Z4 CFG2 / ZPOS (4:0) CFG2 / ZMODE(1:0) CFG3 / NOSEL, NONOFFS
Nominal signal amplitude	660 mV _{pp} , 250 mV _{pp} , 120 mV _{pp} , 60 mV _{pp}	CFG1 / GAIN(1:0) alternative: Pins CFGGAIN(1:0)
Low pass filter (1dB)	150kHz, 75kHz, 10kHz (all +/-10%), Disable	CFG1 / CFGAF(1:0) alternative: Pins CFGAF(1:0)
Digital hysteresis	0 (Disable), 1 ... 7	CFG3 / DH(2:0)
Output signals A/B/Z	ABZ-Mode, DSP-Mode, sensor adjustment, Reference mark adjustment Optional Master-SPI on ABZ-channel 2	CFG1 / MODE(2:0) CFG1 / MODE(3)
Error processing	Masking, latch enable Output configuration in case of errors for ABZ	CFG1 / Mxxx, Lxxx CFG1 / HLD, TRI
Phase correction	± 10° step width 0.15°, ±5° step width 0.08°	CFG2 / PH_1(5:0) CFG2 / PH_2(5:0) CFG2 / PHBER,
Gain controller	Default setting / time constant / enable, disable	CNTRLG, CFG3 / GAINCTL(1:0), DISCTL
Offset controller	Default setting / time constant / enable, disable	CNTRLO, CFG3 / OFFSCTL(1:0), DISCTL
Hardware address	0-15	CMD / SETHWA Pins HWA(3:0)
Special functions	Trigger pulse edge rising/falling Preset function active / inactive Absolute position offset Counter preset position Absolute counter direction	CFG1 / TRGSLP CFG2 / PREENA NONOFFS, PREST2 PREST1,PREST2 Pin CFGDIR
Interface configuration	Format of the position values Resolution of the position, position value select SPI-Mode synchronous, asynchronous BiSS interface active, inactive SSI-Timing, ring mode BiSS-Timing BiSS data format 8Bit, 32 Bit SPI for manipulation of the SSI/BiSS-Data	CFGBISS / SSI20, GRAY CFGBISS / STBIT(4:0), STSEL(1:0) CFG2 / ASYNC, SYNC(6:0) CFGBISS / SSI CFGBISS / SSITO(11:0), RING CFGBISS / BISSTO(4:0) CFGBISS / READ32 CFG1 / MODE(3)
Power saving options	Deactivation of the monitor outputs Deactivation V0-Pins Deactivation of channel 2 Deactivation of the nonius correction Deactivation of the reference mark processing	CFG2 / DISMON CFG2 / DISV0 CFG3 / DISCH2(1:0) CFG3 / MXFEED CFG3 / DISZ1, DISZ2

¹⁾ The reference value is set from NONOFFS either at noius or two-channel mode

Table 8: Default configuration

Configuration	Default (EEPROM with factory settings)		Default (Pin)	
Analogue	Phase correction	0°	Phase correction	0°
	Low pass -1dB	150 kHz	Low pass -1dB	configured via pin
	Nominal signal amplitude	660 mVpp	Nominal signal amplitude	configured via pin
	Power saving options	inactive	Power saving options	inactive
Interpolation Nonius	Interpolation rate	8000	Interpolation rate	2000
	Controller	active, timing 01	Controller	active, Timing 01
	Controller start values	Average	Controller start values	Average
	Reference mark position	at 45°	Reference mark position	at 45°
	Nonius pitch	125	Nonius pitch	125
	Correction	none	Correction	none
	Count direction	configured via pin DIR	Count direction	configured via pin DIR
	Power saving options	inactive	Powers saving options	inactive
ABZ output signals	Mode	ABZ	Mode	ABZ
	TPP	0	TPP	0
	Digital hysteresis	1	Digital hysteresis	1
	Z	active, 1 increment	Z	active, 1 increment
	Output in case of error	Hold	Output in case of error	Hold
	Power saving options	inactive	Power saving options	inactive
Error processing	Error monitoring	all errors	Error monitoring	all errors
	Error storage	inactive	Error storage	inactive
Special functions	Preset (Nonius)	inactive	Preset (Nonius)	always inactive
	Preset values	0x00	Preset values	0x00
	Nonius offset	0x00	Nonius offset	0x00
	Trigger pulse edge	falling	Trigger pulse edge	falling

Table 9: Default configuration interfaces

Configuration	Default (EEPROM with factory settings)		Default (Pin)	
SPI interface	Activate via Pin SEN Hardware-address at $HWA(3:0)$		Activate via Pin SEN Hardware-address at $HWA(3:0)$	
SSI interface	Activate via Pin SEN Timeout 20 μs @ 26 MHz, Ring mode Format 20Bit Direct output (no Simple-SPI)		Activate via Pin SEN Timeout 20 μs @ 26 MHz, Ring mode Format 20Bit Direct output (no Simple-SPI)	
BiSS interface	Hardware-address at $HWA(3:0)$ Timeout 19.7 μs @ 26MHz Format 30Bit Singleturn Direct output (no Simple-SPI)		Hardware-address at $HWA(3:0)$ Timeout 19.6 μs @ 26 MHz Format 30Bit Singleturn Direct output (no Simple-SPI)	

7 Functional description

7.1 Input amplifier / Low pass filter

The GC-NIP incorporates six instrumentation amplifiers with adjustable gain factors. Incremental encoders with voltage interface and measuring bridges can be connected directly. Sensors with current-interface are adapted via a simple external circuit (see 11.1). The IC operates with both, single-ended and differential input signals. The amplification is identical for all signals of the sensor (sinusoidal, cosinusoidal, index/reference). To adapt the GC-NIP to customized sensors, the mean voltage of the instrumentation amplifiers for each channel is provided at pins V0CH1 and V0CH2.

The instrumentation amplifiers are connected to the internal A/D converters. Alternatively, this connection is done directly or via a configurable low-pass filter. The cut-off frequencies given in Table 11 are achieved with an accuracy of +/-10%. The conversion range of the analogue digital converter and the reference voltages of the instrumentation amplifiers are pre-adjusted, so that internal offset error is already compensated. The signals on the input of the analogue digital converters can be monitored using the pins SMON1, CMON1, SMON2, and CMON2¹.

Table 10: Configuration signal amplitude (nominal) (Register CFG1)

CFG1/GAIN(1:0)	00	01	10	11
Input voltage for differential supply ¹⁾ (mV _{pp})	330	125	60	30
Input voltage V _{DiffNom} nominal (mV _{pp})	660	250	120	60
Input voltage range for V _{Diff} (mV _{pp})	400...800	150...300	75...145	36...72
Input voltage for maximal ADC-range V _{DiffMAX} (mV _{pp})	990	375	180	90
Reference voltage on V0CH1 nominal	1.1	1.1	1.1	1.1
Output voltage V _{MON} nominal on SMON / CMON (V _{pp})	1.27	1.27	1.27	1.27
Amplification (V _{MON} / V _{DIFF})	1.92	5.08	10.6	21.2

¹⁾ at each of the inputs SINP, SINN, COSP, COSN

Table 11: Configuration low pass filter (Register CFG1)

Cut-Off frequency -1dB	CFG1/CFGAF(1:0)
150 kHz (-0.5dB)	00
75 kHz	01
10 kHz	10
low-pass disabled	11

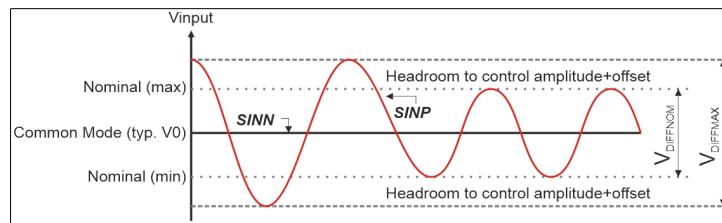


Figure 3: Input signals (single ended)

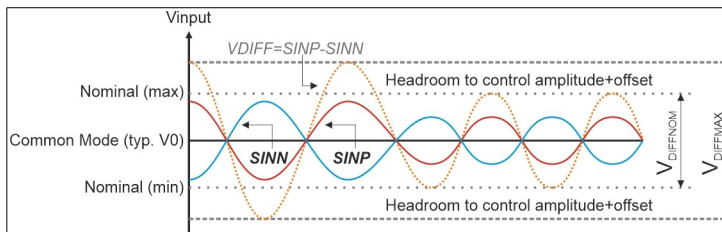


Figure 4: Input signals (differential)

1 The analogue low-pass filter must be activated when using the monitor outputs (CFG1/CFGAF ≠ '11').

The input voltage for the instrumentation amplifiers is limited in a range from $V_{in}=0.35V$ to $VDDA-1.00V$. According to the common mode voltage at the analogue input, this may limit the operating range for the Gain-Configuration "00" ($V_{NOM} = 660 mV_{pp}$).

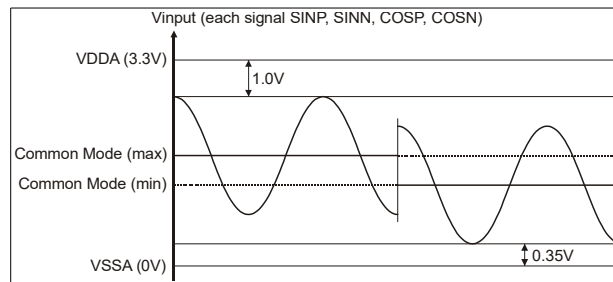


Figure 5: CMIR (input signals)

Exemplary, the following table shows combinations for common mode voltage and supply voltage for single-ended input signals with maximum amplitude of $880mV_{pp}$ and maximum offset of $\pm 70 mV$:

Table 12: Example for common mode input voltage (CMIR)

VDDA	Common-mode voltage (Min)	Common-mode voltage (Max)
3.30V	0.82V	1.83V
3.15V	0.82V	1.63V
3.00V	0.82V	1.53V

7.2 Signal correction

The input signals are subjected to an AMAC-specific internal gain and offset control. The amplitudes are controlled in the range between 60 % and 120 % of the standard amplitude. The control range for the offset of the two input signals is $\pm 10\%$ of the nominal amplitude. The phase displacement of the input signals can be corrected statically in 64 steps using a digital potentiometer. The setting range of the phase is set to approx. $\pm 5^\circ$ or approx. $\pm 10^\circ$ via configuration bit.

After resetting the IC, start values to correct amplitude and offset of the input signals are loaded from the EEPROM. If the EEPROM content is not marked valid (see 4), these values are set to the center of the control range. The full measuring accuracy of the IC, however, is only achieved after settling of the internal signal control which is about 10...50 periods of the input signals. Alternatively, start values for the controller can be stored in EEPROM, so that the settling time of the controller after reset of the IC can be shortened.

To achieve the maximum possible accuracy in the amplitude and offset control, the phase potentiometer must be matched with the sensor connected to the GC-NIP. Amplitude and offset errors are treated as a unit in the GC-NIP. This means that for particular applications a larger permissible error may be permitted for the respectively other parameter under certain circumstances. The attenuation of the controlled system implemented in the GC-NIP can be adjusted (registers CFG3/GAINCTL and CFG3/OFFSCTL).

Table 13: Signal correction

Parameter	as a percentage referred to the nominal amplitude (PEAK-PEAK)	as a percentage referred to the ADC-maximum (PEAK-PEAK)	in mV referred to the standard signal (0.66 V _{pp})	in V on the pin SMON and CMON (PEAK-PEAK)
Maximal value at the input (V_{maxpp})	150	100	990	1.90
Nominal value of the input signal (V_{nompp})	100	66.7	660	1.27
Guaranteed control range for the amplitude	60... 120	40... 80	400... 800	0.76 ... 1.52
Setting range of the amplitude controller	56... 168 ¹⁾	38... 112 ¹⁾	370... 1110 ¹⁾	0.71 ... 2.13 ¹⁾
Vector monitoring ²⁾	30	20	200	0.38
Guaranteed control range for the offset (sensor)	± 15	± 10	± 70	± 0.133
Setting range of the offset controller	± 25	± 17	± 165	± 0.315

¹⁾ The setting range for the amplitude is greater than the control range of the ADC.

²⁾ An aggregate signal from sine and cosine is monitored. See chapter 7.5 Bit V_{LOW}

7.3 Interpolation / nonius calculation

The signal periods of the analogue sinusoidal and cosinusoidal signals are divided according to the selected interpolation rate and provided to the serial interfaces (SPI/SSI/BISS) as phase and count value. In parallel, square-wave sequences with 90° phase shift (A/B/Z signals) are generated.

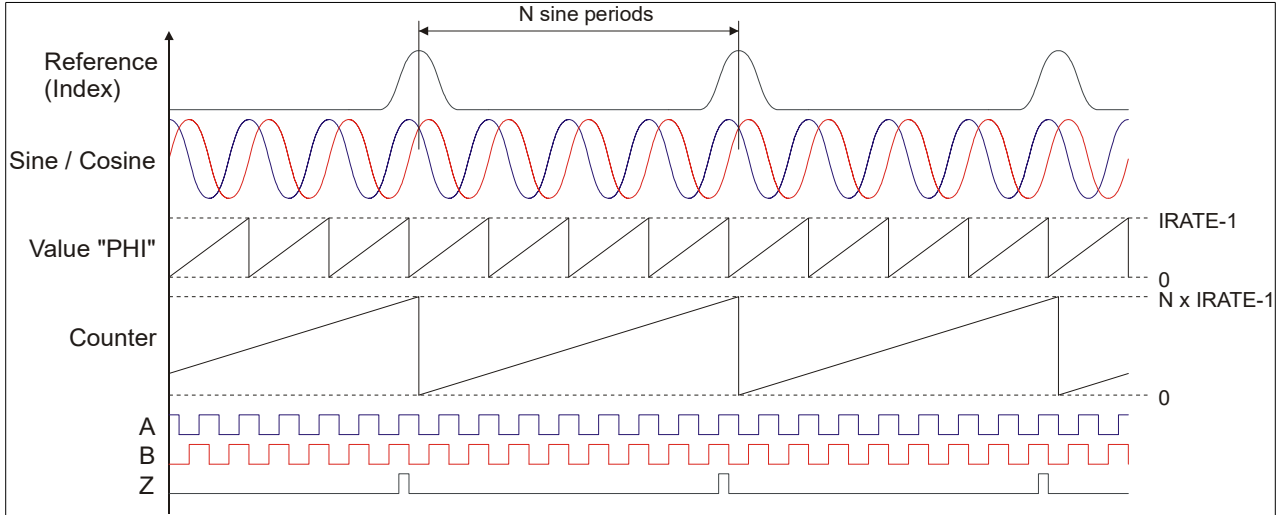


Figure 6: Interpolation

Using the phase values of the two channels and the vernier scale (nonius) method, the absolute position of the sensor is determined on the measuring scale. Errors of the sensor signal or resulting from inaccuracies of the measuring scale can be suppressed via an integrated correction. Therefore, 16 correction coefficients, determined by a software-based calibration algorithm, can be stored in the IC's internal EEPROM.

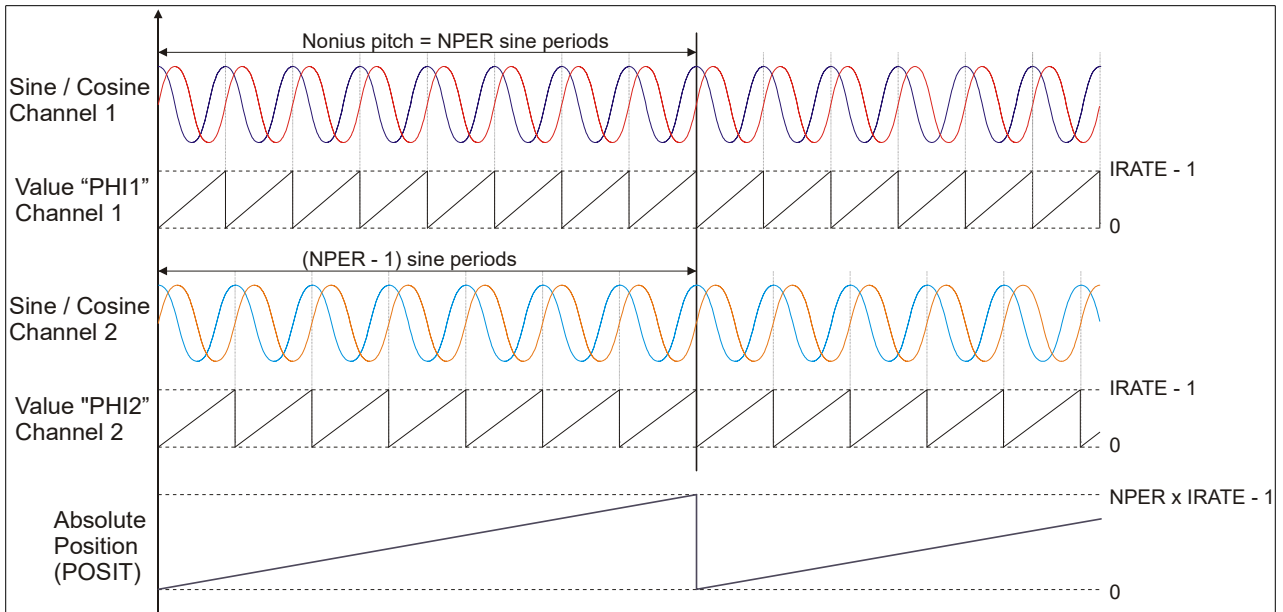


Figure 7: Nonius calculation

Table 14: Operating modes of the GC-NIP

Mode	CFG1/MODE(3:0)	Sensor type	Measuring values
Nonius + ABZ	X000	Nonius sensor	Register CNT1: Incremental position channel 1 Register CNT2: Incremental position channel 2 Register POSIT: Absolute position ¹⁾ SSI/BiSS: Absolute position ¹⁾ ABZ1: Incremental signals channel 1 ABZ2: Incremental signals channel 2
Two channel	X000	Two independent sensors	Register CNT1: Incremental position channel 1 Register CNT2: Incremental position channel 2 Register POSIT: Incremental position 1 and/or 2 ¹⁾ SSI/BiSS: Incremental position 1 and/or 2 ¹⁾ ABZ1: Incremental signals channel 1 ABZ2: Incremental signals channel 2
Calibration	0101	Nonius sensor	Register CNT1: Incremental position channel 1 Register CNT2: Incremental position channel 2 Register POSIT: Absolute position ¹⁾ SSI/BiSS: Absolute position ¹⁾ ABZ1: Test signals for sensor adjustment ABZ2: Test signals for sensor adjustment

¹⁾ The content of register POSIT is selected via CFGBiSS/STSEL (1:0) (see Register description)

7.3.1 Interpolation rate / nonius pitch

The term “interpolation rate” (IRATE) is here understood as the number of increments, into which the sinusoidal/cosinusoidal period of the input signals is divided. “Nonius pitch” describes the number of periods of the input signals, where the absolute position can be clearly assigned using the vernier (nonius) method. Possible interpolation rates for the nonius calculation of the GC-NIP can be selected between 256 and 8,192. And the defined value must be divisible by 8. The nonius pitch (NPER) is selectable between the values IRATE/8, IRATE/16, IRATE/32, or IRATE/64.

Additionally, the interpolation rate for the integrated interpolation counters and the square-wave-signal outputs (A/B) can be divided by a selectable factor (IRDIV) of 1, 2, 4, or 8 (each channel independently). If IRDSEL= 0, the IRDIV1 divider is applied to both channels. When IRD2SEL = 1, IRDIV1 and IRDIV2 are applied to channel one and two correspondingly. The notation IRDIV will refer to IRDIV1 if not stated differently.

The divided interpolation rate of the incremental counters corresponds the number of signal transitions at the A/B outputs per input signal period. The number of square-wave periods at the outputs A and B amounts to ¼ of the divided interpolation rate.

Following table shows possible combinations and limitations of interpolation rate and nonius pitch for different interfaces and use cases.

Table 15: Selecting interpolation rate and nonius pitch

Interface / use case	Interpolation rate	Requirement/Limitation	Possible values for nonius pitch
Single turn Nonius	IRATE from EEPROM 256 ... 8,192	IRATE is divisible by 8	NPER = IRATE / DIV DIV = [8, 16, 32, 64] If DIV = 8: IRATE ≤ 4096 ¹⁾
Internal interpolation counter	IRATE from EEPROM / IRDIV IRDIV = [1, 2, 4, 8]	IRATE is divisible by 8	No influence
A/B-Output	IRATE from EEPROM / IRDIV IRDIV = [1, 2, 4, 8] IRDIV2 = [1, 2, 4, 8] (IRD2SEL = 1)	IRATE is divisible by 8 IRATE/IRDIV is divisible by 4	No influence

¹⁾ This means that the NPER max. value is 512

Example 1

The resolution of the absolute position should be at least 17 bit
 The IC is used with measuring scales with a nonius pitch of 50 ... 70 (channel 1)
 The incremental resolution, using the AB-signals for a motor controller should be at least 10 bit

NPER	IRATE	DIV	Resolution	Bit	IRDIV	IRATE (ABZ)	NPER	IRATE	DIV	Resolution	Bit	IRDIV	IRATE (ABZ)
50	3200	64	160000	17.29	2	1600	61	3904	64	238144	17.86	2	1952
51	3264	64	166464	17.34	2	1632	62	3968	64	246016	17.91	2	1984
52	3328	64	173056	17.40	2	1664	63	4032	64	254016	17.95	2	2016
53	3392	64	179776	17.46	2	1696	64	2048	32	131072	17.00	2	1024
54	3456	64	186624	17.51	2	1728	65	2080	32	135200	17.04	2	1040
55	3520	64	193600	17.56	2	1760	66	2112	32	139392	17.09	2	1056
56	3584	64	200704	17.61	2	1792	67	2144	32	143648	17.13	2	1072
57	3648	64	207936	17.67	2	1824	68	2176	32	147968	17.17	2	1088
58	3712	64	215296	17.72	2	1856	69	2208	32	152352	17.22	2	1104
59	3776	64	222784	17.77	2	1888	70	2240	32	156800	17.26	2	1120
60	3840	64	230400	17.81	2	1920							

Example 2

The maximum resolution for the absolute position should be achieved
 The IC is used with measuring scales with a nonius pitch of 30 ... 40 (channel 1)
 The maximum interpolation rate for the AB-output is 128. This value has been calculated using the maximum input frequency and the maximum output frequency for the ABZ-outputs (see chapter 7.4).

NPER	IRATE	DIV	Resolution	Bit	IRDIV	IRATE (ABZ)	NPER	IRATE	DIV	Resolution	Bit	IRDIV	IRATE (ABZ)
30	960	32	28800	14.81	8	120	36	576	16	20736	14.34	8	72
31	992	32	30752	14.91	8	124	37	592	16	21904	14.42	8	74
32	1024	32	32768	15.00	8	128	38	608	16	23104	14.49	8	76
33	528	16	17424	14.09	8	66	39	624	16	24336	14.57	8	78
34	544	16	18496	14.17	8	68	40	640	16	25600	14.64	8	80
35	560	16	19600	14.26	8	70							

7.3.2 Edge distance control / Interval time t_{pp} / Hysteresis

The minimum time interval t_{pp} , at which the output signals A, B, and Z may switch, can be adjusted in binary steps between $1/f_{OSZ}$ and $128/f_{OSZ}$ using the configuration bits $CFG1/CFGTPP(2:0)$. After switching one of the outputs, the subsequent edge of the other signal will only be visible at the IC output after time t_{pp} has elapsed. Thus, in case of a short-time disturbance of the input signals, a subsequent interpolation counter will operate without errors. The configuration of the edge interval t_{pp} depends on the counter connected to A, B, and Z (see section 11.6). Please note the discretization of time at the output of the IC due to the edge interval setting.

The GC-NIP uses a digital interpolation method. This causes the speed-proportional A/B/Z output signals to be overlaid by the inevitable quantization errors (the so called $\pm 1INK$ errors) resulting from the A/D converters. The quantization noise can be suppressed by activating the digital hysteresis using register $CFG3/DH(2:0)$. This prevents switching of the outputs with static input signals. In this case, all output signals are delayed by one increment.

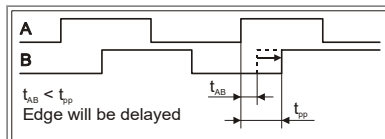


Figure 8: Edge interval setting

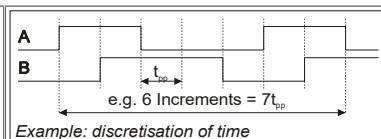


Figure 9: Discretization of time

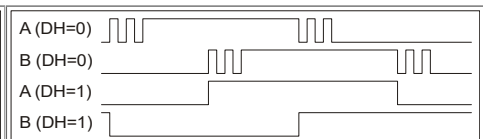


Figure 10: Digital hysteresis

7.3.3 Index signal /Reference signal/ Signal Z

The reference signal Z is generated when the sinusoidal and cosinusoidal analogue signals display a phase angle defined by ZPOS (configured in register CFG2/ZPOS (4:0)) and at the same time the differential voltage of the reference inputs REFP and REFN exceeds the switching point. The default configuration for the phase angle is set to 45° at manufacturing (ZPOS = 4). It has a range from 0° to 348.75°, in 11.25° steps. The switching points of the reference signal must lie in the range between ZPOS ± [90°...150°]. The width of the zero signal Z (reference pulse) at the output can be switched between 1 and 4 increments, i.e. between ¼ and 1 period of the output signals A and B, CFG2/Z4. If the IC is configured to the reference width of 1 increment (¼ period), the outputs A and B carry H level with activated Z signal. The adjustment of the phase angle for matching the IC to the reference signal of the sensor is supported by the IC. Setting ZPOS can be done using test signals or the trigger mode for reference point adjustment (see section 7.6.3 and 7.7).

The relationship between the analogue input signals, the output signals A, B, and Z and the value of the incremental counter is shown in Figure 11.

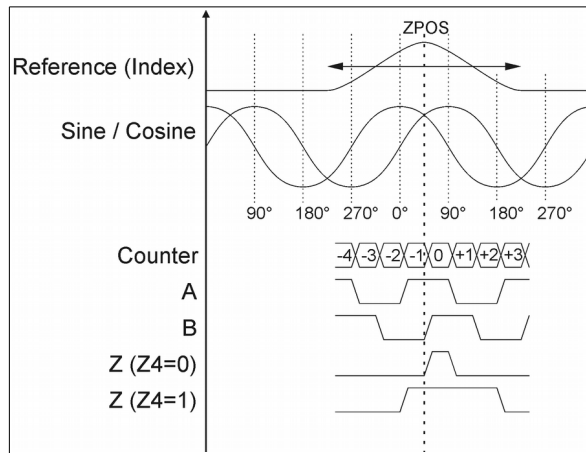


Figure 11: Interpolation (detail)

7.4 Maximum input frequency

The maximum input frequency depends on the selected interface at the output. The maximum input frequency for the absolute position calculation nonius is limited by the f_{OSZ} and corresponding divider. When only using the internal count value through the serial interface, the maximum input frequency is determined by the clock frequency of the circuit (f_{OSZ}). If the square-wave-signals (A/B/Z) are used at the output, the maximum input frequency is limited by the interpolation rate for the A/B-signals and the minimum edge interval (t_{pp}).

The bit CFG1/MNON activates the monitoring for the nonius result error. The reported error at ENON also indicates when the input frequency for the absolute position calculation is exceeded. As for bits MABZ and MFAST, in register CFG1, activates the ABZ error (EABZ) and speed error monitoring (EFAST). The errors show accordingly the limit for the input frequency. If only using the absolute position value (nonius) via the serial interface, MABZ and MFAST can be deactivated. If the serial interface is used and the counters are read, MABZ can be deactivated. For using the square wave signals at the output, requires initializing of both MABZ and MFAST with '1'.

Table 16: Maximum input frequency

Mode	MABZ	MFAST	MNON	Max. frequency for nonius calculation	Max. frequency for the counter	Max. frequency for the ABZ output
Nonius	0	0	1	$f_{MAX} \approx f_{OSZ} / 198$	No error detection	No error detection
Counter	0	1	x		$f_{MAX} \approx f_{OSZ} / 280$	No error detection
Square-wave, $t_{pp} = 2^{TPP} / f_{OSZ}$ $N = 2^{TPP}$ $IRABZ = IRATE / 2^{IRDIV}$	1	1	x		$f_{MAX} \approx 0.9 \cdot f_{OSZ} / (N \cdot IRABZ) < f_{OSZ} / 280$	

TPP is taken from CFG1/CFGTPP
IRDIV is taken from CFG1/IRDIV1

① All values are valid with matched phase between the input signals (SIN and COS) and after the settling of the internal signal control. Up to this time, the input frequency may only amount up to 50% of the specified maximum frequency. Additionally, the maximum input frequency will be limited by the analogue low-pass filter at the input dependent on its configuration.

7.5 Sensor monitoring

The GC-NIP provides 9 sources on each channel for the signal monitoring. The sources can be activated or deactivated using the relevant bit in the register `CFG1`. Storage of the individual error flags can be activated using one further configuration bit each. The OR combination of the error signals saved or masked in this way is provided at the pin `NERR` (L-active). Additional warnings and error information and the individual error conditions can be read via the serial interface (SPI, SSI, BiSS).

The behaviour of the square wave outputs in case of error can be configured. If the bit `HLD` in register `CFG1` is set to the value '1', the outputs will not change in case of error. Setting the bit `TRI` in register `CFG1` to '1' leads to setting the output to a high-impedance state in case of error.

① If the error signal has been activated or one of the error bits has been set in the result register, the current measurement result and all subsequent results must be discarded. After rectification of the error cause, the error bits can be reset by command `RESCNT` or by `PRESET` impulse. For measurements using a reference mark, it is imperative to pass through the reference point to be able to perform further absolute measurements.

Table 17: Overview sensor monitoring

NAME	Description	SPI	ABZ	SSI / BiSS
EVLOW	The signal vector generated from sine- and cosine-signal is too small.	Status bit	Error	Error
ECADC/ ESADC	One or both A/D converters are overdriven.	Status bit	Error	Error
EOFF/ ESOFF	The offset controller has reached its limit.	Status bit	Error	Warning
ECGAIN/ ECSAIN	The gain controller has reached its limit.	Status bit	Error	Warning
EFAST	The input frequency is too high.	Status bit	Error	Error
EABZ	The Signals A, B and Z are invalid.	Status bit	Error	-
ENON ¹⁾	The nonius calculation result is implausible.	Status bit	Error	Error

1) The error can indicate error from input signal or unfavourable correction coefficients.

The error monitoring is configured by the user by switching the relevant bits in register `CFG1` (`MNON`, `MOFF`, `MGAIN`, `MABZ`, `MFAST`, and `MVLOW`). In principle, it is recommended to activate all monitoring sources by setting the bit to '1'. When not using the square wave outputs A, B, and Z, the monitoring of the maximum ABZ-frequency (bit `MABZ`) can be switched off. If only the absolute position is used, the frequency monitoring can also be switched off via `MFAST`. See chapter 7.4 for further information. If the GC-NIP is used as a one- or two-channel interpolation circuit without calculation of the absolute position (nonius), the monitoring of the nonius calculation can be switched off via `MNON`.

Table 18: Recommended configuration of the sensor monitoring

	ABZ-Interface	SPI-Interface	SSI-Interface	BiSS-Interface
Activated monitoring bits	EVLOW EADC EOFFS EGAIN EFAST (ENON) EABZ	EVLOW EADC EOFFS EGAIN (EFAST) (ENON)	EVLOW EADC EOFFS EGAIN (EFAST) (ENON)	EVLOW EADC EOFFS EGAIN (EFAST) (ENON)
Indication in case of error	Error signal on pin <code>NERR</code>	Error bit in register <code>STAT</code> Error bit in <code>POSIT</code> register Error signal at pin <code>NERR</code>	2 bits warning and error in the SSI data stream	2 bits warning and error in the BiSS data stream
Error storage	Deactivate	Activate	Activate	Activate
Clearing of the error storage	-	Command <code>RESCNT</code> <code>PRESET</code> -signal	<code>PRESET</code> -signal	Command <code>RESCNT</code> <code>PRESET</code> -signal
ABZ behaviour in case of error	Hold and/or Tristate	-	-	-
Register <code>CFG1</code> (31:16)	0x00FF (nonius) 0x00BF (no nonius calculation)	0x77F7 (nonius and counter) 0x73F3 (only nonius) 0x37B7 (only counter)	0x77F7 (nonius and counter) 0x73F3 (only nonius) 0x37B7 (only counter)	0x77F7 (nonius and counter) 0x73F3 (only nonius) 0x37B7 (only counter)

The following section describes the monitored sensor signal characteristics and shows the corresponding bits in the registers `CFG1` and `STAT`.

Vector error

The signal vector generated from the sinusoidal and cosinusoidal signals is smaller than 30 percent of the nominal amplitude. Usually, the cause is a partly or completely disconnected sensor. Another cause are input signals with very large offset at simultaneously low amplitude.

Masking	Error storage	STAT-Register	BISS/SSI-SCD
Bit MVLOW	Bit LVLOW	Bit EVLOW	Bit1 – error

ADC error

One or both A/D converters are over-driven. The cause is that the signal amplitude is too high. Another cause are input signals with very large offset at simultaneously high amplitude. If appropriate pull-up or pull-down resistors are connected to the signal inputs, partly or fully disconnected sensors can also be detected by way of this error bit.

Masking	Error storage	STAT-Register	BISS/SSI-SCD
Bit MADC	Bit LADC	Bit ESADC (sine) Bit ECADC (cosine)	Bit1 – error

Offset error

The offset controller has reached its limit. The cause is an excessive signal offset, a partly or fully disconnected sensor or an invalid value for initialization of the offset controller.

Masking	Error storage	STAT-Register	BISS/SSI-SCD
Bit MOFF	Bit LOFF	Bit ESOFF (sine) Bit ECOFF (cosine)	Bit0 – warning

Amplification error

The gain controller has reached its limit. The cause is either that the signal amplitude is too low or the sensor is partly or fully disconnected.

Masking	Error storage	STAT-Register	BISS/SSI-SCD
Bit MGAIN	Bit LGAIN	Bit ESGAIN (sine) Bit ECGAIN (cosine)	Bit0 – warning

Speed error

The input frequency is so high that no A/B signals can be generated or the direction can no longer be detected. The monitored frequency is different depending on whether an internal counter or the square wave outputs A/B/Z are used. See section 7.4. For sole use of the GC-NIP for absolute position calculation (nonius) the detection of this error can be deactivated (MFAST = 0).

Masking	Error storage	STAT-Register	BISS/SSI-SCD
Bit MFAST	Bit LFAST	Bit EFAST	Bit1 – error

A/B/Z error

The signals A, B, and Z are invalid. The cause is an excessive input frequency. The monitored frequency depends on the set minimum edge interval t_{pp} . This error bit will also be set, if the interpolation rate or the minimum edge interval t_{pp} is changed. The detection of this error must be deactivated, if the square wave outputs of the GC-NIP are not in use (MABZ = 0).

Masking	Error storage	STAT-Register	BISS/SSI-SCD
Bit MABZ	Bit LABZ	Bit EABZ	-

Nonius error

The calculated absolute position is invalid. Cause are either errors of the input signals, which can not be internally corrected, or unfavourable combinations of the correction coefficients stored in the EEPROM. The nonius sensor shall be calibrated. See section 7.6.4.

Masking	Error storage	STAT-Register	BISS/SSI-SCD
Bit MNON	Bit LNON	Bit ENON	Bit1 – error

7.6 Pins A/B/Z

The meaning of the signals at the pins A, B, and Z can be modified using the bits `MODE(2:0)` in register `CFG1`. By default, the standard square wave sequences offset by 90° are generated. If only the internal counter of the IC is used, the mode “Controller/DSP” can be activated. Thus, it is possible to carry out equidistant measurements, to synchronize additional components with the IC and to transfer measured values to a processing IC controlled by way of interrupts.

Additional modes are available providing test signals at the pins A, B, and Z for sensor adjustment.

Setting the bit `CFG1/MODE(3)` to '1' enables a SPI-Master interface at the `ABZ`-pins of channel 2. If enabled, this SPI-Master cyclically sends the actual position value (register `POSIT`). Simultaneously, the data of the receiving register of the Master-SPI is used as value for the BiSS/SSI-stream data. This way, the position value can be read and modified from outside, i.e. to provide additional information from a battery powered multiturn-counter, or to transmit additional error information to a control. Also see section 8.4.

Table 19: ABZ modes

Mode	CFG1/MODE	A1	B1	Z1	A2	B2	Z2	TRG	BiSS/SSI
ABZ and nonius	0 000	A1	B1	Z1	A2	B2	Z2	TRG	direct
DSP and nonius	0 001	nINT	StartSample	ZCNT1	nINT	StartSample	ZCNT2	TRG	direct
Sensor adjust 1 ¹⁾ Calibration	0 101	IR4C_1	IR4S_1	RCOMP_1	IR4C_2	IR4S_2	RCOMP_2	TRG Calibration trigger	direct
Sensor adjust 2 ¹⁾	0 110	IR8C_1	IR16C_1	NDEV1	IR8C_2	IR16C_2	NDEV2	TRG	direct
Sensor adjust Z	0 111	REF_SYNC1	ZCNT1	Z1	REF_SYNC2	ZCNT2	Z2	TRG	direct
ABZ and nonius SPI master	1 000	A1	B1	Z1	StartSample	MCMOSI	MCCK	TXENA	Indirect via external controller
DSP and nonius SPI master	1 001	nINT	StartSample	ZCNT1	StartSample	MCMOSI	MCCK	TXENA	Indirect via external controller

¹⁾ For adjustment of the sensor signals the value `ZPOS` in register `CFG2` has to be configured to 00000.

7.6.1 Standard ABZ (mode X000 and X010)

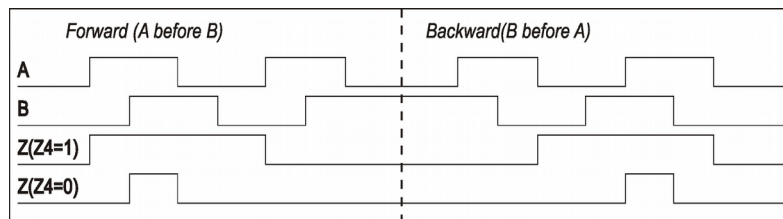


Figure 12: ABZ-Signals

7.6.2 Controller / DSP (mode X001 and X011)

If the measured values of the GC-NIP are transferred exclusively via the serial interface (SPI/BiSS/SSI), additional signals can be provided at the pins A, B, and Z. The pin `NERR` maintains its meaning. It is designed as an open-drain pin, so that the error signals of several ICs can be connected to each other. The detection of ABZ-errors must be deactivated in mode Controller/DSP by writing '0' to `CFG1/MABZ`.

Table 20: DSP-Mode

Pin	Signal	Meaning
A	nINT	Interrupt; L-active; an active signal indicates that at least one of the trigger holding registers is occupied. A read access to the register <code>MVAL</code> provides the 'oldest' measured value saved in the registers. The interrupt can be triggered either by the reference signal at the input or by a signal at the pin <code>TRG</code> . See section 7.7.
B	StartSample	Synchronous signal; this signal delivers the sampling time of the integrated ADC. It can be used to synchronize further systems.
Z	ZCNT	Counter zero signal; this signal indicates that the internal counter of the GC-NIP is reset at the reference point (index point).

7.6.3 Reference point adjustment (sensor adjustment Z - mode 0111)

The phase value for detection of the reference mark (index point) can be moved within one sine period to match different sensors (see Figure 11). The adjustment can be configured in two step sizes; see Table 21.

Table 21: Reference point adjustment

Reference point position	Coarse adjustment	Fine adjustment
Configuration	CFG3/NOSEL = 0	CFG3/NOSEL = 1
Set position in register:	CFG2/ZPOS	NONOFFS
Step size	11.25°	360° / Interpolation rate
Comments	Adjustment value is used for both channels	Each channel can be configured separately. Offset value for the nonius calculation is not available.

For adjustment of the reference point position, test signals can be provided at the pins A, B, and Z (see Figure 13). Additionally, the measured value trigger can be used for adjustment by setting the configuration bits CFG2/ZMODE (1:0) to "01". This way, the GC-NIP triggers the counter on detection of the reference point. Reading the register MVAL (with CFGBISS/STSEL = "01" for channel 1 or CFGBISS/STSEL = "10" for channel 2, provides the values TRGVAL1 and TRGVAL2 (see Figure 13, Table 22). In regard to the interpolation rate, the width of the reference signal and the position of the index signal related to the input signals can be calculated.

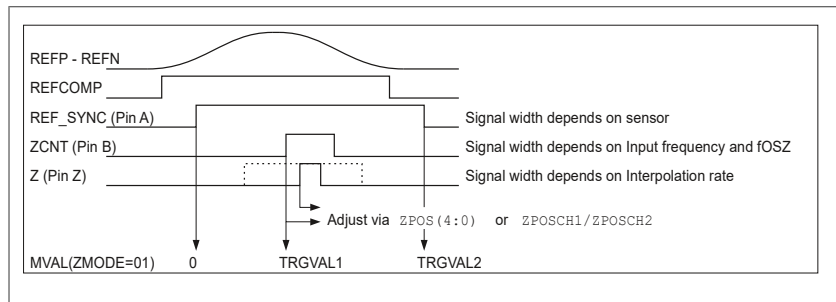


Figure 13: Adjustment of the reference (index) point with trigger mode

Table 22: Example: Reference point adjustment using the trigger mode

Value	Coarse adjustment (CFG3/NOSEL = 0)	Fine adjustment (CFG3/NOSEL = 1)
Width of the reference signal	$Zwidth = TRGVAL2 / IRATE \cdot 360^\circ$	$Zwidth = TRGVAL2 / IRATE \cdot 360^\circ$
Position of the reference signal	$Zstart = Value \cdot ZPOS \cdot 11.25^\circ - (TRGVAL1 / IRATE \cdot 360^\circ)$	$Zstart = (ZPOSCH12 - TRGVAL1) / IRATE \cdot 360^\circ$
Target position	$TRGVAL1 = TRGVAL2 / 2$	$TRGVAL1 = TRGVAL2 / 2$
New calculated position value	$ZPOS_new = (Zstart + Zwidth / 2) / 11.25^\circ$	$ZPOSCH12_new = (Zstart + Zwidth / 2) \cdot IRATE / 360^\circ$
Write to register	CFG2/ZPOS	NONOFFS/ ZPOSCH12

ⓘ The software for evaluation of the triggered values TRGVAL1 and TRGVAL2 should be able to handle the case, that no reference signal is available on the output (Z) or that the trigger values TRGVAL1 or TRGVAL2 are implausible because of multiple index signals. It is recommended to run reference point adjustment at a low signal frequency compared to the ICs clock frequency.

Another option for reference point adjustment is the usage of the preset function. If the preset is enabled via CFG2/PRENA = '1' and the fine adjustment is selected via CFG3/NOSEL = '1' an edge on the pin PRESET stores the actual phase values in register NONOFFS. This way, the reference point position can be selected by an external impulse. For further information to the preset function see section 7.10.

7.6.4 Calibration mode (mode 0101)

The calibration mode of the GC-NIP serves to determine the correction coefficients which are used to enable the 2 track nonius calculation even with distorted sine- and cosine-input-signals. In calibration mode, the registers of the corrected input values (CADC) and a quadrant counter are held on an external event (command TRGCAL or pin TRG) and can be read via the SPI interface. Based on software, these values can be read for the whole period of the nonius scale and the 16 correction coefficients can be calculated and stored in the integrated EEPROM. The evaluation of the coefficients can be realized with the software NIP-Monitor.

7.6.5 Sensor adjustment (mode 0101 and 0110)

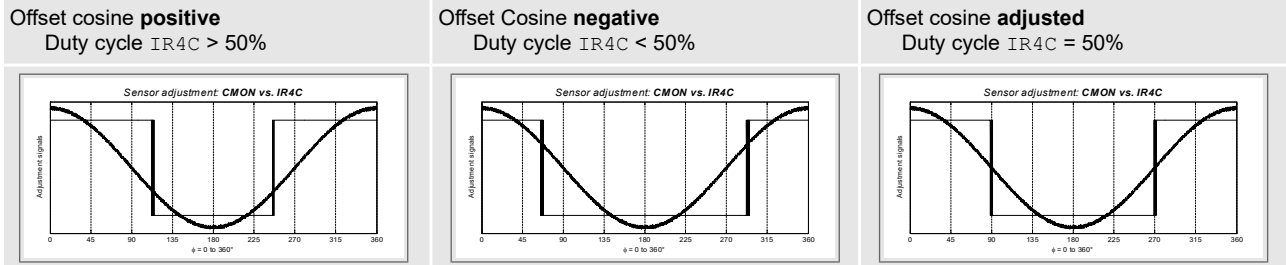
The IC GC-NIP performs an automatic adjustment of amplitude and offset of both signals, sine and cosine, of the encoder. It is reasonable to correct static errors of the sensor previously to use the full control range for dynamic errors. Therefore, supplementary signals at the pins A, B, and Z are available for the modes "Sensor adjustment 1" and "Sensor adjustment 2". The output signals of the instrumentation amplifiers can be measured at the pins SMON and CMON¹⁾. A description of the adjustment sequence can be found in Table 23. Additionally, the following figures show typical signal characteristics.

¹⁾ The analogue low-pass filter must be activated when using the monitor outputs (CFG1/CFGAF ≠ '11').

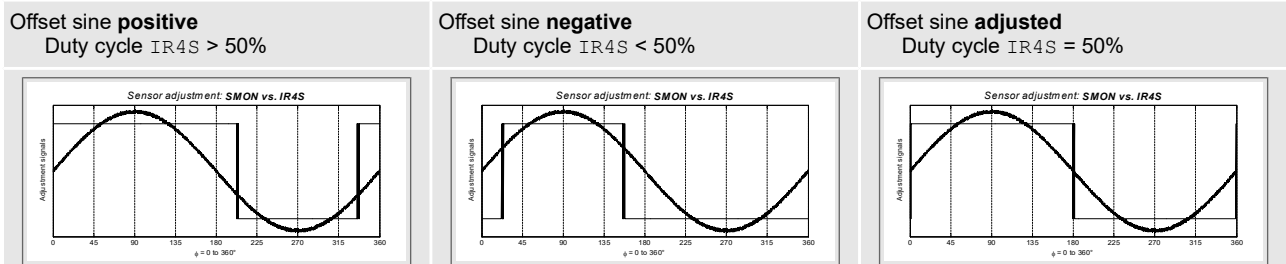
Table 23: Sensor adjustment

No.	Adjustment	Settings of the registers CFG1 / CFG2	Instruction
1	Amplitude Sine/Cosine	Setting of the gain factor	Move sensor; measure on the pins SMON and CMON. Adjustment until both amplitudes display approx. 1.27V _{pp} .
2	Reference	Mode: "Sensor adjustment 1"	Measure signal REFCOMP; adjustment until the signal width corresponds to approx. one period of the sinusoidal signals.
3	Offset Cosine	Mode: "Sensor adjustment 1" Deactivate controller (Bit DISCTRL = '1') Controller disabled; correction values in the middle of the setting range	Move sensor; measure on CMON and signal IR4C. Adjustment until mark-to-space ratio at IR4C is 50% of the period at CMON.
4	Offset Sine	Mode: "Sensor adjustment 1" Deactivate controller (Bit DISCTRL = '1') Controller disabled; correction values in the middle of the setting range	Move sensor; measure on SMON and signal IR4S. Adjustment until mark-to-space ratio at IR4S is 50% of the period at SMON.
5	Phase (coarse)	Mode: "Sensor adjustment 2" Activate controller (Bit DISCTRL = '0').	Move sensor; measure on the pins CMON and signal IR16C, coarse adjustment of the phase until all edges on IR16C are distributed evenly within the sinusoidal period.
5	Phase (fine)	Mode: "Sensor adjustment 2" Activate controller (Bit DISCTRL = 0).	Move sensor; measure at CMON and signal NDEV, adjust phase until frequency at NDEV does not correlate with the frequency of the sinusoidal signal.
6	Amplitude equality	Mode: "Sensor adjustment 2" Deactivate controller (Bit DISCTRL = 1). Controller disabled; correction values in the middle of the setting range	Move sensor, measure at CMON and signal IR8C, adjust signal amplitudes until all edges at IR8C are distributed evenly within the sinusoidal period.

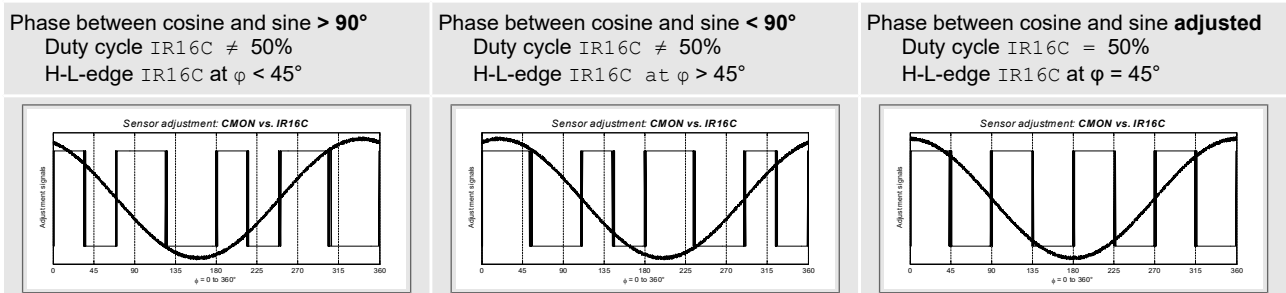
Adjustment offset cosine – signals CMON and IR4C (pin A)
Mode '0101' (sensor adjustment 1), signal controller inactive / correction values in the middle of the setting range



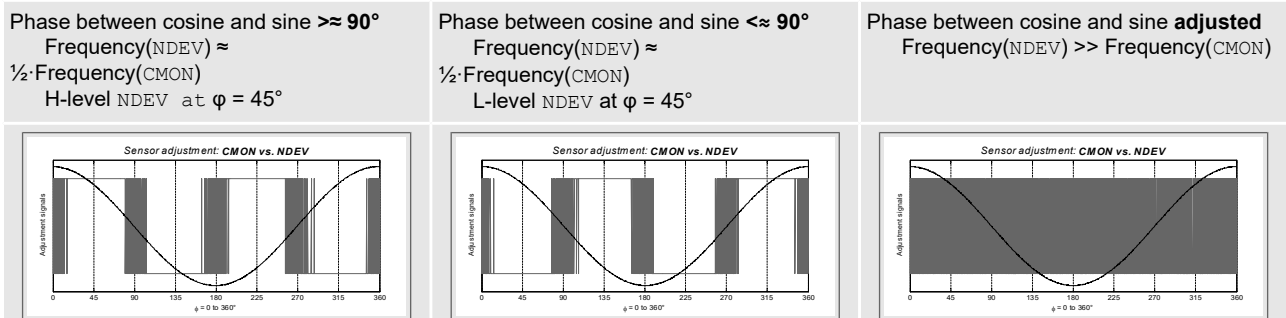
Adjustment offset sine – signals SMON and IR4S (pin B)
Mode '0101' (sensor adjustment 1), signal controller inactive / correction values in the middle of the setting range



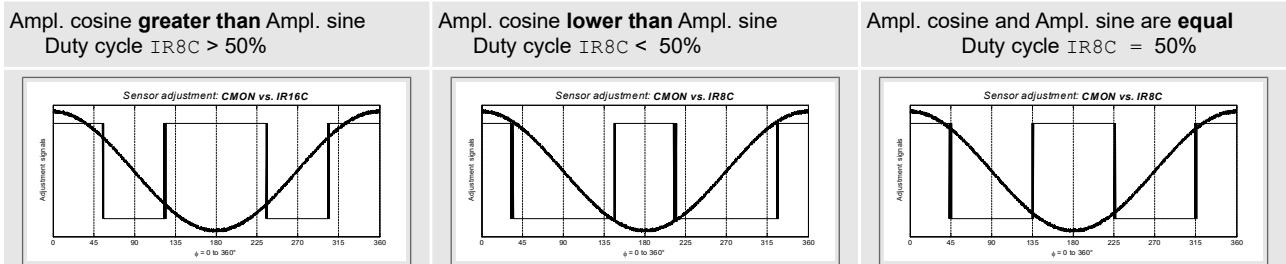
Adjustment phase (coarse) – signals CMON and IR16C (pin B)
Mode '0110' (sensor adjustment 2), signal controller active



Adjustment phase (fine) – signals CMON and NDEV (pin Z)
Mode '0110' (sensor adjustment 2), signal controller active



Adjustment amplitude coincidence – signals CMON and IRC8 (pin A)
Mode '0110' (sensor adjustment 2), signal controller inactive / correction values in the middle of the setting range



7.7 Measured value trigger

The GC-NIP contains two trigger holding registers. The current value of the `POSIT` register can be written to one of these registers controlled by the hardware, for instance by signal edge at pin `TRG`. The respectively 'oldest' value from the trigger holding registers is provided when accessing the register `MVAL` for reading. If no value is saved, the current value of register `POSIT` is displayed. The trigger holding register is re-enabled after reading. The trigger source of the **next** value to be read from `MVAL` is saved in the status register `STAT` (bits `TRG` and `TRGZ`). Furthermore, the bit `TRGOVL` indicates whether a trigger pulse was lost, because both trigger holding registers were occupied when a new trigger pulse appeared. In register `MVAL`, it can be detected from the bit `TRG`, whether the value was triggered by hardware event. If the IC is configured in mode ("Controller/DSP"; see section 7.6.2), the signal `nINT` on pin `A` is switched to L-level, if one of the trigger holding registers is occupied. An example of the program sequence for reading the triggered values using the registers `MVAL` and `STAT` is shown in section 11.3.

Table 24: Trigger mode / reference mode

Trigger source	Usage
TRG-pin	Trigger event from an external device (i.e. measuring probe). Trigger by microcontroller for equidistant measurements.
Reference signal CFG2 / ZMODE='01'	Trigger by the reference (index) signal for evaluation in software.
Reference signal CFG2 / ZMODE='10'	Trigger by the reference (index) signal for adjustment of the reference position in software.
Reference signal CFG2 / ZMODE='11'	Trigger by the reference (index) signal for the evaluation of distance coded reference marks.

TRG-pin

The current count value is written to one of the two trigger holding registers via a signal edge at pin `TRG`. The trigger edge (falling or rising) can be set using the bit `TRGSLP` in register `CFG1`.

Reference index trigger

Any occurrence of a reference index at the input triggers the actual position value (register `POSIT`) to be stored in one of the two trigger holding registers.

Adjustment of the reference index position

The method of adjusting the reference point position is described in section 7.6.3. A rising edge at the reference signal resets the internal counter. The detection of the index in the IC (at the configured phase angle) triggers the counter to be stored into the first trigger holding register. The falling edge at the reference signal at the input triggers the counter to be stored in the second trigger holding register. After the two trigger events, the bit `ZSTAT` in register `STAT` is set and the trigger processing is locked until release by one of the SPI/BiSS commands `RESCNT` or `CLRZ`. See Figure 13.

Processing of distance coded reference marks

The first reference mark resets the internal counter. The second reference mark triggers the counter to be stored in the first trigger holding register. After these two events, the bit `ZSTAT` in register `STAT` is set and the trigger processing is locked until release by one of the SPI/BiSS commands `RESCNT` or `CLRZ`. The distance between the two reference marks must be at least two periods of the input signals. The calculation of the absolute position of the sensor from any further values is handled by the evaluation software. See section 11.5 and Figure 50 for more information.

7.8 Measured value register CNT, POSIT, MVAL, and STAT

The interpolated counter values, the position values, the trigger register holding values, as well as the sensor monitoring information can be read out from various registers via the serial interfaces. Following table shows the measured value registers for the different interfaces. For the BiSS interface it can be chosen between register data (slow communication) and single-cycle-data (SCD; fast communication). Programming examples for reading the measured value registers are shown in section 11.3.

Table 25: Measured value registers

Registers	SPI	SSI	BiSS
Register CNT1 / CNT2	Interpolation counter 30 bit Index status 1 bit Error status 1 bit	-	Interpolation counter 30 bit Index status 1 bit Error status 1 bit
Register POSIT	See Table 27 Error status 2 Bit	-	Use SCD
Register MVAL	Register MVAL: see Table 28 Error status 1 bit Trigger status 1 bit	-	-
Register STAT	Error status 19 bit Trigger status 3 bit Index status 2 bit	-	Error status 19 bit Trigger status 3 bit Index status 2 bit
SCD (BiSS) / SSI-data ¹⁾	-	See Table 26 Error status 2 bit	See Table 26 Error status 2 bit 2 MSB: 00

¹⁾ This data can be modified by an externally connected microcontroller → see section 8.4

Content and format of the position data register (POSIT) can be selected in register CFGBISS using the bits STBIT, STSEL, and GRAY. The configuration bits STBIT (4:0) describe the bit-length (LSB) of the position value, unused MSBs are set to '0'. The coding of the position data can be switched between gray code (GRAY = '1') and binary code (GRAY = '0'). The value STSEL defines the content in the position data register POSIT (see table 26 and table 27).

Table 26: Position data SSI/BiSS/SPI

STSEL	Position data SSI ¹⁾	Position data BiSS	Position data SPI
00	30 Bit Position / 8-30 Bit Resolution 1 Bit Error / 1 Bit Warning	32 Bit Position / 8-30 Bit Resolution 1 Bit Error / 1 Bit Warning	30 Bit Position / 8-30 Bit Resolution 1 Bit Error / 1 Bit Warning
01	30 Bit Counter Channel 1 / 8-30 Bit Resolution 1 Bit Error / 1 Bit Warning	32 Bit Counter Channel 1 / 8-30 Bit Resolution 1 Bit Error / 1 Bit Warning	30 Bit Counter Channel 1 / 8-30 Bit Resolution 1 Bit Error / 1 Bit Warning
10	30 Bit Counter Channel 2 / 8-30 Bit Resolution 1 Bit Error / 1 Bit Warning	32 Bit Counter Channel 2 / 8-30 Bit Resolution 1 Bit Error / 1 Bit Warning	30 Bit Counter Channel 2 / 8-30 Bit Resolution 1 Bit Error / 1 Bit Warning
11	15 Bit Counter Channel 1 15 Bit Counter Channel 2 1 Bit Error / 1 Bit Warning	2 Bits unused 15 Bit Counter Channel 1 15 Bit Counter Channel 2 1 Bit Error / 1 Bit Warning	15 Bit Counter Channel 1 15 Bit Counter Channel 2 1 Bit Error / 1 Bit Warning

1) The length of the SSI-data depends on the configuration of bit CFGBISS/SSI20 (32 or 20 bit)

Table 27: Configuration of the position data register POSIT(31:0)

STSEL	Register POSIT	ERROR (1Bit)	WARNING(1Bit)
00	Nonius position (max. 30Bit)	ERROR (1Bit)	WARNING(1Bit)
01	Interpolation counter 1 (max. 30Bit)	ERROR (1Bit)	WARNING(1Bit)
10	Interpolation counter 2 (max. 30Bit)	ERROR (1Bit)	WARNING(1Bit)
11	Interpolation counter 2 (15Bit) ¹⁾ Interpolation counter 1 (15Bit) ¹⁾	ERROR (1Bit)	WARNING(1Bit)

1) It is recommended to use binary representation only (GRAY = '0') for STSEL = '11'

Table 28: Configuration of the position data register MVAL(31:0)

STSEL	Register MVAL	Trigger status (1 Bit)	ERROR(1Bit)
00	Nonius position (max. 30Bit)	Trigger status (1 Bit)	ERROR(1Bit)
01	Interpolation counter 1 (max. 30Bit)	Trigger status (1 Bit)	ERROR(1Bit)
10	Interpolation counter 2 (max. 30Bit)	Trigger status (1 Bit)	ERROR(1Bit)
11	Interpolation counter 2 (15Bit) Interpolation counter 1 (15Bit)	Trigger status (1 Bit)	ERROR(1Bit)

The register MVAL contains a triggered position value if bit 1 (Trigger status) is set. Otherwise, the actual value from the POSIT register is displayed. → see section 7.7

7.9 Count direction switch (pin DIR)

The configuration pin `CFGDIR` serves to set the count direction for the absolute (nonius) position calculation. The pin has no effect on the count direction of the two interpolation counters or the A/B output signals.

7.10 Counter preset / Nonius offset / Commands / Control signals

The reset values for the integrated interpolation counters, the nonius offset value and the reference point position can be configured. The GC-NIP contains the preset registers `PREST1` and `PREST2`, and the nonius offset register `NONOFFS/ZPOSCH12`, which allow the user to set the zero position of the sensor independently from a reference signal. In conjunction with the integrated EEPROM, the zero position can be stored during system power-down. An overview about available signals and commands is shown in table 29.

Table 29: Commands

Command / Signal	Action
Reference signal (at pin <code>REFP/REFN</code>)	Resets the counter to zero
Reset SPI/BiSS command <code>RESIC</code> ¹⁾	The registers <code>PREST1</code> and <code>PREST2</code> , the controller parameters and the register <code>NONOFFS</code> are loaded from the EEPROM. The content of the preset registers (<code>PREST1/PREST2</code>) is transferred into the counter registers.
Reset SPI/BiSS command <code>RESIC</code> ²⁾	The registers <code>PREST1</code> and <code>PREST2</code> and the register <code>NONOFFS</code> are not changed. The content of the preset registers (<code>PREST1/PREST2</code>) is transferred into the counter registers.
SPI/BiSS command <code>RESCNT</code>	The content of the preset registers <code>PREST1</code> and <code>PREST2</code> is transferred into the counter registers.
SPI/BiSS command <code>RESCTL</code>	The controller parameters are set to the center of their value range.
SPI/BiSS command <code>WCFCG</code>	The registers <code>PREST1</code> , <code>PREST2</code> and the controller parameters are written to the EEPROM.
SPI/BiSS command <code>PRESET</code>	If the bit <code>CFG2/PREENA</code> is set, the offset for the nonius position <code>NONOFFS</code> is re-calculated using the preset value for the nonius from register <code>PREST2</code> and the new offset value (<code>NONOFFS</code>) is stored in the EEPROM.
Falling edge at pin <code>PRESET</code>	<code>CFG3/NOSEL = '0'</code> The content of the preset registers <code>PREST1</code> and <code>PREST2</code> is transferred into the counter registers. If the bit <code>CFG2/PREENA</code> is set, the offset for the nonius position <code>NONOFFS</code> is re-calculated using the preset value for the nonius from register <code>PREST2</code> and the new offset value (<code>NONOFFS</code>) is stored in the EEPROM. <code>CFG3/NOSEL = '1'</code> If the bit <code>CFG2/PREENA</code> is set, the actual phase angle of both channels is written to the register <code>NONOFFS</code> and stored in EEPROM. The contents of register <code>NONOFFS/ZPOSCH12</code> define the reference index position (fine adjustment of signal Z) (see register description and section 7.6.3)

¹⁾ if EEPROM is valid, see section 6.1 ²⁾ if EEPROM is invalid, see section 6.1

The signal on pin `PRESET` is debounced in the IC. After a falling edge of the signal, the signal event generation is locked for a time of t_{debounce} , which is about 60 ms for a clock frequency of 26 MHz. The function of the pin `PRESET` depends on the configuration bit `CFG2/PREENA`. Additionally, the configuration of `CFG3/NOSEL` switches the meaning of the register `NONOFFS/ZPOSCH12`.

7.10.1 Nonius-Offset

Configuring the bit `CFG3/NOSEL` with '0' defines the register `NONOFFS` to be used as offset register for the nonius calculation. The content of the `POSIT` register is then calculated from the physical nonius position and the nonius offset (see table 31). Using the preset function, the offset can be calculated in hardware, where the register `PREST2` holds the target position for the absolute position (see tables 30 and 31).

Table 30: PRESET-Pin

<code>CFG2/PREENA = '0'</code>	<code>CFG2/PREENA = '1'</code>
The interpolation counter register <code>CNT1</code> is loaded with the content of the <code>PREST1</code> register.	The interpolation counter register <code>CNT1</code> is loaded with the content of the <code>PREST1</code> register.
The interpolation counter register <code>CNT2</code> is loaded with the content of the <code>PREST2</code> register.	The interpolation counter register <code>CNT2</code> is loaded with the content of the <code>PREST2</code> register.
	The offset for the nonius position is re-calculated using the actual position and the content of the <code>PREST2</code> register.
	The calculated value for the nonius offset (<code>NONOFFS</code>) is stored in the EEPROM.

The following relationship applies for the registers `PREST2`, `POSIT` (Nonius = absolute position value) and `NONOFFS`:

Table 31: Nonius offset

Measurement	Preset-Function, <code>CFG2/PREENA = '1'</code>
$POSIT = \text{Nonius}(\text{PHI1}, \text{PHI2}, \text{CFGDIR}) + \text{NONOFFS}$	$\text{NONOFFS (new)} = \text{PREST2} - \text{Nonius}(\text{PHI1}, \text{PHI2}, \text{CFGDIR})$ $POSIT = \text{Nonius}(\text{PHI1}, \text{PHI2}, \text{CFGDIR}) + \text{NONOFFS (new)} = \text{PREST2}$

7.10.2 Configuration of the reference point position

Setting the configuration bit `CFG3/NOSEL` to '1' defines the register `NONOFFS/ZPOSCH12` to store the reference mark position for both channels. A signal edge at pin `PRESET` then stores the actual phase angle information (`PHI1` and `PHI2`) to be used as reference point position. This way, the zero position of the internal counters and A/B signals can be set freely within one sine period (see Figure 14 and Table 32). The resolution of the phase angle information depends on the selected interpolation rate.

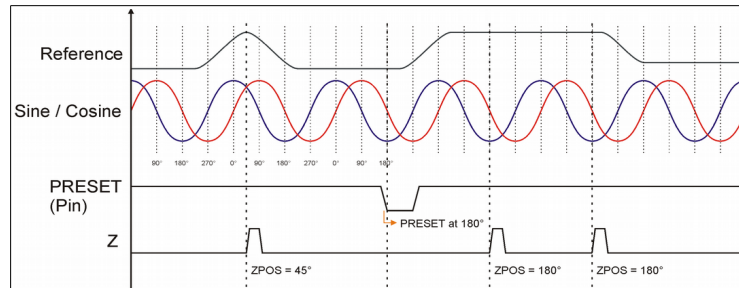


Figure 14: Fine adjustment of the reference point position

Table 32: `PRESET-Pin` `CFG3/NOSEL = 1`

<code>CFG2/PRENA = 0</code>	<code>CFG2/PRENA = 1</code>
The interpolation counter register <code>CNT1</code> is loaded with the content of the <code>PREST1</code> register.	The interpolation counter register <code>CNT1</code> is loaded with the content of the <code>PREST1</code> register.
The interpolation counter register <code>CNT2</code> is loaded with the content of the <code>PREST2</code> register.	The interpolation counter register <code>CNT2</code> is loaded with the content of the <code>PREST2</code> register.
The reference point position is loaded from register <code>NONOFFS</code> .	The actual phase angle of channel 1 is stored in register <code>NONOFFS/ZPOSCH1(15:0)</code> .
	The actual phase angle of channel 2 is stored in register <code>NONOFFS/ZPOSCH1(31:16)</code> .
	The register <code>NONOFFS</code> is stored in EEPROM.
	The reference point position is loaded from register <code>NONOFFS</code> .

Note: Since counter, controller and EEPROM are influenced by several sources, the following notes apply:

- During EEPROM access, the `PRESET`-signal is suppressed
- During EEPROM access, the command `RESCNT` is suppressed.
- During EEPROM access, the command `WCFG` is suppressed.
- If the `PRESET` signal is active during write access to the registers `PREST1`- or `PREST2`, faulty values may be written into the counter registers.
- If the `PRESET` signal is active or was active up to 40 ms before, the register `NONOFFS` shall not be written via the serial interface. Otherwise, faulty values may be written to the EEPROM.
- Please pay attention to the maximum number of write cycles for the EEPROM when using the commands `WCFG` and the `PRESET` signal.

7.11 Power saving options

To lower the current consumption of the GC-NIP, several functions of the IC can be disabled:

Table 33: Power saving options

Configuration bit	Effect	Typical Application
<code>CFG2/DISMON = '1'</code>	The pins <code>SMON1</code> , <code>SMON2</code> , <code>CMON1</code> and <code>CMON2</code> are inactive.	Adjustment of the analogue signals has been finished. These pins are not required for operation of the GC-NIP.
<code>CFG2/DISV0 = '1'</code>	The pins <code>VOCH1</code> and <code>VOCH2</code> are inactive.	The mean voltages of the GC-NIP are not used for the sensor (i.e. for measuring bridges).
<code>CFG3/DISCH2 = '01'</code>	<code>ABZ</code> output ¹⁾ and counter value of channel 2 is not calculated. The phase angle value of channel 2 is calculated.	GC-NIP is used for nonius calculation.
<code>CFG3/DISCH2 = '11'</code>	Nonius position, phase angle, <code>ABZ</code> output ¹⁾ and counter value of channel 2 are not calculated.	GC-NIP is used as one channel interpolator.
<code>CFG3/DISZ1 = '1'</code>	The reference point processing on channel 1 is deactivated.	GC-NIP is used for nonius calculation or as interpolator for measuring scales without reference mark.
<code>CFG3/DISZ2 = '1'</code>	The reference point processing on channel 2 is deactivated.	GC-NIP is used for nonius calculation. GC-NIP is used as interpolator for measuring scales without reference mark. GC-NIP is used as one channel interpolator.

¹⁾ The `ABZ` output is de-/activated after reset of the IC.

7.12 Signal propagation time

The propagation delay of the input signals through the instrumentation amplifier of the GC-NIP is given by the chosen gain factor and the setting of the cut-off frequency of the low-pass filter. The following table shows approximate values for some configurations.

Table 34: Propagation delay analogue (td_{ANA})

Configuration	Min	Typ.	Max
CFG1/CFGAF = '00' (150 kHz)	720 ns	800 ns	880 ns
CFG1/CFGAF = '01' (75 kHz)	1.0 μ s	1.2 μ s	1.4 μ s
CFG1/CFGAF = '10' (10 kHz)	2.1 μ s	2.4 μ s	2.7 μ s
CFG1/CFGAF = '11' (inactive) CFG1/CFGGAIN = '00'	70 ns	100 ns	120 ns
CFG2/LP CFGAF = '11' (inactive) CFG1/CFGGAIN = '11'	70 ns	130 ns	180 ns

The propagation delay td_{DIG} between sampling and measurement value in the registers MVAL, POSIT or CNT1, CNT2 as well as at the pins ABZ depends on the selected operating mode:

Table 35: Propagation delay digital (td_{DIG})

Mode	Configuration	Register CNT1/2	Register POSIT	ABZ
Nonius	CFG1/Mode = 'X000'	112 clock cycles f_{OSZ}	181 clock cycles f_{OSZ}	208 clock cycles f_{OSZ}
	CFGBISS/STSEL = '00'	4.3 μ s @ f_{OSZ} = 26MHz	7 μ s @ f_{OSZ} = 26MHz	8 μ s @ f_{OSZ} = 26MHz
Two channel	CFG1/Mode = 'X000'	112 clock cycles f_{OSZ}	115 clock cycles f_{OSZ}	208 clock cycles f_{OSZ}
	CFGBISS/STSEL \neq '00'	4.3 μ s @ f_{OSZ} = 26MHz	4.4 μ s @ f_{OSZ} = 26MHz	8 μ s @ f_{OSZ} = 26MHz

ⓘ Please note that the **constant** propagation delay of the IC (as in every digital system) causes a frequency dependent phase shift between the analogue input signals and the output signals of $\varphi = 2\pi \cdot f \cdot td$. The following figures show this behaviour for the output signal Z for two different input signal frequencies as an

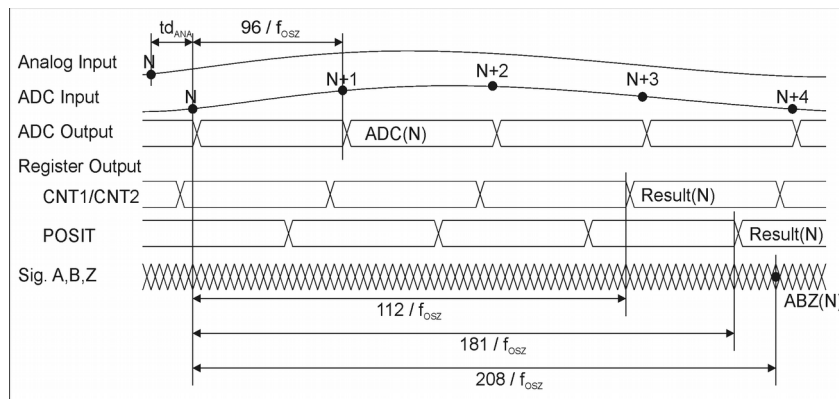


Figure 15: Signal propagation time

example. The behaviour of the signals A and B is equivalent.

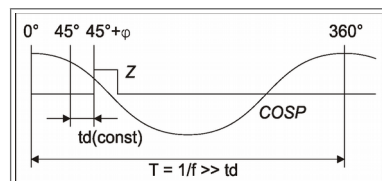


Figure 16: Constant propagation delay

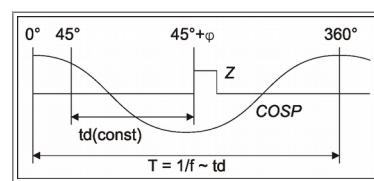


Figure 17: Constant propagation delay

8 Digital interfaces

8.1 Serial interface SPI

The serial interface SPI of the GC-NIP is activated if the pin *SEN* is held on H-level during reset of the IC. The GC-NIP operates in slave mode. In other words: It cannot start communication itself. Up to sixteen ICs can be operated on a single interface bus. The interface is compatible to the most important microcontroller families in SPI mode 0 (16 bit data, MSB first, SCK default low, sampling with rising clock signal edge).

8.1.1 Signals

Signal	Meaning	Direction
SEN	Enable Low: Interface is enabled High: Interface is disabled, <i>MISO</i> becomes high-resistant or is set to <i>nWAIT</i> Rising edge: Command is executed	IN
SCK	Clock cycle The data at <i>MOSI</i> is sampled by the IC with the rising edge at <i>SCK</i> . The data at <i>MISO</i> is modified by the IC with the falling edge at <i>SCK</i> .	IN
MOSI	Master-OUT / Slave-IN Data input	IN
MISO/nWAIT	Master-IN / Slave-OUT Data output and status signal Please note: A Pull-Up resistor is required at this pin!	OUT (tristate-capable)

While the IC is reset or during the waiting time of a synchronous SPI read command, the *MISO* line is kept at L level (meaning of *nWAIT*).

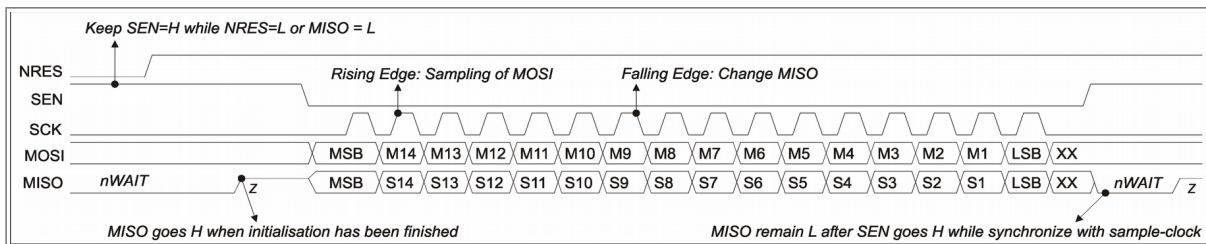


Figure 18: SPI-transfer (1)

8.1.2 Protocol

OP-Code	Description	Bit at signal MOSI															
		15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
WRA	Write address	1	0	0	nB	H3	H2	H1	H0	A7	A6	A5	A4	A3	A2	A1	A0
WRD	Write data	1	0	1	nB	H3	H2	H1	H0	D7	D6	D5	D4	D3	D2	D1	D0
RD0/ST	Read Bytes 0+1 (2 LSB)	1	1	0	X	H3	H2	H1	H0	A7	A6	A5	A4	A3	A2	A1 ^{*)}	0
RD1	Read Bytes 2+3 (2 MSB)	1	1	1	X	X	X	X	X	X	X	X	X	X	X	X	X
NOP	Output read register	0	X	X	X	X	X	X	X	X	X	X	X	X	X	X	X

H (3:0) : Hardware address, default: '0000', Is not evaluated if nB = 0
 A (7:0) : Register address within an IC
 D (7:0) : Data word / write data (read data will appear at the pin *MISO*)
 nB: Broadcast (L-active) 0: Command to all ICs
 1: Command to the IC addressed by way of H (3:0)

Default-OP-Codes

WRA = 0x80 + address WRD = 0xA0 + data
 RD0 = 0xC0 + address RD1 = 0xE000
 NOP = 0x0000

*) Some registers can be addressed for reading 16 bit values. Usually the command *RD0* has to be sent with *A1* set to 0 for reading of the registers.

Any data transfer is initiated by the host processor sending of an SPI word. An SPI word consists of: 4 bits OP code, 4 bits hardware address, and up to 8 bits data. OP codes are only accepted if the hardware address sent matches with the hardware address of the GC-NIP. The hardware address of the IC after a reset is '0000'. The command `SETHWA` (see section 9, register `CMD`) can be used to read the pins `HWA<3:0>` into the IC as the new hardware address. OP codes for reading of a register result in data output at the pin `MISO` in the subsequent SPI access (regardless of the hardware address in the new SPI word).

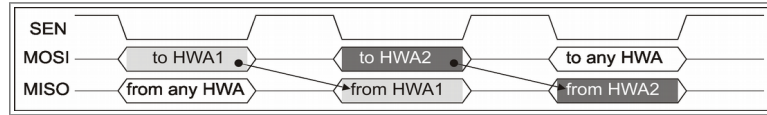


Figure 19: SPI transfer (2)

8.1.3 Register access

Register access in the GC-NIP is done by writing 8 bit and reading 16 bit. The registers of the IC are organized in 32 bit blocks. Therefore, the IC contains a 32 bit holding register for read access. Data to be read is stored in the holding register using the SPI opcode `RD0/ST`. The two least significant bytes of the data to be read is output at the pin `MISO` during the next SPI access (see Figure 21). The two most significant bytes of the read access are output with the SPI cycle following the command `RD1`. To read a 32 bit register, the commands `RD0/ST`, `RD1` and `NOP` are usually executed one after another. To read several registers in succession, the sequence `RD0 – RD1 – RD0 – RD1...` can be used (see Figure 23).

To write a GC-NIP register, first the register address must be set using the SPI opcode `WRA`. Subsequently, the register can be programmed using `WRD`. The register is programmed byte by byte (see Figure 22).



Figure 20: Write access 8 Bit



Figure 21: Read access 32 Bit



Figure 22: Write access 32 Bit

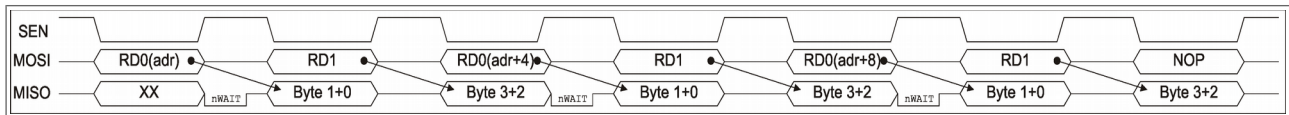


Figure 23: Read access 3 x 32 Bit

8.1.4 Synchronous / asynchronous access

Reading a register, the data is taken over into the holding register synchronously to the internal IC sequence. The value `SYNC` in the register `CFG2` can be used to shift the time relative to the sampling time of the A/D converter. Thus, it is possible to carry out equidistant measurements with small delays.

The pin `MISO` is low during the waiting time (`nWAIT`). If the bit `ASYN` in the register `CFG2` is set, the data is stored immediately after the rising edge at the signal `SEN`. The time reference to the sampling of the analogue signals will be lost. Thus, higher baud rates are achieved.

① To read the registers `IP11`, `IP12`, `IP21`, `IP22`, `Nonius` or for read access at SPI page 1, a value `SYNC(6:0) = 64 (decimal)` must be used.

8.2 BiSS interface

The BiSS C-mode interface of the GC-NIP is activated if the pin `SEN` is held on L-level during reset of the IC and the bit `SSI` in register `CFGBISS` is set to '0'. Please note that the level at the pins `HWA (3:0)` is read in for use as the 4 LSB of the BiSS serial number. Thus multiple ICs can be used on a single interface bus.

For use of the BiSS interface, the integrated EEPROM must contain a valid configuration, because essential operating parameters are stored in EEPROM. The configuration bits `BISSTO` and `READ32` in register `CFGBISS` can be used to adapt the interface to the system parameters.

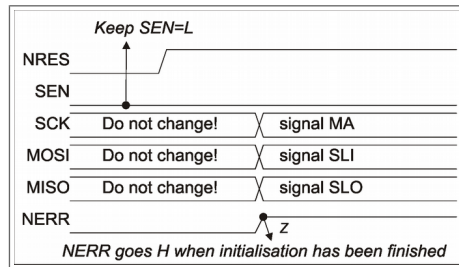


Figure 24: BiSS interface initialization

The Single Cycle Data (SCD) transferred in BiSS C-mode contains the actual position value from register `POSIT` (see section 7.8) with an overall length of 40 bit. This includes the 32 bit position (see table 26), two bits of error information (error and warning bit) and the CRC checksum (polynomial 0x43, 6 bit, inverted).

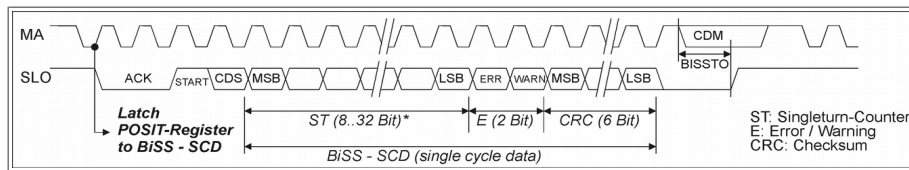


Figure 25: BiSS SCD (Single Cycle Data)

Using the BiSS register access, all registers of the GC-NIP are attainable. Reading of 32 bit registers requires the bit `READ32` in register `CFGBISS` to be set. Read access is then handled in 32 bit format, so 4 subsequent addresses, beginning with the least significant address (divisible by 4), must be read by the master. Additionally, the hints for configuring the bits `SYNC (6:0)` of register `CFG2` must be taken into account (see section 8.1.4).

Table 36: Register `CFGBISS` (BiSS mode)

Bit	Meaning	Vendor configuration	User configuration
BISSTO	BiSS-Timeout	19.7µs at 26 MHz	BISSTO = $\log_2(\text{Timeout} \cdot f_{\text{osz}})$
READ32	Data format read access	Reading of configuration registers	Reading of data- or configuration registers

Table 37: Default values BiSS register

Register	Vendor configuration	User configuration
BiSS serial number	MSB: 0 LSB: level at pins <code>HWA (3:0)</code>	MSB: unique serial number LSB: level at pins <code>HWA (3:0)</code>
BiSS Vendor ID	0x47 0x43 („GC“)	User defined ID
BiSS Device ID	0x32 0x03 0x00 0x00	User defined ID
BiSS-Profile + Electronic data sheet (EDS)	unused	User profile

Further specification of the BiSS interface, signal waveforms, register description as well as information to the electronic data sheet (EDS) can be found on the website www.biss-interface.com.

8.3 SSI interface

The SSI interface of the GC-NIP is activated if the pin SEN is held on L-level during reset of the IC and the bit SSI in register CFGBISS is set to '1'. For use of the SSI interface, the integrated EEPROM must contain a valid configuration, because essential operating parameters are stored in EEPROM. The configuration bits SSITO and RING in register CFGBISS can be used to adapt the interface to the system parameters.

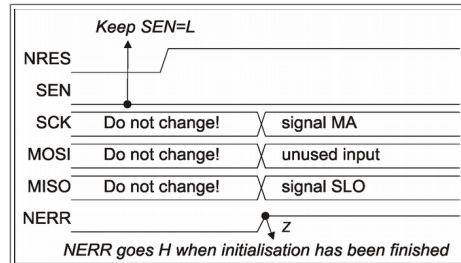


Figure 26: SSI initialization

The SSI data output contains the position value (register POSIT, see section 7.8) with an overall length of 20 or 32 bit. This contains the measured position and two bits of error information (error and warning). Setting the bit RING in register CFGBISS enables continuous transmission of the measurement value in SSI ring mode.

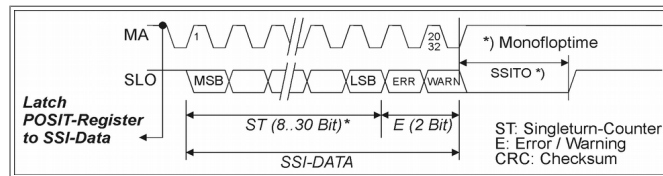


Figure 27: SSI

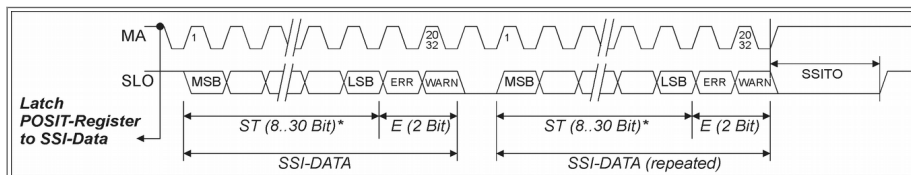


Figure 28: SSI (ring mode)

Table 38: Register CFGBISS (SSI-Mode)

Bit	Meaning	Vendor configuration	User configuration
SSITO	SSI-Timeout		SSITO = (Timeout·f _{OSZ})-3
RING	SSI ring mode	Ring mode	Adapt the operating mode to the master.
SSI20	Output data length	20 bit	0 for 32 bit / 1 for 20 bit

8.4 Simple SPI Master

Setting bit `CFG1/MODE(3)`, enables the SPI-master at the pins A, B, Z of channel 2 which sends the position data (register `POSIT`) cyclically to a connected slave. Additionally, the received data of this SPI form the bits (31:0) of the SSI- or BiSS-data. This way, additional information, for example from an external multturn counter, can be added to the measurement value or extra error information can be transferred to a control. The microcontroller connected to the SPI must be able to send and receive 32 bits of data with an SPI clock of $f_{osz}/2$. It operates in SPI mode 0 (MSB first, sampling on the rising clock edge, clock default low).

The following figures show the flow of data and the operational sequence in GC-NIP and in the microcontroller.

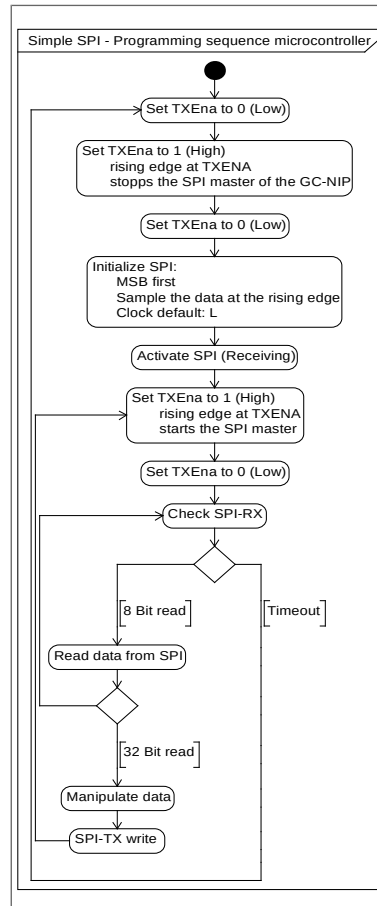


Figure 29: Program sequence microcontroller

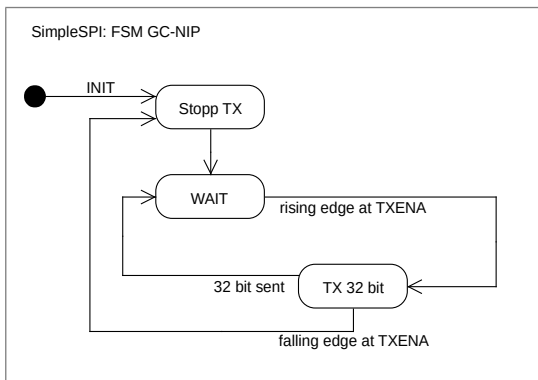


Figure 30: Program sequence GC-NIP

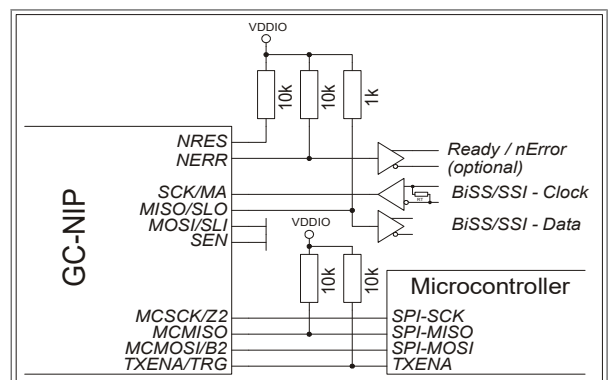


Figure 31: Simple SPI Master

8.5 EEPROM

The GC-NIP contains an integrated EEPROM for permanent storage of the user configuration. The IC checks during reset, if the EEPROM content is valid and sets the configuration. **For validation of the different areas, the EEPROM contains the identifier 0x134A at the EEPROM addresses 0x00-0x02.** Reading from and writing to the EEPROM is handled by an internal interface, which can be accessed using the register `EEP`.

Using the BiSS interface, read access to the EEPROM can be done directly using the BiSS address (40). The address allocation in the EEPROM differs from the addressing via the SPI or BiSS interface.

Table 39: EEPROM addressing

	Register	EEPROM
Word size data	8 Bit	16 Bit
Word size address	8 Bit / 2 Pages	8 Bit / EEPROM address = Register-address / 2
Endianness (user register)	Little Endian	Little Endian or special format
Endianness (BiSS register)	Big Endian	Big Endian

Table 40: Address mapping

Area	Usage	Address SPI	Address BiSS	Address EEPROM	Format
User register	Config	0x00...0x3F (Page 0)	0x00...0x3F (Page 0)	0x00 ... 0x1F	Little Endian
BiSS register	BiSS	-	0x40...0x47	0x20 ... 0x23	Big Endian
User register	Special	0x48...0x77 (Page 0)	0x48...0x77	0x24 ... 0x3B	Little Endian
BiSS register	BiSS	-	0x78...0x7F	0x3C ... 0x3F	Big Endian
BiSS EDS-Common	BiSS	-	0x00...0x3F (Page 1)	0x40 ... 0x5F	Big Endian
BiSS EDS-Profile 1	BiSS	-	0x00...0x3F (Page 2)	0x60 ... 0x7F	Big Endian
BiSS EDS-Profile 2	BiSS	-	0x00...0x3F (Page 3)	0xC0 ... 0xDF	Big Endian
User register	Coefficients	0x40...0x7F (Page 1)	0x00...0x3F (Page 4)	0xA0 ... 0xBF	Little Endian
User register	IP-Table	0x80...0xBF (Page 1)	0x00...0x3F (Page 5)	0x80 ... 0x9F	Special format 20 bit
User register	IP-Table (fix)		0x00...0x3F (Page 6)	0xE0 ... 0xFF	Special format 20 bit
User register	Read-Register	0x80...0xBF (Page 0)	0x00...0x3F (Page 7)	-	Little Endian
Page register SPI	Page	0xFF (any Page)	-	-	Byte

The sequence for reading and writing the EEPROM are described in section 11.4. It is important to ensure not to write the `EEP` register if the bit `EEPBSY` in register `EEP` is set.

9 Register

Table 41: Register overview

Register	Access ¹⁾	Address SPI	BiSS-Page	Address BiSS	Address EEPROM ²⁾	Hints
IDREV + Status	R	0x00	0	0x00...0x03	0x00...0x01	0x00: valid flag for configuration 0x01: valid flag for the coefficients
CFGEEP	!	0x04...0x07		0x04...0x07	0x02...0x03	
CFG1	R/W	0x08...0x0B		0x08...0x0B	0x04...0x05	
CFG2	R/W	0x0C...0x0F		0x0C...0x0F	0x06...0x07	
CFG3	R/W	0x10...0x13		0x10...0x13	0x08...0x09	
Unused	R/W	0x14...0x17		0x14...0x17	0x0A...0x0B	
Unused	R/W	0x18...0x1B		0x18...0x1B	0x0C...0x0D	
CTRLG1	R/W	0x1C...0x1F		0x1C...0x1F	0x0E...0x0F	
CTRLO1	R/W	0x20...0x23		0x20...0x23	0x10...0x11	
PREST1	R/W	0x24...0x27		0x24...0x27	0x12...0x13	
CTRLG2	R/W	0x28...0x2B		0x28...0x2B	0x14...0x15	
CTRLO2	R/W	0x2C...0x2F		0x2C...0x2F	0x16...0x17	
PREST2	R/W	0x30...0x33		0x30...0x33	0x18...0x19	
NONOFFS	R/W	0x34...0x37		0x34...0x37	0x1A...0x1B	Change EEP at Preset signal / no update at cmd WCFG
CFGBISS	R/W	0x38...0x3B		0x38...0x3B	0x1C...0x1D	No update at cmd WCFG
CFGGEMAC	!	0x3C...0x3F		0x3C...0x3F	0x1E...0x1F	No update at cmd WCFG /write protect
BiSS-PAGE	R/W	-	-	0x40		SPI-Page
BiSS-EDS-Bank	R (EEP)	-		0x41	0x20 (MSB)	
BiSS-Profile	R (EEP)	-		0x42...0x43	0x21	
BiSS-Serial-Number	R (EEP)	-		0x44...0x47	0x22 ... 0x23	
EEP_DAT	R/W	0x48...0x49		0x48...0x49		
EEP_ADR / EEP_STAT	W / R	0x4A		0x4A		
EEP_OPC / EEP_MSB	W	0x4B		0x4B		
CFGTM	R/W	0x4C...0x4F		0x4C...0x4F		
CMD (16 Bit)	W	0x50...0x51		0x50...0x52		
TSTCMD (16 Bit)	W	0x52...0x53		0x52...0x53		Write protected
Unused	R/W / !	0x54...0x67		0x54...0x67		
Adjust3	!	0x68...0x6B		0x68...0x6B	0x34...0x35	Write protected
Adjust2	!	0x6C...0x6F		0x6C...0x6F	0x36...0x37	
Adjust1	!	0x70...0x73		0x70...0x73	0x38...0x39	
Adjust0	!	0x74...0x77		0x74...0x75	0x3A...0x3B	
BiSS-Device-Identifier	R (EEP)	-		0x78...0x7B	0x3C...0x3D	
BISS-Timeout	R/W	-		0x7C...0x7D	0x3E	
BiSS-Vendor-Identifier	R (EEP)	-		0x7E...0x7F	0x3F	
EDS-Common	R (EEP)	-	1	0x00...0x3F	0x40...0x5F	
EDS-Profil 1	R (EEP)	-	2	0x00...0x3F	0x60...0x7F	
EDS-Profil 2	R (EEP)	-	3	0x00...0x3F	0x80...0x9F	
Coefficients	R (SPI1)	0x40...0x7F	4	0x00...0x3F	0xA0...0xBF	
IP-Table	R (SPI1)	0x80...0xBF	5	0x00...0x3F	0xC0...0xDF	
IP-Table (fix)	R (SPI1)		6	0x00...0x3F	0xE0...0xFF	Write protected
MVAL	R	0x80	7	0x00...0x03		
CNT1	R	0x84		0x04...0x07		
POSIT	R	0x88		0x08...0x0B		
ADC1	R	0x8C		0x0C...0x0F		
CADC1	R	0x90		0x10...0x13		
IP11	R	0x94		0x14...0x17		
IP21	R	0x98		0x18...0x1B		
IP3	R	0x9C		0x1C...0x1F		
Unused	R	0xA0		0x20...0x23		
CNT2	R	0xA4		0x24...0x27		
Unused	R	0xA8		0x28...0x2B		
ADC 2	R	0xAC		0x2C...0x2F		
CADC2	R	0xB0		0x30...0x33		
IP12	R	0xB4		0x34...0x37		
IP22	R	0xB8		0x38...0x3C		
NONIUS	R	0xBC		0x3C...0x3F		
SPI-Page = SPI-MSB	R/W	0xFF	-	-	-	Any SPI-Page

¹⁾ R: Read only (register 32 Bit) W: Write only (register) R/W: Read/Write (register)
R (SPI): Read only via SPI-Page 1 !: Vendor register. shall/can not be changed!

²⁾ The EEPROM address is used for the internal EEPROM interface (register EEP).

dark gray: This register is loaded from EEPROM during reset

blue: BiSS-Information, direct read from EEPROM

white: EEPROM contains the valid identifier 0x134A

CNT1	Counter value (Interpolation counter) channel 1
CNT2	Counter value (Interpolation counter) channel 2

31:2	1	0
CNT	ZSTAT	ERR

Bit	Name	Reset value	Format	Value	Meaning
31:2	CNT	0	Signed	CNT	Counter value
1	ZSTAT	0	Bit	0	The reference mark (index) of the scale has not yet been passed or the reference of count value and reference mark was lost due to an error.
				1	The reference mark (index) of the scale has been passed; GC-NIP and scale operate synchronously.
0	ERR	0	Bit	0	Measured value is valid.
				1	An error occurred. The current measurement value and all subsequent values are to be discarded. After rectification of the error cause and resetting of the error bits (command RESCNT or PRESET pulse) it is imperative to pass through the reference point to be able to perform further absolute measurements.

POSIT	Position data (also: BiSS/SSI-SCD)
--------------	---

31:17	16:2	1	0
NONIUS		ERR	WARN
CNT1		ERR	WARN
CNT2		ERR	WARN
CNT2	CNT1	ERR	WARN

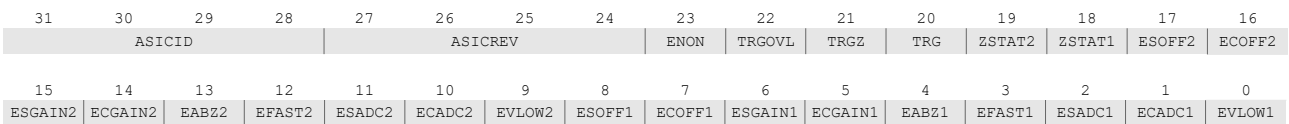
Bit	Name	Reset value	Format	Value	Meaning
31:2	NONIUS	0	Unsigned (optional: Gray)		CFGBISS/STSEL (1:0) = '00' Absolute position calculated by the nonius method. The data format is selected by CFGBISS/GRAY.
	CNT1	0	Signed (optional Gray)		CFGBISS/STSEL (1:0) = '01' Counter value of channel 1. The data format is selected by CFGBISS/GRAY.
	CNT2	0	Signed (optional Gray)		CFGBISS/STSEL (1:0) = '10' Counter value of channel 2. The data format is selected by CFGBISS/GRAY.
	CNT1 CNT2	0	Signed 15 Bit		CFGBISS/STSEL (1:0) = '11' Counter value of channel 1 and channel 2 (15 bit each). CFGBISS/GRAY should be 0.
1	ERR	0	Bit	0	Measured value is valid.
				1	An error occurred. The current measurement value and all subsequent values are to be discarded. After rectification of the error cause and resetting of the error bits (command RESCNT or PRESET pulse) it is imperative to pass through the reference point to be able to perform further absolute measurements.
0	WARN	0	Bit	0	Measured value is valid.
				1	The measured value has a limited accuracy.

MVAL Measured value / trigger value



Bit	Name	Reset value	Format	Value	Meaning
31:2	POSIT/TVAL	0	→ POSIT		Measured value; value corresponds to register POSIT or contents of a trigger holding register. A trigger holding register may be freed by reading this register. → see sections 7.7, 7.8
1	TRG	0	Bit	0	Measured value corresponds to content of register POSIT.
				1	Measured value corresponds to contents of a trigger holding register.
0	ERR	0	Bit	0	Measured value is valid.
				1	An error occurred. The current measurement value and all subsequent values are to be discarded. After rectification of the error cause and resetting of the error bits (command RESCNT or PRESET pulse) it is imperative to pass through the reference point to be able to perform further absolute measurements.

STAT / ID / REV ASIC identifier / status



Bit	Name	Reset value	Format	Value	Meaning
31:28	ASICID	0011	Binary	0011	The IC is a GC-NIP
27:24	ASICREV	0010	Binary		Silicon revision of the IC
23	ENON	0	Bit	0	No error in calculation of the nonius value. Measurement value is plausible.
				1	The calculated absolute position is invalid. Cause are either errors of the input signals, which can not be internally corrected, or unfavourable combinations of the correction coefficients stored in the EEPROM. The nonius sensor shall be calibrated.
22	TRGOVL	0	Bit	0	No overflow of the trigger holding register.
				1	Overflow of the trigger holding register; trigger event was lost.
21	TRGZ	0	Bit	0	Next measured value read from register MVAL was not triggered by the reference signal.
				1	Next measured value read from register MVAL was triggered by the reference signal.
20	TRG	0	Bit	0	Next measured value read from register MVAL was not triggered by the pin TRG.
				1	Next measured value read from register MVAL was triggered by the pin TRG.
19	ZSTAT2	0	Bit	0	The reference mark (channel 2) of the scale has not yet been passed or the reference of count value and reference mark was lost due to an error.
				1	The reference mark (channel 2) of the scale has been passed. Counter and scale operate synchronously.
18	ZSTAT1	0	Bit	0	The reference mark (channel 1) of the scale has not yet been passed or the reference of count value and reference mark was lost due to an error.
				1	The reference mark (channel 1) of the scale has been passed. Counter and scale operate synchronously.
17	ESOFF2	0	Bit	0	No offset error at sinusoidal signal at channel 2
				1	The offset controller for the sinusoidal signal has reached its limit. The cause is an excessive signal offset, a partly of fully disconnected sensor on an invalid value for initialization of the controller.
16	ECOFF2	0	Bit	0	No offset error at cosinusoidal signal at channel 2
				1	The offset controller for the cosinusoidal signal has reached its limit. The cause is an excessive signal offset, a partly of fully disconnected sensor on an invalid value for initialization of the controller.
15	ESGAIN2	0	Bit	0	No amplitude error at sinusoidal signal at channel 2
				1	The gain controller for the sinusoidal signal has reached its limit. The cause is either that the signal amplitude is too low or the sensor is partly or fully disconnected.
14	ECGAIN2	0	Bit	0	No amplitude error at sinusoidal signal at channel 2

Bit	Name	Reset value	Format	Value	Meaning
				1	The gain controller for the cosinusoidal signal has reached its limit. The cause is either that the signal amplitude is too low or the sensor is partly or fully disconnected.
13	EABZ2	0	Bit	0	No error at A,B,Z channel 2
				1	The signals A, B, and Z are invalid. The cause is an excessive input frequency. The monitored frequency depends on the set minimum edge interval t_{pp} . This error also occurs if the interpolation rate or the minimum edge interval is changed. Detection of this error is deactivated for the counter mode.
12	EFAST2	0	Bit	0	No speed error at channel 2
				1	The input frequency is so high that no A/B signals can be generated or the direction can no longer be detected. The monitored frequency is different depending on whether an internal counter or the square-wave outputs A/B/Z are used.
11	ESADC2	0	Bit	0	No ADC error at the sinusoidal signal at channel 2
				1	The A/D converter for the sinusoidal signal is overdriven. The cause is that the signal amplitude is too high. This error may also occur with signals with very large offset at simultaneously high amplitude.
10	ECADC2	0	Bit	0	No ADC error at the cosinusoidal signal at channel 2
				1	The A/D converter for the cosinusoidal signal at is overdriven. The cause is that the signal amplitude is too high. This error may also occur with signals with very large offset at simultaneously high amplitude.
9	EVLOW2	0	Bit	0	No vector error at channel 2
				1	The signal vector generated from the sinusoidal and cosinusoidal signals is too small. Usually, the cause is a partly or completely disconnected sensor. This error may also occur with signals with very large offset at simultaneously low amplitude.
8	ESOFF1	0	Bit	0	No offset error at sinusoidal signal at channel 1
				1	The offset controller for the sinusoidal signal has reached its limit. The cause is an excessive signal offset, a partly of fully disconnected sensor on an invalid value for initialization of the controller.
7	ECOFF1	0	Bit	0	No offset error at sinusoidal signal at channel 1
				1	The offset controller for the cosinusoidal signal has reached its limit. The cause is an excessive signal offset, a partly of fully disconnected sensor on an invalid value for initialization of the controller.
6	ESGAIN1	0	Bit	0	No amplitude error at sinusoidal signal at channel 1
				1	The gain controller for the sinusoidal signal has reached its limit. The cause is either that the signal amplitude is too low or the sensor is partly or fully disconnected.
5	ECGAIN1	0	Bit	0	No amplitude error at sinusoidal signal at channel 1
				1	The gain controller for the cosinusoidal signal has reached its limit. The cause is either that the signal amplitude is too low or the sensor is partly or fully disconnected.
4	EABZ1	0	Bit	0	No error at A,B,Z channel 1
				1	The signals A, B and Z are invalid. The cause is an excessive input frequency. The monitored frequency depends on the set minimum edge interval t_{pp} . This error also occurs if the interpolation rate or the minimum edge interval is changed. Detection of this error is deactivated for the counter mode.
3	EFAST1	0	Bit	0	No speed error at channel 1
				1	The input frequency is so high that no A/B signals can be generated or the direction can no longer be detected. The monitored frequency is different depending on whether an internal counter or the square-wave outputs A/B/Z are used.
2	ESADC1	0	Bit	0	No ADC error at the sinusoidal signal at channel 1
				1	The A/D converter for the sinusoidal signal is overdriven. The cause is that the signal amplitude is too high. This error may also occur with signals with very large offset at simultaneously high amplitude.
1	ECADC1	0	Bit	0	No ADC error at the sinusoidal signal at channel 1
				1	The A/D converter for the cosinusoidal signal is overdriven. The cause is that the signal amplitude is too high. This error may also occur with signals with very large offset at simultaneously high amplitude.
0	EVLOW1	0	Bit	0	No vector error at channel 1
				1	The signal vector generated from the sinusoidal and cosinusoidal signals is too small. Usually, the cause is a partly or completely disconnected sensor. This error may also occur with signals with very large offset at simultaneously low amplitude.

CMD		Command					
7	6	5	4	3	2	1	0
TRGCAL	PRESET	SETHWA	WCFG	RESIC	CLRZ	RESCTL	RESCNT
Bit	Name	Reset value	Format	Value	Meaning		
7	TRGCAL		Bit write-only	1	The registers used for calibration are refreshed. Also, the registers are refreshed if the trigger (Pin TRG) is active.		
6	PRESET		Bit write-only	1	The offset for the nonius position NONOFFS is re-calculated using the preset value for the nonius from register PREST2 and the new offset value (NONOFFS) is stored in the EEPROM. The bit CFG2/PREENA has to be set for this function.		
5	SETHWA		Bit write-only	1	The pins HWA3, HWA2, HWA1 and HWA0 are read into the IC as hardware addresses. If several ICs are to be connected to one SPI interface, this command must be sent first to all connected ICs. This command is automatically set during initialization if the BiSS interface is activated.		
4	WCFG		Bit write-only	1	The content of the registers CFG1, CFG2, CFG3, CNTRLG, CNTRLO, PREST1, and PREST2 are transferred to the EEPROM.		
3	RESIC		Bit write-only	1	The IC is reset and re-configured.		
2	CLRZ		Bit write-only	1	The status bit ZSTAT is reset. For the trigger modes "Adjustment Z" and "Distance coded" this command starts a new measurement (see 7.7).		
1	RESCTL		Bit write-only	1	The internal controller for gain and offset is reset, i.e. all correction values for offset and gain are set to the center of their value range.		
0	RESCNT		Bit write-only	1	The counter values (CNT1/CNT2) are set to the value in the registers PREST1/PREST2. All error flags in the status register are reset and the bit ZSTAT is reset. For the trigger modes "Adjustment Z" and "Distance coded" this command starts a new measurement (see 7.7).		

CFG1				Configuration 1											
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
TRI	LNON	LOFF	LGAIN	LABZ	LFAST	LADC	LVLOW	HLD	MNON	MOFF	MGAIN	MABZ	MFAST	MADC	MVLOW
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
GAIN		CFGAF		MODE				CFGTPP			TRGSLP	IRMAP	IRD2SEL	IRDIV1	

Bit	Name	Reset value	Format	Value	Meaning
31	TRI	0	Bit	0	The behaviour of the signals A, B, and Z is determined by bit HLD in case of error.
				1	The signals A, B, and Z are high resistant in case of error.
30	LNON	0	Bit	0	Detected nonius errors (ENON) are not saved.
				1	Detected nonius errors (ENON) are saved.
29	LOFF	0	Bit	0	Detected offset errors (EOFF) are not saved.
				1	Detected offset errors (EOFF) are saved.
28	LGAIN	0	Bit	0	Detected gain errors (EGAIN) are not saved.
				1	Detected gain errors (EGAIN) are saved.
27	LABZ	0	Bit	0	Detected A/B/Z errors (EABZ) are not saved.
				1	Detected A/B/Z errors (EABZ) are saved.
26	LFAST	0	Bit	0	Detected speed errors (EFAST) are not saved.
				1	Detected speed errors (EFAST) are saved.
25	LADC	0	Bit	0	Detected ADC errors (EADC) are not saved.
				1	Detected ADC errors (EADC) are saved.
24	LVLOW	0	Bit	0	Detected vector errors (ELVLOW) are not saved.
				1	Detected vector errors (ELVLOW) are saved.
23	HLD	1	Bit	0	The behaviour of the signals A,B and Z is not defined in case of error.
				1	The signals A,B and Z do not change in case of error.
22	MNON	1	Bit	0	The detection of nonius errors (ENON) is deactivated.
				1	The detection of nonius errors (ENON) is activated.
21	MOFF	1	Bit	0	The detection of offset errors (EOFF) is deactivated.
				1	The detection of offset errors (EOFF) is activated.
20	MGAIN	1	Bit	0	The detection of gain errors (EGAIN) is deactivated.
				1	The detection of gain errors (EGAIN) is activated.
19	MABZ	0 (EEP) 1 (Pin)	Bit	0	The detection of A/B/Z errors (EABZ) is deactivated; the IC operates in the counter mode.
				1	The detection of A/B/Z errors (EABZ) is activated; the IC operates in the square-wave mode.
18	MFAST	1	Bit	0	The detection of speed errors (EFAST) is deactivated.
				1	The detection of speed errors (EFAST) is activated.
17	MADC	1	Bit	0	The detection of ADC errors (EADC) is deactivated.
				1	The detection of ADC errors (EADC) is activated.
16	MVLOW	1	Bit	0	The detection of vector errors (ELVLOW) is deactivated.
				1	The detection of vector errors (ELVLOW) is activated.
15:14	GAIN	Pins CFGGAIN	binary	00	Nominal signal amplitude 660 mV _{pp}
				01	Nominal signal amplitude 250 mV _{pp}
				10	Nominal signal amplitude 120 mV _{pp}
				11	Nominal signal amplitude 60 mV _{pp}
13:12	CFGAF	Pins CFGAF	binary	00	Analogue low-pass filter cut-off frequency 150kHz -0.5dB
				01	Analogue low-pass filter cut-off frequency 75kHz -1dB
				10	Analogue low-pass filter cut-off frequency 10kHz -1dB
				11	Analogue low-pass filter inactive
11:8	MODE	0000	binary		Configuration of the output signals and operating mode as per Table 19
7:5	CFGTPP	000	binary	TPP	Configuration of the minimum edge interval t _{pp} ; t _{pp} = 2 ^{TPP} / f _{osz} see Table 58
				000	Range low: t _{pp} = 1/ f _{osz}
				111	Range high: t _{pp} = 128/ f _{osz}
4	TRGSLP	0	Bit	0	A falling edge at pin TRG accepts the measured value into the trigger holding register.
				1	A rising edge at pin TRG accepts the measured value into the trigger holding register.
3	IRMAP	0 (EEP) 1 (Pin)	Bit	0	Base interpolation rate and nonius pitch are read from EEPROM
				1	Base interpolation rate = 2000, nonius pitch = 125
2	IRD2SEL	0	Bit	0	The divider factor for the interpolation rate of channel 1 and 2 are the same
				1	The divider factor for the interpolation rate of channel 2 is set by CFG3/IRDIV2.
1:0	IRDIV1	00	binary	N	The base interpolation rate (set by IRMAP) used for the counter and the A/B output is divided by 2 ^N .
				00	2 ^N - divider 1
				01	2 ^N - divider 2
				10	2 ^N - divider 4
				11	2 ^N - divider 8

CFG2		Configuration 2													
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
DISMON	DISV0	PH_2				PREENA	PHBER	PH_1							
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
ASYNC	SYNC						ZMODE	Z4	ZPOS						

Bit	Name	Reset value	Format	Value	Meaning
31	DISMON	0	Bit	0	The pins SMON1, SMON2, CMON1, and CMON2 are active.
				1	The pins SMON1, SMON2, CMON1, and CMON2 are inactive (power saving option).
30	DISV0	0	Bit	0	The pins VOCH1, VOCH2 are active.
				1	The pins VOCH1, VOCH2 are inactive (power saving option).
29:24	PH_2	0	signed	-32	Largest phase displacement negative.
				PH	Setting value of the phase correction potentiometer of channel 2.
				+31	Largest phase displacement positive.
23	PREENA	0	Bit	0	The preset function for calculating a new offset value is inactive.
				1	The preset function for calculating a new offset value is active (see 7.10.1).
22	PHBER	0	Bit	0	The setting range of the phase correction potentiometer is $\pm 5^\circ$. The step size is 0.156° .
				1	The setting range of the phase correction potentiometer is $\pm 10^\circ$. The step size is 0.313° .
21:16	PH_1	0	signed	-32	Largest phase displacement negative.
				PH	Setting value of the phase correction potentiometer of channel 1.
				+31	Largest phase displacement positive.
15	ASYNC	0	Bit	0	The data to be read are accepted into a 32-bit holding register synchronously to the internal sequence using the SPI word RD0/ST. The time of acceptance can be shifted relative to the sampling time using the value of SYNC.
				1	Data to be read are accepted asynchronously into a 32-bit holding register using the SPI word RD0/ST. The value of SYNC is not evaluated.
14:8	SYNC	0	unsigned		Displacement of an SPI read access relative to the sampling time. To read the registers IP11, IP12, IP21, IP22, Nonius and to read at SPI page 1 a value of 64 (dec) must be used.
7:6	ZMODE	00	binary	00	Reference point evaluation mode Incremental
				01	Reference point evaluation mode Trigger
				10	Reference point evaluation mode Adjustment Z
				11	Reference point evaluation mode Distance coded
5	Z4	0	Bit	0	The width of the zero signal Z is one increment = $\frac{1}{4}$ period
				1	The width of the zero signal Z is four increments = 1 period
4:0	ZPOS	00100 (45°)	unsigned	0	Reference position = 0°
				ZPOS	Displacement of the reference point position according to the sinusoidal signal at the input. Reference position = $ZPOS \cdot 11.25^\circ$ This register is valid only if CFG3/NOSEL = 0 (see section 7.6.3)
				31	Reference position = 348.75°

CFG3				Configuration 3											
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
-	-	-	-	-	-	-	-	-	-	-	-	-	NOSEL	IRDIV2	
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
DISCH2	MXFEED	MXSHR	ZDEL2	ZDEL	DISZ2	DISZ1		DH		OFFSCTL		GAINCTL		DISCTL	

Bit	Name	Reset value	Format	Value	Meaning
18	NOSEL	0	Bit	0	The register <code>NONOFFS</code> is used for calculation of the absolute position (nonius). The signal <code>PRESET</code> causes the re-calculation of the offset value.
				1	The register <code>NONOFFS</code> contains the reference point position for channel 1 and channel 2. the signal <code>PRESET</code> causes the re-calculation of the reference point positions.
17:16	IRDIV2	00	binary	N	The base interpolation rate (set by <code>IRMAP</code>) used for the counter and the A/B output on channel 2 is divided by 2^N , if enabled by <code>CFG1/IRD2SEL = 1</code> .
				00	$2^N \rightarrow$ divider 1
				01	$2^N \rightarrow$ divider 2
				10	$2^N \rightarrow$ divider 4
15:14	DISCH2	00	Bit	00	Nonius, phase angle, ABZ outputs ¹⁾ and counter of channel 2 are calculated. → Nonius, one channel and two channel operating mode possible.
				01	Counter and ABZ outputs ⁰⁾ of channel 2 are not calculated. The phase angle value of channel 2 is calculated. → Nonius and one channel operating mode possible.
				10	invalid
				11	Nonius, phase angle, ABZ outputs ¹⁾ and counter of channel 2 are not calculated. → One channel operating mode possible.
13	MXFEED	0	Bit	0	The correction value is calculated from the coefficients stored in EEPROM.
				1	Coefficient 14 is applied as correction value. → One channel and two channel operating mode possible.
12	MXSHR	1	Bit	0	Coefficient scaling 16 Bit
				1	Coefficient scaling 16/18 Bit
11	ZDEL2	0	Bit	0	Default value
				1	Additional internal delay of the reference signal Z of $96/f_{osZ}$.
10	ZDEL	0	Bit	0	Default value
				1	Additional internal delay of the reference signal Z of $96/f_{osZ}$.
9	DISZ2	00	Bit	0	Reference point processing at channel 2 is activated. Activation (<code>DISZ2</code> changes from 1 to 0) requires a processing time of 100µs.
				1	Reference point processing at channel 2 is deactivated.
8	DISZ1	00	Bit	0	Reference point processing at channel 1 is activated. Activation (<code>DISZ1</code> changes from 1 to 0) requires a processing time of 100µs
				1	Reference point processing at channel 1 is deactivated.
7:5	DH	01	binary	000	A value of 0 deactivates the digital hysteresis.
				DH	Threshold value for the digital hysteresis.
				111	Highest digital hysteresis threshold value 7
4:3	OFFSCTL	01	binary	00	Maximum settling time for the offset controller. This configuration must be selected if the sensor signal has a lower input frequency or is overlaid by noise, or the phase between sinusoidal and cosinusoidal signals cannot be fully adjusted using the phase correction potentiometer.
				01	Reduction of the settling time of the offset controller by a factor of approx. 2
				10	Reduction of the settling time of the offset controller by a factor of approx. 4
				11	Reduction of the settling time of the offset controller by a factor of approx. 8
2:1	GAINCTL	01	binary	00	Maximum settling time for the gain controller. This configuration must be selected if the sensor signal has a lower input frequency or is overlaid by noise, or the phase between sinusoidal and cosinusoidal signals cannot be fully adjusted using the phase correction potentiometer.
				01	Reduction of the settling time of the gain controller by a factor of approx. 2
				10	Reduction of the settling time of the gain controller by a factor of approx. 4
				11	Reduction of the settling time of the gain controller by a factor of approx. 8
0	DISCTL	0	Bit	0	The internal controller for gain and offset is activated.
				1	The internal controller for gain and offset is deactivated.

¹⁾ The ABZ output is de-/activated after reset of the IC.

CFGBISS Configuration SSI and BiSS interface

31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16	
SSI	-	SSI20	RING	SSITO (11:0)												
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	
STSEL1	STSEL0	GRAY	STBIT					-	READ32	-	BISSTO (4:0)					

Bit	Name	Reset value	Format	Value	Meaning
31	SSI	1	Bit	0	BiSS is used as digital interface if BiSS/SSI is enabled by SEN=Low during reset.
				1	SSI is used as digital interface if BiSS/SSI is enabled by SEN=Low during reset.
29	SSI20	1		0	32 Bit SSI-Data
				1	20 Bit SSI-Data
28	RING	1	Bit	0	SSI Ring mode inactive
				1	SSI Ring mode active
27:16	SSITO	517 decimal	unsigned	SSITO	Configuration SSI timeout parameter Timeout = (SSITO+3)/f _{OSZ} or SSITO = (Timeout·f _{OSZ})-3 Example: f _{OSZ} = 26MHz → SSITO = 23(1µs) ... 517(20µs) ... 2047 (79 µs)
15:14	STSEL	00	binary	00	The position value is determined by the nonius calculation.
				01	The counter value of channel 1 is used as position value.
				10	The counter value of channel 2 is used as position value.
				11	For debug purpose only: / the position value contains the counter values of channel 1 and channel 2 with 16 bit each. Gray-coding shall not be used.
13	GRAY	0	Bit	0	The position data is binary coded.
				1	The position data is gray coded.
12:8	STBIT	30 decimal	binary	STBIT	Resolution of the position data in bits; range: 8-30 bit. Unused MSB are filled with zero This value has no effect if STSEL = 11.
6	READ32	0	Bit	0	8 bit BiSS read access. Suitable for reading configuration registers.
				1	32 bit BiSS read access. 4 subsequent addresses, beginning with the least significant address (divisible by 4), must be read. Suitable for reading data and user registers.
4:0	BISSTO	9 decimal	unsigned	BISSTO	Configuration of the BiSS-Timeout; values: 12µs...40µs. Timeout = 2 ^{BISSTO} /f _{OSZ} or BISSTO = log ₂ (Timeout·f _{OSZ}) Example: f _{OSZ} = 26MHz → BISSTO = 9(19.7µs) or 10(39.4µs)

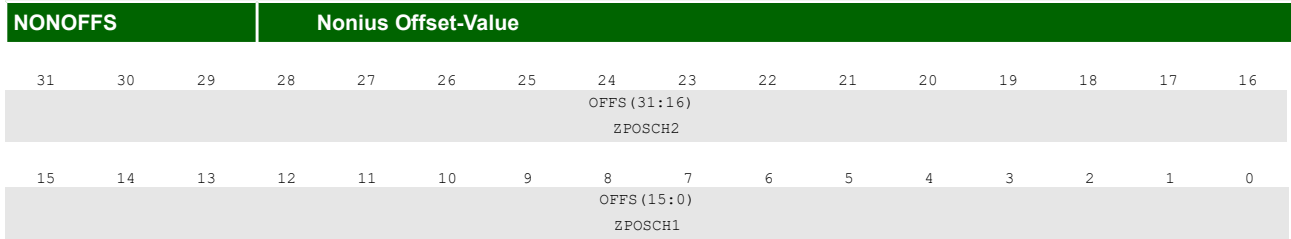
ⓘ This register has to be configured in EEPROM via SPI to ensure correct BiSS functionality.

PREST1 Preset-Value channel 1

PREST2 Preset-Value channel 2 / nonius

31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
PRE (31:16)															
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
PRE (15:0) / PRE_NONIUS															

Bit	Name	Reset value	Format	Value	Meaning
31:0	PRE	0	unsigned	PRE	Preload-Value; → see section 7.10



Bit	Name	Reset value	Format	Value	Meaning
31:0	OFFS	0	unsigned	OFFS	Offset value for nonius calculation → see section 7.10
31:16	ZPOSCH2	0	unsigned	ZPOS	Configuration of the reference position referred to the sinusoidal signal. Condition: CFG3/NOSEL = 1 → see section 7.6.3, 7.10.2
15:0	ZPOSCH1	0	unsigned	0 IRATE-1	

CNTRLG1	Controller: Gain correction value channel 1
CNTRLG2	Controller: Gain correction value channel 2

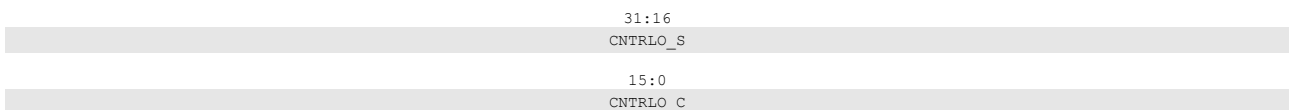
When writing the bits 26:16, the bits 23:16 must be written first. Subsequently, the whole correction value is refreshed in the register by writing of the bits 26:24.
 When writing the bits 10:0, the bits 7:0 must be written first. Subsequently, the whole correction value is refreshed in the register by writing of the bits 10:8.
 Please note that the correction values are changed automatically by the IC with active signal control.



Bit	Name	Reset value	Format	Value	Meaning
26:16	CNTRLG_S	0x400	unsigned	CNTRLG	$CADC_S = [ADC_S + CNTRLO_S] \cdot (0.5 + CNTRLG_S/2048)$ $CADC_C = [ADC_C + CNTRLO_C] \cdot (0.5 + CNTRLG_C/2048)$ Minimum value; the offset-corrected ADC values of the sinusoidal signal are multiplied by 0.5.
10:0	CNTRLG_C	0x400	unsigned	0x400 0x7FF	

CNTRLO1	Controller: Offset correction value channel 1
CNTRLO2	Controller: Offset correction value channel 2

When writing the bits 31:16, the bits 23:16 must be written first. Subsequently, the whole correction value is refreshed in the register by writing of the bits 31:24. If the value to be written lies outside the valid range of -2730...+2729, the correction register is no longer refreshed, and the bit **ESOFF** in the register **STAT/ERR** is set.
 When writing the bits 15:0, the bits 7:0 must be written first. Subsequently, the whole correction value is refreshed in the register by writing of the bits 15:8. If the value to be written lies outside the valid range of -2730...+2729, the correction register is no longer refreshed, and the bit **ECOFF** in the register **STAT/ERR** is set.
 Please note that the correction values are changed automatically by the IC with active signal control.



Bit	Name	Reset value	Format	Value	Meaning
31:16	CNTRLO_S	0x0000	signed	CNTRLO	$CADC_S = [ADC_S + CNTRLO_S] \cdot (0.5 + CNTRLG_S/2048)$ $CADC_C = [ADC_C + CNTRLO_C] \cdot (0.5 + CNTRLG_C/2048)$ Minimum value -2730
15:0	CNTRLO_C	0x0000	signed	0x0000 0x0AA9	

ADC1	ADC values channel 1
ADC2	ADC values channel 2

31:16	ADC_S
15:0	ADC_C

Bit	Name	Reset value	Format	Value	Meaning
31:16	ADC_S	-	signed	0xE000	Minimum value -8192; corresponds to a differential voltage of approx. -495mV at the input of the instrumentation amplifier (@GAIN=00).
				0x0000	Mean value 0; corresponds to a differential voltage of approx. 0mV at the input of the instrumentation amplifier.
15:0	ADC_C	-	signed	0x1FFF	Maximum value +8191; corresponds to a differential voltage of approx. +495mV at the input of the instrumentation amplifier (@GAIN=00).

CADC1	Corrected ADC values channel 1
CADC2	Corrected ADC values channel 2

In calibration mode (CFG1/MODE = 0101), the registers are refreshed by command TRGCAL or by an edge at the pin TRG. In the other operational modes, the registers always contain the actual corrected ADC-values.

31	30	29:16
VZ (CADC_S)	0	Abs (CADC_S)
15	14	13:0
VZ (CADC_C)	0	Abs (CADC_C)

Bit	Name	Reset value	Format	Value	Meaning
31	VZ(CADC_S)	-	Bit	0 1	Corrected ADC value sinusoidal ≥ 0 Corrected ADC value sinusoidal < 0
29:16	Abs(CADC_S)	-	unsigned	0 0x3FFF	Corrected ADC value sinusoidal (absolute value) Minimum value Maximum value
15	VZ(CADC_C)	-	Bit	0 1	Corrected ADC value cosinusoidal ≥ 0 Corrected ADC value cosinusoidal < 0
13:0	Abs(CADC_C)	-	unsigned	0 0x3FFF	Corrected ADC value cosinusoidal (absolute value) Minimum value Maximum value

Bit	Name	Reset value	Format	Value	Meaning
31:16	CADC_S	-	Sign + absolute value	CADC	Corrected ADC value sinusoidal $CADC_S = [ADC_S + CNTRLO_S] \cdot (0.5 + CNTRLG_S/2048)$
15:0	CADC_C	-	Sign + absolute value	CADC	Corrected ADC value cosinusoidal $CADC_C = [ADC_C + CNTRLO_C] \cdot (0.5 + CNTRLG_C/2048)$

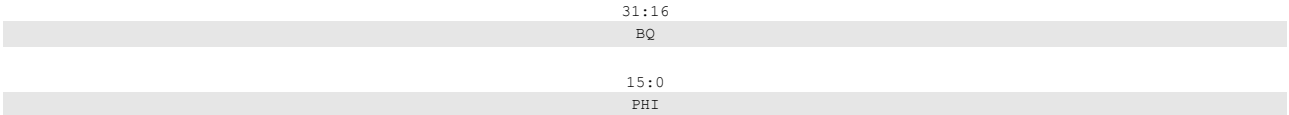
IP11	Interpolation register 1 – Angular value / angle difference – channel 1
IP12	Interpolation register 1 – Angular value / angle difference – channel 2

31:16	DPHI
15:0	PHI

Bit	Name	Reset value	Format	Value	Meaning
31:16	DPHI	-	signed	DPHI	The value DPHI is the difference of the phase angle of sinusoidal and cosinusoidal signals between two samplings. The range of values is dependent on the set interpolation rate. This value represents the speed of the measuring system. $f_{input} = DPHI/(96 \cdot IRATE) \cdot f_{OSZ}$
15:0	PHI	-	unsigned	0x0000 IRATE-1	The phase angle of sinusoidal and cosinusoidal signal is 0° The phase angle of sinusoidal and cosinusoidal signal is 360° - ε

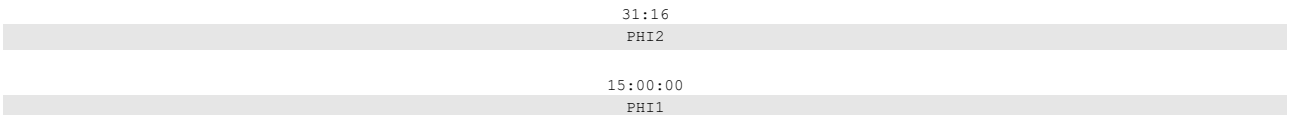
IP21	Interpolation register 2 – Angular value / controller value – channel 1
IP22	Interpolation register 2 – Angular value / controller value – channel 2

In calibration mode (CFG1/MODE = 0101) this register contains a quadrant counter at the bits 31:16, which is refreshed by a trigger event or by the SPI command TRGCAL.



Bit	Name	Reset value	Format	Value	Meaning
31:16	BQ	-	unsigned	BQ	The value BQ contains the deviation of the gain and offset controller from the setpoint. If offset and gain are adjusted completely, the value of this register is 321.
15:0	PHI	-	unsigned	0x0000 IRATE-1	The phase angle of sinusoidal and cosinusoidal signal is 0° The phase angle of sinusoidal and cosinusoidal signal is 360° - ε

IP3	Interpolation register 3 – Angular value 1 / Angular value 2
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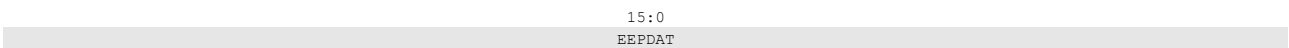


Bit	Name	Reset value	Format	Value	Meaning
31:16	PHI2	-	unsigned	0x0000 IRATE-1	The phase angle of sinusoidal and cosinusoidal signal at channel 2 is 0° The phase angle of sinusoidal and cosinusoidal signal at channel 2 is 360° - ε
15:0	PHI1	-	unsigned	0x0000 IRATE-1	The phase angle of sinusoidal and cosinusoidal signal at channel 1 is 0° The phase angle of sinusoidal and cosinusoidal signal at channel 1 is 360° - ε

EEP	EEPROM-Interface
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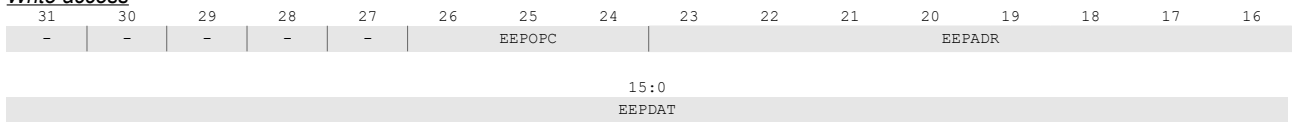
Read access

31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
-	-	-	-	-	-	-	INIT	-	AUT	RSV	EEPVALID	EER	EWR	EEPBSY	EEPLOCK



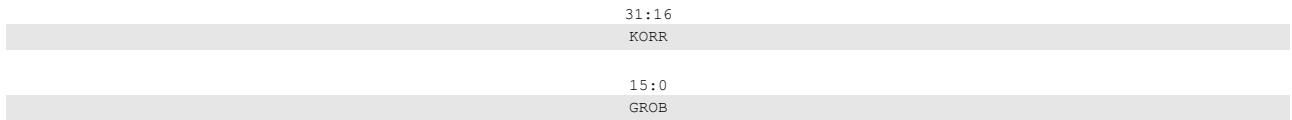
Bit	Name	Reset value	Format	Value	Meaning
24	INIT		Bit	1	This bit is reserved for test purposes.
22	AUT	0	Bit		This bit is reserved for test purposes.
21	RSV	0	Bit		This bit is reserved for test purposes.
20	EEPVALID	0/1	Bit	0	EEPROM valid identifier 0x134A at address 0x00 was not found
				1	EEPROM valid identifier 0x134A at address 0x00 was found. EEPROM content has been loaded into the registers during reset.
19	EER	0	Bit	0	No EEPROM delete access active.
				1	EEPROM delete access active.
18	EWR	0	Bit	0	No EEPROM write access active.
				1	EEPROM write access active.
17	EEPBSY	0	Bit	0	No EEPROM access active.
				1	EEPROM access active; No further command may be sent to the EEPROM.
16	EEPLOCK	0	Bit	0	The EPROM is free for use.
				1	EEPROM locked
15:0	EEPDAT	0x0000	binary		EEPROM-Data

Write access



Bit	Name	Reset value	Format	Value	Meaning
26:24	EEPOPC	000	binary		EEP-OPCode; Writing to this register triggers an EEPROM access. The register must not be written if the <code>EEPBSY</code> is set. <code>EEPADR</code> and <code>EEPDAT</code> must be valid.
				000	NOP – No action
				001	WRITE – write 16 bit
				010	READ – read 16 bit
				100	ERASE – delete 16 bit
	other	Undefined behaviour. EEPROM content may be lost.			
23:16	EEPADR	0x00	binary		EEPROM address. To program or read the EEPROM, the address must be written to this register before activating the OPCode. The register must not be written if the bit <code>EEPBSY</code> is active.
15:0	EEPDAT	0x0000	binary		EEPROM-Data; To program the EEPROM, the data must be written to this register before activating the OP code. The register must not be written if the bit <code>EEPBSY</code> is set.

NONIUS *Nonius register*



Bit	Name	Reset value	Format	Value	Meaning
31:16	KORR	-	signed		Correction value calculated from the coefficients and the input signals. For test purpose only.
15:0	GROB	-	unsigned	0x0000 IRATE-1	Corrected nonius position (PHI1-PHI2) with a value range of 0 ... IRATE-1 For test purpose only.

10 Characteristic values

Table 42: Absolute maximum ratings

Symbol	Characteristic value	Min.	Typ.	Max.	Unit
VDDA	Analogue supply voltage			TBD ¹⁾	V
VDD	Digital supply voltage			TBD ¹⁾	V
TJ	Operating temperature	-40		125	°C
TS	Storage temperature	-55		150	°C
V(AIN)	Voltage at the analogue inputs	-0.3		VDDA+0.3	V
V(DIN)	Voltage at the digital inputs	-0.3		VDDIO+0.3	V
ESD	ESD sensitivity (HBM)			2	kV

¹⁾ t < 250ms, T < 60°C

Table 43: Operating conditions

Symbol	Characteristic value	Min.	Typ.	Max.	Unit
VDDA	Analogue supply voltage	3.15 (3.0) ¹⁾	3.30	3.60	V
VDD	Digital supply voltage	3.00	3.30	3.60	V
I(VDDA)	Current consumption, analogue		25		mA
I(VDD)	Current consumption, digital		25		mA
T	Operating temperature	-40		100 (125) ¹⁾	°C

¹⁾ Controller ranges and interpolation accuracy are limited between 3.0V and 3.15V resp. between 100°C and 125°C.

Table 44: Characteristic values clock / reset

Symbol	Characteristic value	Min.	Typ.	Max.	Unit
f _{OSZ}	External Clock : frequency	4		26	MHz
TH/TL	External Clock : duty-cycle	40	50	60	%
t _{INIT}	Initialization time Time between NRES rising edge and Ready (MISO, NERR)		40	80	ms

Table 45: Characteristic values for interpolation

Symbol	Characteristic value	Min.	Typ.	Max.	Unit
f _{IP}	Input frequency (nonius calculation)	0		150	kHz
f _{IP}	Input frequency (interpolation)	0		90	kHz
IRATE	Interpolation rate (nonius calculation)	256		8192	Increments
IRATE_AB	Interpolation rate (ABZ)	32		8192	Increments
CTRL(A)	Amplitude control	60		120	%VINNOM ²⁾
CTRL(O)	Offset control	-15		15	%VINNOM ²⁾
VTH(INP)	Threshold voltage for vector monitoring		30		%VINNOM ²⁾
EABS	Absolute angle error ¹⁾		TBD	TBD	Increments
EDIFF	Differential angle error ¹⁾		TBD		Increments
t _{pp}	Minimum edge distance A/B	1/f _{OSZ}		128/f _{OSZ}	ns
t(TRG)	Pulse width of the trigger signal	3/f _{OSZ}			ns
t _p (Preset)	Pulse width PRESET signal	60			ms
td(CNT)	Delay time 'Analogue input to nonius result'		181/f _{OSZ} + 100		ns
td(ABZ)	Delay time 'Analogue input to A/B'		208/f _{OSZ} + 100		ns

¹⁾ Input voltage range 0.66 V_{pp} / matched phase deviation between sinusoidal and cosinusoidal signal

²⁾ Nominal value of the differential voltage of SINP-SINN or COSP-COSN

Table 46: Digital characteristic values

Symbol	Characteristic value	Min.	Typ.	Max.	Unit
VOH	Output voltage H	80			%VDDIO
VOL	Output voltage L			0.4	V
VIH	Input voltage H	70			%VDDIO
VIL	Input voltage L			30	%VDDIO
I(DIG1)	Output current digital			6	mA
I(DIG2)	Output current digital at MISO and NERR			12	mA
R(PU)	Internal Pull-Up resistors	90k		210	K Ω
R(PD)	Internal Pull-Down resistors	75k		250	K Ω

Table 47: Analogue characteristic values

Symbol	Characteristic value	Min.	Typ.	Max.	Unit
Z(AIN)	Input impedance		1G Ω 8pF		
Gain	Gain factor as per 10 @1kHz	97	100	103	%
f _g	Cut-off-frequency analogue low-pass filter according to 11	90	100	110	%
f _{gMATCH}	Deviation of the cut-off-frequencies among the channels	-1	0	+1	%
V(AIN)	Voltage at the analogue inputs	0.35		VDDA-1.0	V
CMIR	Common mode input voltage		1.1		V
CMRR	Common mode rejection ratio (@ f < 1kHz, GAIN maximum)	65			dB
V(V0)	Voltage at pin V0 / DC-voltage at SMON/CMON	1.08	1.1	1.12	V
VMON	AC-voltage at SMON/CMON @ nominal amplitude		1.27		V _{pp}
I(V0)	Output current at V0CH1/2			1	mA
CL(V0)	Capacitive load at Pin V0CH1/2			300	pF
VTH(REF)	Switching threshold reference-point-comparator ²⁾	-1		1	mV
VH(REF)	Hysteresis reference-point-comparator ²⁾		15		%VINNOM ¹⁾
I(OUTX)	Output current at pin SMON1/CMON1/SMON2/CMON2			0.5	mA
CL(OUTX)	Capacitive load at pin SMON1/CMON1/SMON2/CMON2			50	pF
ϕ K1	Phase correction in range 1	± 4.5	± 5	± 5.5	$^{\circ}$
ϕ K2	Phase correction in range 2	± 9	± 10	± 11	$^{\circ}$

¹⁾ Nominal value of the difference voltage of SINP-SINN or COSP-COSN

²⁾ Voltage difference REFPP-REFN

Table 48: Characteristic values EEPROM

Symbol	Characteristic value	Min.	Typ.	Max.	Unit
t _{READEEP}	EEPROM read duration	20		85	us
t _{PROGEEP}	Programming time / delete time	4	9.5	11	ms
t _{RETENTIONEEP}	Data retention @ T < 85 $^{\circ}$	10			Years
N _{ProgEEP}	Endurance @ T = 25 $^{\circ}$ @ T = 125 $^{\circ}$	10 ⁴ 10 ³			Write cycles

Table 49: Characteristic values SSI interface

Symbol	Characteristic value	Min.	Typ.	Max.	Unit
f _{MA}	Clock frequency @ f _{OSZ} \geq 4MHz @ f _{OSZ} \geq 8MHz @ f _{OSZ} \geq 10MHz @ f _{OSZ} \geq 20MHz			2 3 4 5	MHz
t _D (MISO)	Delay time MA rising until SLO			25	ns
t _{TIMEOUT}	Timeout \rightarrow CFGBISS	3/f _{OSZ}	10	4095 / f _{OSZ}	us

Table 50: Characteristic values BiSS interface

Symbol	Characteristic value	Min.	Typ.	Max.	Unit
f_{MA}	Clock frequency			10	MHz
$t_D(\text{MISOBISS})$	Delay time M_A rising until SLO			20	ns
t_{BUSY_S}	Start bit delay SCD		0		ns
t_{BUSY_R}	Start bit delay Register data		0		ns
t_{BUSY_E}	Start bit delay EEPROM data		$t_{READEEP}$		ns
$t_{TIMEOUT}$	Timeout → $CFGBISS$	$2/f_{OSZ}$	25	$2^{31}/f_{OSZ}$	us

Table 51: Characteristic values SPI interface

Symbol	Characteristic value	Min.	Typ.	Max.	Unit
$t_{HIGH}(\text{SCK})$	SPI-Clock, H time	20			ns
$t_{LOW}(\text{SCK})$	SPI-Clock, L time	20			ns
$t_s(\text{SEN})$	Setup time SEN falling before SCK rising	15			ns
$t_h(\text{SEN})$	Hold time SEN rising after SCK falling	15			ns
$t_s(\text{MOSI})$	Setup time $MOSI$ before SCK rising	5			ns
$t_h(\text{MOSI})$	Hold time $MOSI$ after SCK rising	5			ns
$t_D(\text{MISO})$	Delay time SCK falling until $MISO$ @CL = 12 pF			20	ns
$t_{ENA}(\text{MISO})$ ¹⁾	Delay time SEN falling until $MISO$ active			25	ns
$t_D(\text{nWAIT})$	Delay time SEN rising until $nWAIT$ active		60	70	ns
$t(\text{nWAIT-L})$	Waiting time after SEN rising	$2/f_{OSZ}$		$4/f_{OSZ} + 25$	ns
	Waiting time after SEN rising (synchronous read)	$2/f_{OSZ}$		$36/f_{OSZ} + 25$	ns
$t(\text{SEN-Wait})$	Time between wait state and next access	0			ns

¹⁾ for non-read commands, the pin $MISO$ may remain in the tristate state (inactive).

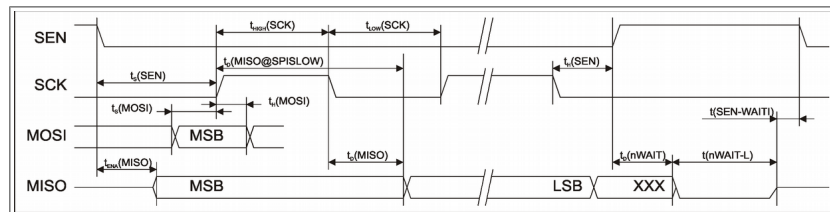


Figure 32: Timing SPI

Table 52: Characteristic values Simple-SPI-Master

Symbol	Characteristic value	Min.	Typ.	Max.	Unit
$t_{HIGH}(\text{SCK})$	SPI clock, H time		$1/f_{OSZ}$		ns
$t_{LOW}(\text{SCK})$	SPI clock, L time		$1/f_{OSZ}$		ns
$t_s(\text{MISO})$	Setup time $MISO$ before SCK rising	5			ns
$t_h(\text{MISO})$	Hold time $MISO$ after SCK rising	5			ns
$t_D(\text{MOSI})$	Delay time SCK falling until $MOSI$ @CL = 12 pF			20	ns
t_{TXENA}	Pulse width $TXENA$ (High)			TBD/f_{OSZ}	ns

11 Application notes

11.1 Application circuit

As the GC-NIP includes two fast A/D converters, the same design rules applicable to A/D converters must be applied. All block capacitors are to be connected closely to the pad. Please note that the quality of the sensor power supply also influences the measuring accuracy of standard sensors. If necessary, additional LC filters to the sensor power supply and to AVDD must be included.

The supply- and reference-voltage pins are to be connected as seen in 53; the connection of unused inputs is shown in 54.

Table 53: IC connection, voltages

Pin	Connection
VSSA	Ground analogue
VSS, VSSIO, Exposed Pad	Ground digital
VDDA	Supply voltage analogue 3.3V Block capacitor 100nF against VSSA
VDD, VDDIO	Supply voltage digital 3.3V Block capacitor 100nF against VSS/VSSIO
R1N,R1M,R1P,R2N,R2M,R2P	a block capacitor each 2.2µF against VSSA and a block capacitor each 10nF against VSSA
V0CH1,V0CH2	A block capacitor 100nF against VSSA open input possible if CFG2/DISV0 = 1

Table 54: IC connection of unused in-/outputs

Pin	Connection, if unused
NRES	Pull-Up 10k against VDDIO
SINN, COSN, REFN	V0
REFP	AVDD or AVSS
CFGAF, CFGGAIN, CFGDIR	VSSIO
PRESET, TXENA/TRG	VDDIO
MISO/SLO	Pull-Up 1k against VDDIO
MOSI/SLI	VSSIO
SCK/MA	VSSIO
SEN	VDDIO
NERR	Pull-Up 10k against VDDIO
TM	VSS

General notes:

- All block capacitors have to be connected closely to the pad.
- Separate ground areas for VSSA resp. for VSS and VSSIO must be used.
- The ground areas for VSSA and VSS/VSSIO must be connected at one point of the PCB.
- The pins NRES, NERR require a pull-up resistor of 10 kΩ each.
- The pin MISO/SLO requires a pull-up resistor of 1 kΩ.
- For using the SPI with high data rates, series resistors of 22...33 Ω each at MOSI, MISO, SCK and SEN are useful.
- The digital outputs A, B, and Z are designed for a maximum load of 6 mA. An external driver-IC is necessary to realize a differential RS422-interface. The outputs can be configured for tristate behaviour in case of error. Depending on the application, pull-up resistors are required.
- For the use of additional termination resistors between SINP and SINN respectively between COSP and COSN please refer to the application notes of the sensor manufacturer.
- Single-ended sensors are typically connected to the inputs SINP and COSP. The DC reference levels of the GC-NIP and of the sensor must concordant in this case.
- The signals V0CH1 and V0CH2 can be used as reference level. The current rating at this pins totals 1 mA. Short and low-capacity wires should be used. A buffer operational amplifier may be included, if necessary.
- For reliable operation of the IC, it is imperative to connect defined levels to the IC inputs. Internal pull-up resistors only prevent unpredictable behaviour of the IC with floating inputs.

The design of the analogue input circuit depends on the type of the sensor that is connected. The following figures show an example of one channel connected to different types of sensors:

Sensor with differential output signals **Sensor with a nominal amplitude of 1V_{pp} or 2V_{pp}**

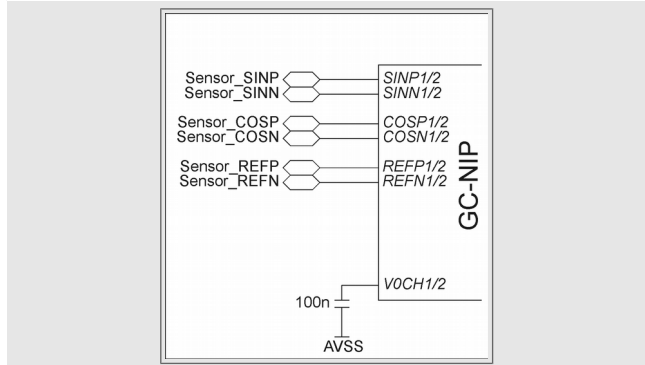


Figure 33: Sensor with differential output signals

- The amplitude of the sensor and the gain factor of the GC-NIP are adapted by the configuration bits GAIN (1 : 0)
- Reference level V0CH1/2 is generated internally.

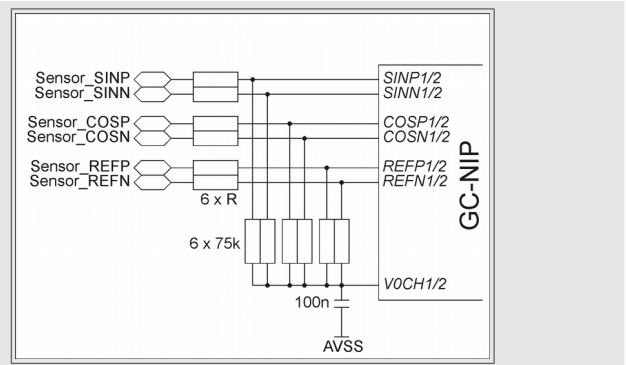


Figure 34: Sensor with a nominal amplitude of 1V_{pp} or 2V_{pp}

- The nominal amplitude of the GC-NIP is configured to the value 660 mV_{pp} using the bits GAIN (1 : 0).
- Reference level V0CH1/2 is generated internally.
- External resistors between the input signals and pin V0CH1/2 are used as voltage divider for the sensor signals. The value for the resistor is calculated as follows:
 $R = (V_{\text{Sensor}} / 660\text{mV} - 1) \cdot 75\text{k}\Omega$
- The amplitude of the sensor signals and the reference level will be divided using the ratio R/75kΩ.
- Alternatively, for sensors with 5V supply, the level shifter IC GC-LS can be used.

Sensor with single-ended output signals (I) **Sensor with single-ended output signals (II)**

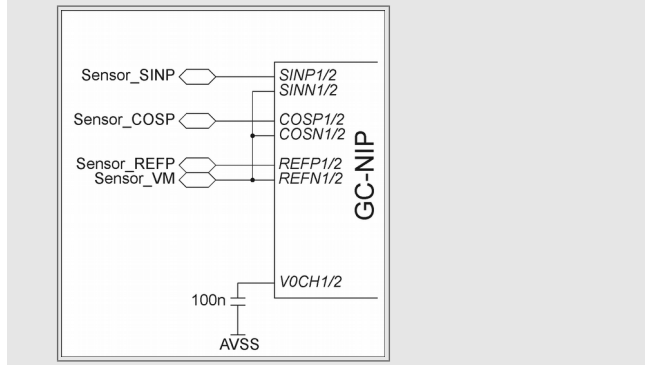


Figure 35: Sensor with single-ended output signals (I)

- The amplitude of the sensor and the gain factor of the GC-NIP are adapted by the configuration bits GAIN (1 : 0)
- Reference level VM is generated by sensor.

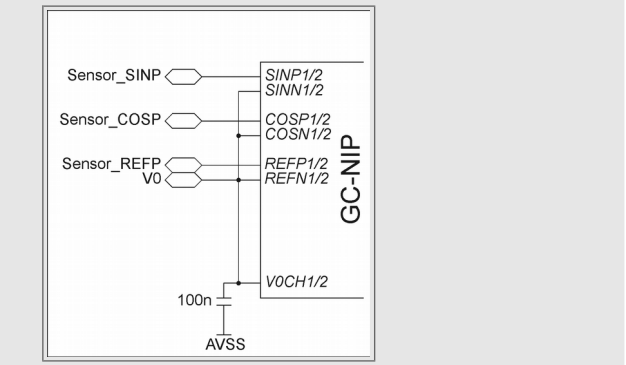


Figure 36: Sensor with single-ended output signals (II)

- The amplitude of the sensor and the gain factor of the GC-NIP are adapted by the configuration bits GAIN (1 : 0)
- Reference level V0CH1/2 is generated internally and is provided to the sensor.

**Sensor containing antiparallel photodiodes
Adjustment of amplitude equality possible**

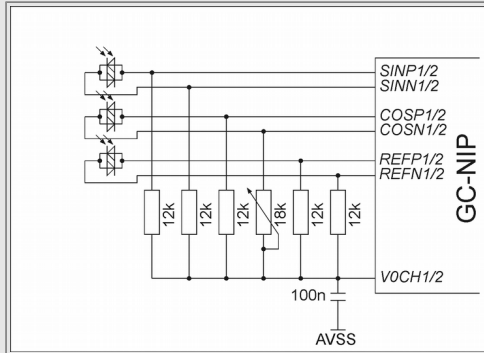


Figure 37: Sensor containing antiparallel photodiodes

- The nominal amplitude of the GC-NIP has to be set to 250mVpp by configuration bits `GAIN (1:0)`.
- Reference level `V0CH1/2` is generated internally.
- The amplitude equality is adjusted by changing the amplitude of the cosine signal. The pins `SMON` and `CMON` are used for the measurement.
- The value of the resistors has to be adjusted to the given sensor: $R_{FIX} = 250 \text{ mV} / (2 \cdot I_{SENSOR})$ and $P_{AMPL} \approx 1.5 \cdot R_{FIX}$
- Example: $I_{SENSOR} = 11 \mu A_{pp}$

**Array of photo diodes with common cathode or anode
Adjustment of amplitude equality and offset possible**

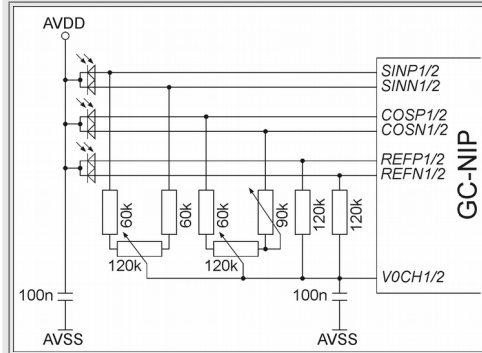


Figure 38: Array of photo diodes with common cathode or anode

- The nominal amplitude of the GC-NIP has to be set to 160mVpp by configuration bits `GAIN (1:0)`.
- Reference level `V0CH1/2` is generated internally.
- The amplitude equality is adjusted by changing the amplitude of the cosine signal. Thereafter the offset for both signals can be adjusted. The pins `SMON` and `CMON` are used for the measurement.
- The values of the resistor has to be adjusted to the given sensor:
- $R = 160\text{mV} / (2 \cdot I_{SENSOR})$.
- This resistor is partly designed as a potentiometer for the adjustment of the offset
 $P_{OFFS} \approx R$; $R_{FIX} \approx \frac{1}{2} R$; $P_{AMPL} \approx 1.5 \cdot R_{FIX}$
- Example: $I_{SENSOR} = 0.5 \mu A_{pp}$

Sensor for current signals 11 μA_{pp}

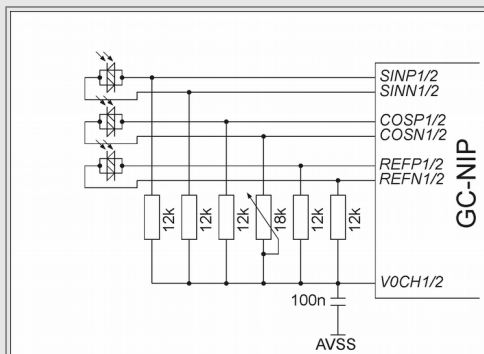
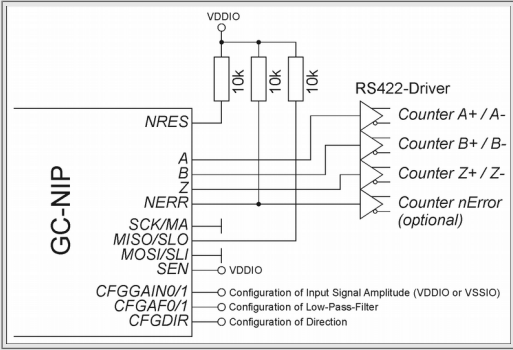
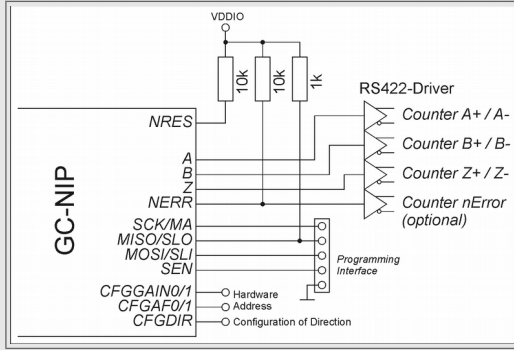
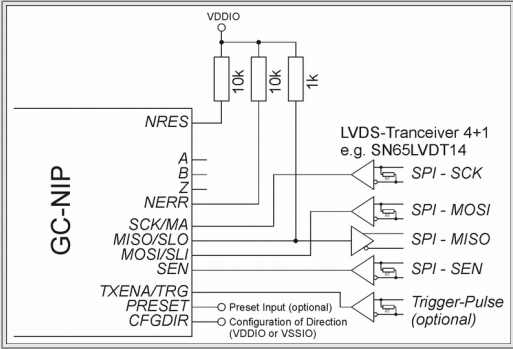
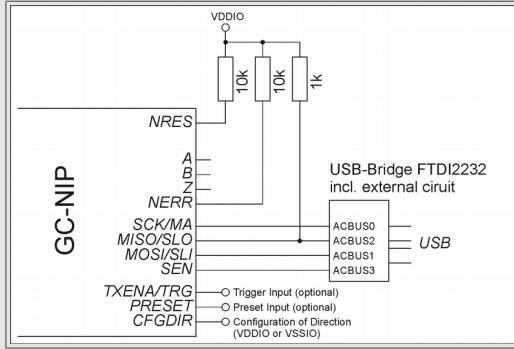


Figure 39: Sensor containing antiparallel photodiodes

- The nominal amplitude of the GC-NIP is set to 250mVpp by configuration bits `GAIN (1:0)`.
- Reference level `V0CH1/2` is generated internally.
- The value of the resistor R is dimensioned as follows:
 $R = 250\text{mV} / (2 \cdot I_{SENSOR})$
- Example: $I_{SENSOR} = 11 \mu A_{pp}$

The following figures show examples for the connection of the various interfaces:

ABZ Output / Configuration via pin	ABZ Output / Configuration via EEPROM
	
<p>Figure 40: ABZ Output / Configuration via pin</p> <ul style="list-style-type: none"> ■ The nominal amplitude and the low-pass filter frequency are set via configuration pins. All other configurations are set as in Table 8. ■ Connection of the second ABZ output is identical. 	<p>Figure 41: ABZ Output / Configuration via EEPROM</p> <ul style="list-style-type: none"> ■ The configuration of the IC is read from the integrated EEPROM. ■ The SPI interface is used as programming interface to the EEPROM ■ Short wires should be used at the pins MOSI, SEN and SCK. Otherwise, pull-up resistors (10 kΩ) are recommended. ■ Connection of the second ABZ output is identical.

SPI interface LVDS	SPI interface to PC via USB
	
<p>Figure 42: SPI interface LVDS</p> <ul style="list-style-type: none"> ■ The IC is configured from the internal EEPROM or via the SPI interface. ■ The LVDS driver IC enables long cable length at high clock frequencies. ■ The trigger pulse is provided as differential signal (optional). 	<p>Figure 43: SPI interface USB</p> <ul style="list-style-type: none"> ■ The IC is configured from the internal EEPROM or via the SPI interface. ■ SPI communication is realized using a bridge-IC with USB interface to the PC.

BiSS interface **SSI interface**

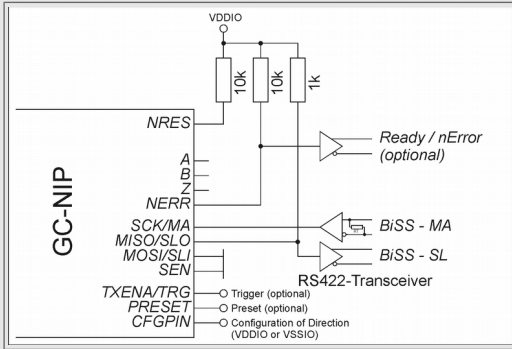


Figure 44: BiSS interface

- The configuration of the IC is read from the integrated EEPROM.
- BiSS is used as point-to-point connection.
- An optional signal indicates the ready/error status if the GC-NIP.

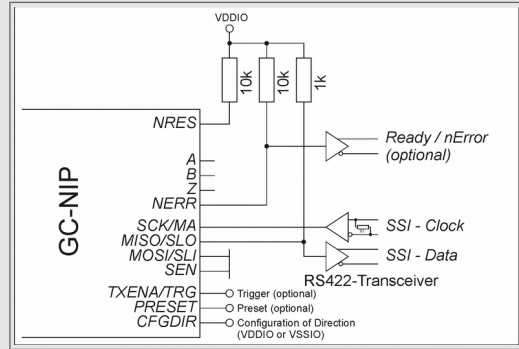


Figure 45: SSI interface

- The configuration of the IC is read from the integrated EEPROM.
- An optional signal indicates the ready/error status if the GC-NIP.

SPI interface to microcontroller **SPI interface to PC via USB**

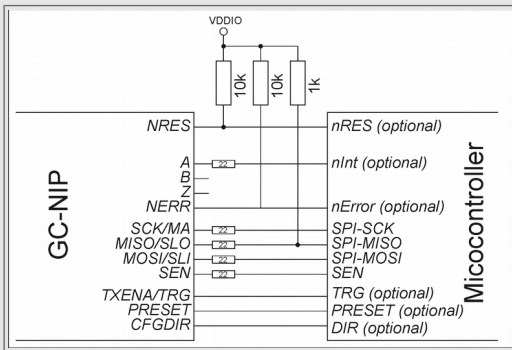


Figure 46: SPI interface to microcontroller

- The IC is configured from the internal EEPROM or via the SPI interface.
- The microcontroller firmware implements the SPI master for communication to the GC-NIP.
- An optional signal is used as trigger, another as interrupt to the microcontroller.
- Optionally, the controller is able to reset the GC-NIP.

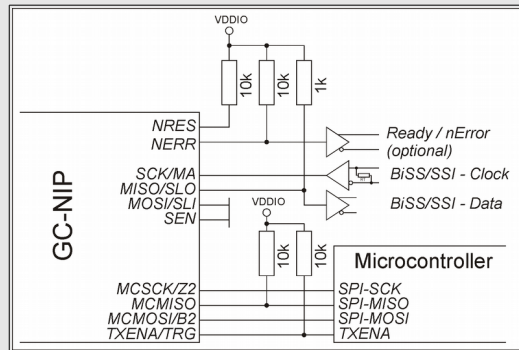


Figure 47: SSI/BiSS interface and Simple SPI master

- The configuration of the IC is read from the integrated EEPROM.
- The measurement values are output via the BiSS- or SSI-interface.
- An external microcontroller enables read and modify of the measurement value. The communication is controlled by the pin TXENA/TRG.

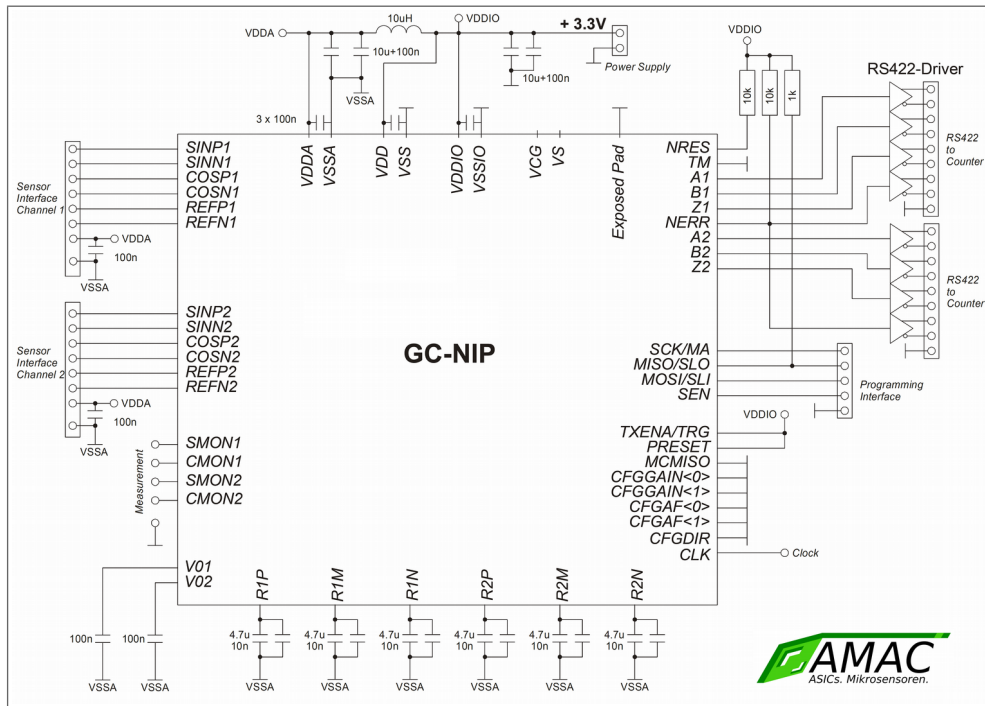


Figure 48: Minimum application circuit (principle)

- The configuration of the IC is read from the integrated EEPROM.
- The SPI interface is used for programming of the EEPROM and for calibration of the nonius scale.
- Short wires should be used at the pins MOSI, SEN and SCK. Otherwise, pull-up resistors (10 kΩ) are recommended.

For further information for connection of the IC and usage of the interfaces please request the detailed circuit of the evaluation board “AM-IPE-NONIUS” and the recommended PCB-layout via E-Mail to support@amac-chemnitz.de.

11.2 Fast equidistant measurements via SPI

Fast and/or equidistant measurements can be realized via the SPI interface:

Table 55: Equidistant measurements

Time base	Pin TRG	SPI interface	Remark
From the SPI-interface	For asynchronous trigger events	SYNC mode	Enabling SEN with a period of $N \cdot 96/f_{OSZ}$ allows equidistant measurement without jitter. Exact synchronization of several ICs is possible.
External	Time base	ASYN mode	Jitter: $96/f_{OSZ}$. Reading of the measures values via SPI must be completed within the measuring interval. Exact synchronization of several ICs is possible.

11.3 Program sequence examples

A measurement example using the trigger and measurement value register MVAL is shown in the following figure. The determination of the trigger values and the sensor monitoring is done using the register STAT:

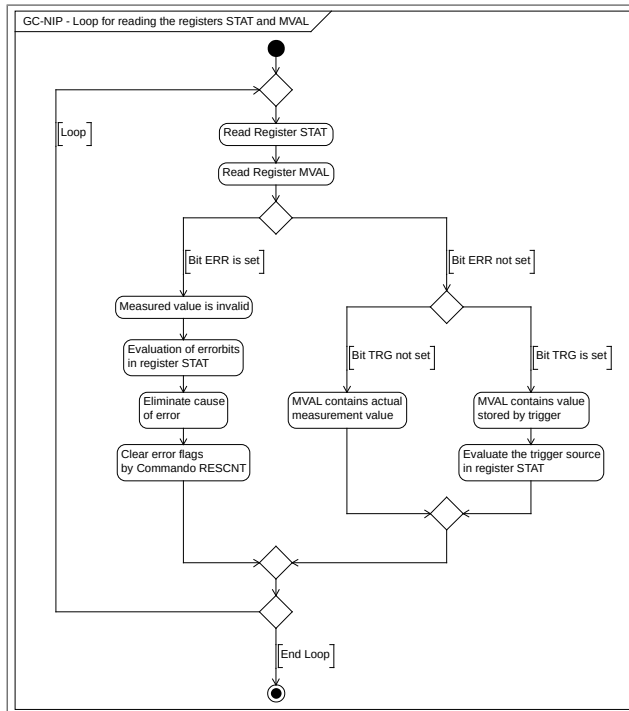


Figure 49: Program sequence for reading MVAL and STAT

For adjustment of the reference position and for evaluation of distance coded reference marks, the sequence can be extended (see sections 7.6.3 and 11.5 for further information):

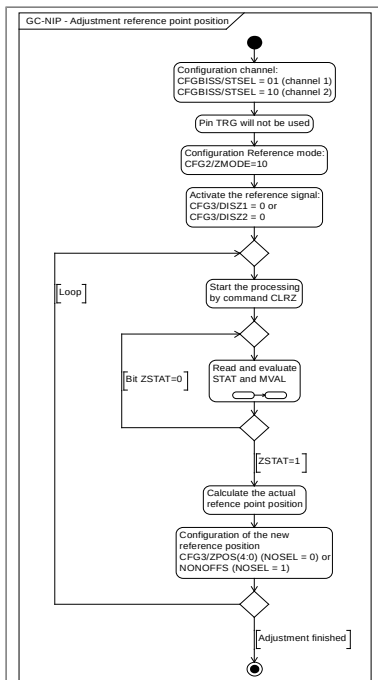


Figure 50: Extended programming sequence for ZMODE 10 see section 7.6.3 and 11.5

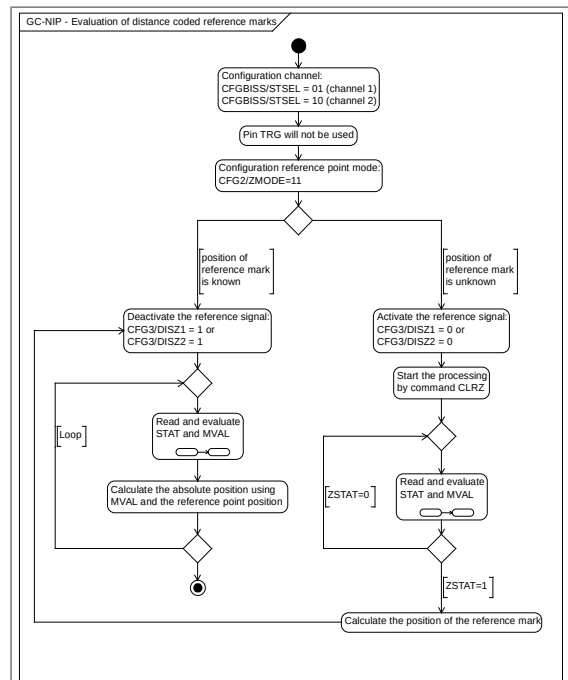


Figure 51: Extended programming sequence for ZMODE 11

11.4 EEPROM

Reading from and writing to the EEPROM is handled by an internal interface, which can be accessed using the register EEP :

Before any write access, the bit EEPBSY must contain the value '0'

Writing an OP-code to the register EEP OPC (Byte 3) trigger a EEPROM access. EEPADR and EEPDAT must be valid. Invalid OP-codes shall not be used.

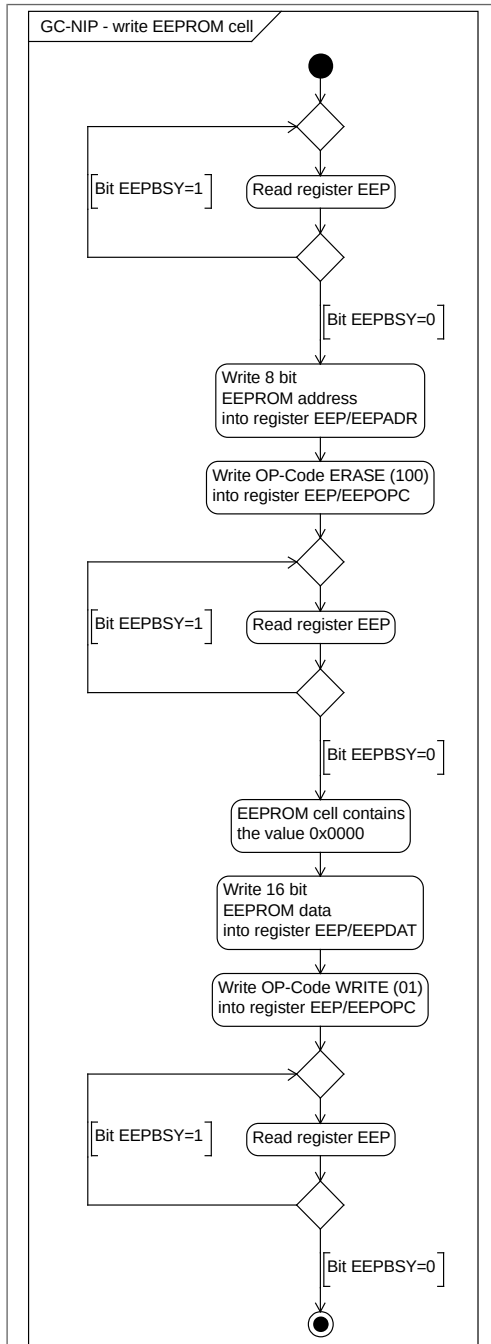


Figure 52: Programming sequence write/read EEPROM

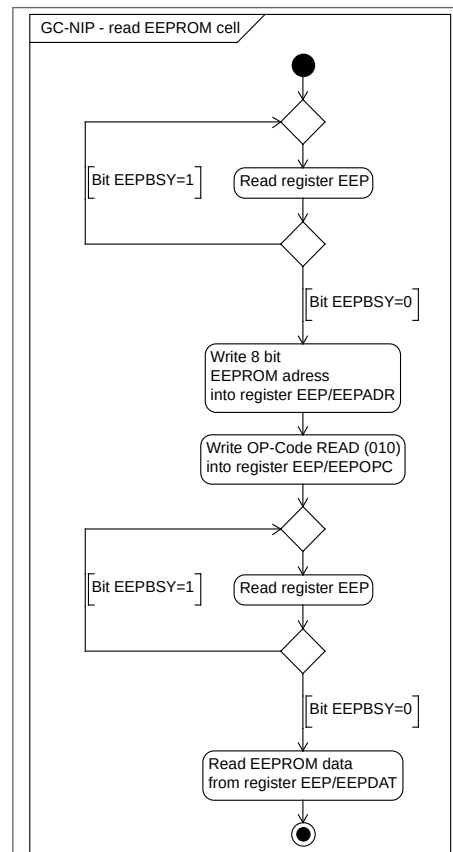


Figure 53: Programming sequence write/read EEPROM

11.5 Evaluation of distance-coded reference marks

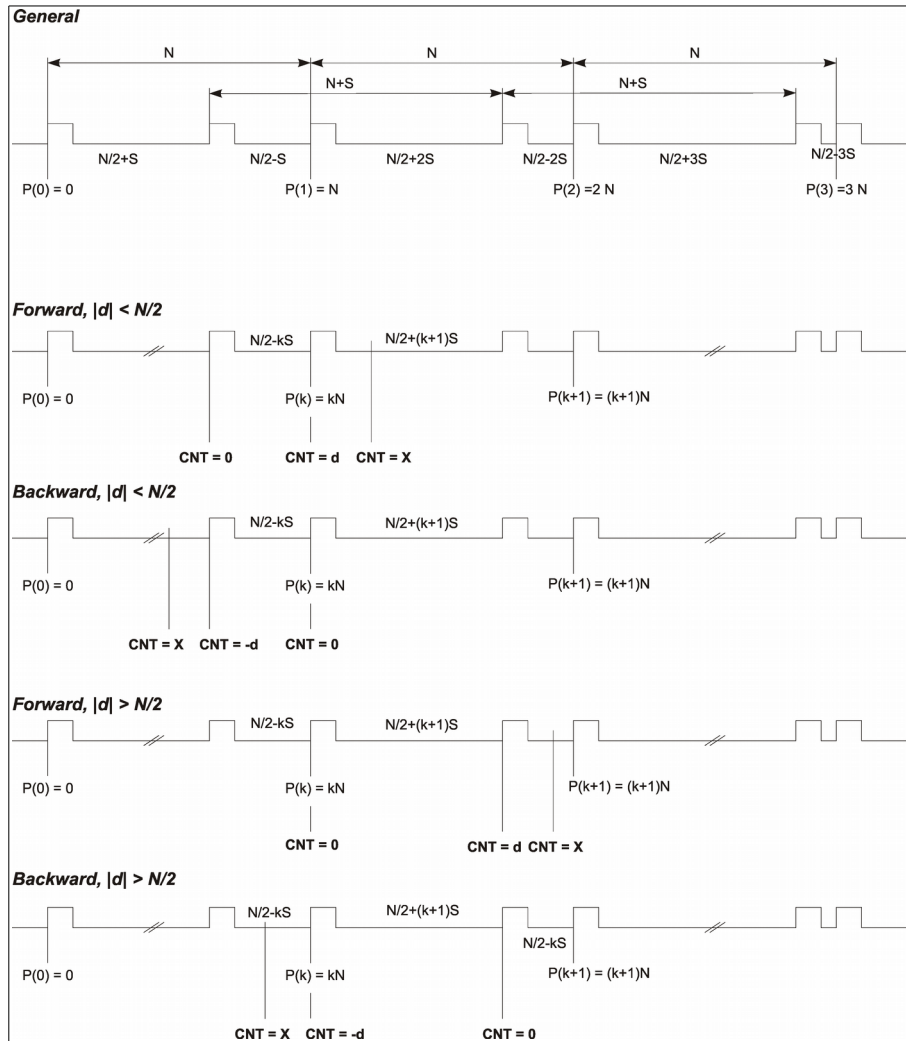


Figure 54: Evaluation of distance-coded reference marks

Table 56: Evaluation of distance-coded reference marks

$0 < d < \frac{1}{2} \cdot N$ Figure 54-1	$-\frac{1}{2} \cdot N < d < 0$ Figure 54-2	$\frac{1}{2} \cdot N < d$ Figure 54-3	$d < -\frac{1}{2} \cdot N$ Figure 54-4
$D = d/IRATE \cdot M$	$D = d/IRATE \cdot M$	$D = d/IRATE \cdot M$	$D = d/IRATE \cdot M$
$D = (N/2 - k) \cdot S$	$D = -(N/2 - k) \cdot S$	$D = N/2 + (k+1) \cdot S$	$D = -N/2 - (k+1) \cdot S$
$P = N/S \cdot (N/2 - D)$	$P = N/S \cdot (N/2 + D)$	$P = N/S \cdot (D - N/2) - N$	$P = -N/S \cdot (D + N/2) - N$
$PX = P - D + X/IRATE \cdot M$	$PX = P + X/IRATE \cdot M$	$PX = P + X/IRATE \cdot M$	$PX = P - D + X/IRATE \cdot M$

With:

- M: Scale graduation (mm)
- N: Segment length of the reference mark on the scale (mm)
- S: Reference point step width on the scale (mm)
- k: Reference mark number on the scale
- P(k): Absolute position of the reference mark k (mm)
- d: Triggered reference mark distance (increments)
- D: Triggered reference mark distance (mm)
- X: Counter value (increments)
- PX: Absolute position of the sensor (mm)

11.6 Configuring t_{pp} and f_{OSZ}

The GC-NIP is configured according to the requirements of the sensor and of the subsequent electronics. Also see sections 7.3 and 7.4

Table 57: Configuration t_{pp} and speed monitoring

ABZ output used?			
no		yes	
CFG1/MABZ=0 CFG1/TPP (2:0) any value		CFG1/MABZ=1 CFG1/MFAST=1 Condition: $t_{pp}(\text{counter at ABZ}) < t_{pp}(\text{GC-NIP})$	
Interpolation counter used ?		Oscillator frequency specified?	
no	yes	no	yes
CFG1/MFAST=0 $f_{MAX} \geq 150\text{kHz}$	CFG1/MFAST=1 $f_{MAX} = f_{OSZ} / 280$	CFG1/TPP (2:0) any value typical: CFG1/TPP (2:0) = '001' $N = 2^{CFG1/TPP(2:0)}$ $4\text{ MHz} \leq f_{OSZ} < N/t_{pp}(\text{counter at ABZ}) \leq 26\text{ MHz}$	$N = 2^{CFG1-TPP(2:0)} > t_{pp}(\text{counter at ABZ}) \cdot f_{OSZ}$
		$t_{pp}(\text{GC-NIP}) = N / f_{OSZ}$ $f_{MAX} < 0.9 \cdot f_{OSZ} \cdot \text{IRDIV} / (N \cdot \text{IRATE})$ and $f_{MAX} < f_{OSZ} / 280$	

Example a)

The minimum edge distance of the electronics connected to A, B and Z is 250 ns ($t_{pp}(\text{counter at ABZ})$)
 The interpolation rate (IRATE) is 4000, IRDIV is configured to '10' (4), (interpolation rate ABZ = 1000).
 The maximum input frequency is 2 kHz.
 The oscillator frequency can be selected freely within the range 4 MHz ... 26 MHz.

CFG1/MFAST = 1 CFG1/MABZ = 1 CFG1/TPP (2:0) = '001' → N = 2 $f_{OSZ} < 2/250\text{ns}$, $f_{OSZ} < 8\text{ MHz}$ $2\text{kHz} < 0.9 \cdot f_{OSZ} \cdot 4 / (2 \cdot 4000)$, $4.44\text{ MHz} < f_{OSZ}$ → 4.44 MHz < f_{OSZ} < 8 MHz	CFG1/MFAST = 1 CFG1/MABZ = 1 CFG1/TPP (2:0) = '010' → N = 4 $f_{OSZ} < 4/250\text{ns}$, $f_{OSZ} < 16\text{ MHz}$ $2\text{kHz} < 0.9 \cdot f_{OSZ} \cdot 4 / (4 \cdot 4000)$, $8.88\text{ MHz} < f_{OSZ}$ → 8.88 MHz < f_{OSZ} < 16 MHz
---	---

Example b)

The minimum edge distance of the electronics connected to A, B, and Z is 150 ns. ($t_{pp}(\text{counter at ABZ})$)
 The interpolation rate (IRATE) is 2000, IRDIV is configured to '11' (8) (interpolation rate ABZ = 250).
 The oscillator frequency is 26 MHz.
 The maximum input frequency is determined on the basis of the specified parameters.

CFG1/MFAST = 1 CFG1/MABZ = 1 $N = 2^{CFG1/TPP(2:0)} > 150\text{ ns} \cdot 26\text{ MHz} \rightarrow N > 3.9$ CFG1/TPP (2:0) = '010' → N = 4 $f_{MAX} = 0.9 \cdot 26\text{ MHz} \cdot 8 / (4 \cdot 2000)$ $f_{MAX} = 23.4\text{ kHz}$
--

Table 58: t_{pp} at GC-NIP

CFG1/TPP(2:0)	N	t_{pp}
000	0	$t_{pp} = 1 / f_{OSZ}$
001	2	$t_{pp} = 2 / f_{OSZ}$
010	4	$t_{pp} = 4 / f_{OSZ}$
011	8	$t_{pp} = 8 / f_{OSZ}$
100	16	$t_{pp} = 16 / f_{OSZ}$
101	32	$t_{pp} = 32 / f_{OSZ}$
110	64	$t_{pp} = 64 / f_{OSZ}$
111	128	$t_{pp} = 128 / f_{OSZ}$

11.7 Configuration of the interface SPI/BiSS/SSI

The interface of the GC-NIP is configured according to the requirements of the interface master and the data format of the position value. See sections 8.1, 8.2, 8.3 and 7.8.

Example a) SPI-Mode

Maximum data rate for reading the counter value and the status information.

Configuration	Value	Details
CFG2/ASYNC	1	Maximum data rate
CFG2/SYNC	any value	Any value for SYNC (6:0) for reading CNT1 and CNT2
CFGBISS/STSEL	any value	Register POSIT is not used
CFGBISS/STBIT	any value	Register POSIT is not used
CFGBISS/GRAY	any value	Register POSIT is not used
CFGBISS/READ32	any value	BiSS-Interface inactive
CFGBISS/BISSTO	any value	BiSS-Interface inactive
CFGBISS/RING	any value	SSI-Interface inactive
CFGBISS/SSITO	any value	SSI-Interface inactive
CFGBISS/SSI20	any value	SSI-Interface inactive

Example b) SPI-Mode

Reading of all data registers using a software timer.

Configuration	Value	Details
CFG2/ASYNC	0	(equidistant) measurement triggered by software timer
CFG2/SYNC	64dez	Reading of some registers requires this value
CFGBISS/STSEL	00bin	Register POSIT contains the absolute position (nonius), Register CNT1 and CNT2 contains the incremental value of the interpolation counters
CFGBISS/STBIT	30dez	The maximum number of bits is used
CFGBISS/GRAY	0	Usually, binary data is transferred via the SPI interface
CFGBISS/READ32	any value	BiSS-Interface inactive
CFGBISS/BISSTO	any value	BiSS-Interface inactive
CFGBISS/RING	any value	SSI-Interface inactive
CFGBISS/SSITO	any value	SSI-Interface inactive
CFGBISS/SSI20	any value	SSI-Interface inactive

Example c) BiSS-C-Mode

The measurement value is binary coded

The clock frequency of the GC-NIP is 26MHz.

Register access via the BiSS interface is only used for reading and writing configuration registers.

Configuration	Value	Details
CFG2/ASYNC	any value	No register access to data (measurement) registers.
CFG2/SYNC	any value	No register access to data (measurement) registers.
CFGBISS/STSEL	00bin	The BiSS-SCD (and register POSIT) provides the nonius result.
CFGBISS/STBIT	30dez	30 Bit single-turn data, two leading zero bits are added to get the total length of 32 bit.
CFGBISS/GRAY	0	Binary code
CFGBISS/READ32	0	No register access to data (measurement) registers.
CFGBISS/BISSTO	9	BiSS-Timeout = $512/26 \text{ MHz} = 19.7 \mu\text{s}$
CFGBISS/RING	any value	SSI-Interface inactive
CFGBISS/SSITO	any value	SSI-Interface inactive
CFGBISS/SSI20	any value	SSI-Interface inactive

Example d) SSI-Mode 20 bit

The measurement value is transferred in gray-code.

The clock frequency of the GC-NIP is 8MHz.

The SSI-Master operates in ring-mode with a timeout of 18 µs.

Configuration	Value	Details
CFG2/ASYNC	any value	any value for SSI
CFG2/SYNC	any value	any value for SSI
CFGBISS/STSEL	00bin	The SSI data (and register <code>POSIT</code>) provides the nonius result.
CFGBISS/STBIT	30dez	The maximum number of bits is used
CFGBISS/GRAY	1	Gray code
CFGBISS/READ32	any value	BiSS-Interface inactive
CFGBISS/BISSTO	any value	BiSS-Interface inactive
CFGBISS/RING	1	Ring operation possible
CFGBISS/SSITO	144dez	SSI timeout = 144 / 8 MHz = 18µs
CFGBISS/SSI20	1	SSI interface is working with 20 bit

11.8 BiSS configuration file *idbiss4743.xml*

To enable auto detection of the GC-NIP on a BiSS master device, the configuration file *idbiss4743.xml* can be used. For the detection of the data format of the single-cycle-data (SCD), it is recommended to program the BiSS vendor identifier according to the data format selected in `CFGBISS/STSEL` (see the following table and section 7.8).

CFGBISS/STSEL	Recommended vendor identifier	SCD (Pos 0)	SCD (Pos 1)	SCD (Pos 2)	SCD (Pos 3)
00bin	0x32 0x03 0x00 0x00	10 bit unused	22 bit Nonius	1 bit error	1 bit warning
01bin	0x32 0x03 0x01 0x00	2 bit unused	30 bit counter 1	1 bit error	1 bit warning
10bin	0x32 0x03 0x02 0x00	2 bit unused	30 bit counter 2	1 bit error	1 bit warning
11bin	0x32 0x03 0x03 0x00	16 bit counter 1	16 bit counter 2	1 bit error	1 bit warning

